

ESE370: Circuit-Level Modeling, Design, and Optimization for Digital Systems

Lec 6: February 6, 2023
 MOS Transistor Operating Regions
 Part 2, Parasitics



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Today

- Operating Regions
 - Resistive
 - Saturation
 - Subthreshold
 - Velocity Saturation
- Short Channel Effects
 - V_{th}
 - Drain Induced Barrier Lowering
- Capacitance

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Velocity Saturation



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Carrier Velocity

- Model assumes carrier velocity increases with field
 - Increases with voltage proportionally to mobility

$$v = \mu_n \cdot \frac{V_{DS}}{L_{eff}} = \left(\frac{\mu_n}{L_{eff}} \right) V_{DS}$$

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Carrier Velocity

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Preclass 1

- (a) What is the electrical field in the channel?

$$L_{eff} = 25nm, V_{DS} = 1V$$

$$\text{Uniform Field} = \frac{V_{DS}}{L_{eff}}$$

- Velocity:

$$v = F \cdot \mu_n$$

- Electron mobility: $\mu_n = 500cm^2 / (V \cdot s)$

- (b) What is the electron velocity?

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Moving Charge

$$I = \left(\frac{1}{R}\right)V$$

- I increases linearly in V
- What's I?

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Moving Charge

$$I = \left(\frac{1}{R}\right)V$$

- I increases linearly in V
- What's I?
 - $\Delta Q/\Delta t$
 - Speed at which charge moves

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Moving Charge

$$I = \left(\frac{1}{R}\right)V$$

- I increases linearly in V

- What's I?
 - $\Delta Q/\Delta t$
 - Speed at which charge moves

$$\text{Field} = \frac{V_{DS}}{L_{eff}}, v = \mu_n \cdot F$$

$$v = \mu_n \cdot \frac{V_{DS}}{L_{eff}} = \left(\frac{\mu_n}{L_{eff}}\right)V_{DS}$$

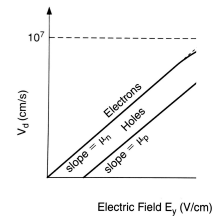
- Velocity increases linearly in V

- What's a moving electron?

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Carrier Velocity

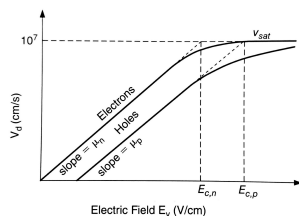


- Velocity –
 - increases for increasing field with slope of mobility

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Carrier Velocity



- Velocity –
 - increases for increasing field with slope of mobility
 - saturates for increasing field
 - More likely to hit the critical field in short channel

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Short Channel

- Model assumes carrier velocity increases with field
 - Increases with voltage proportionally to mobility
- There is a limit to how fast carriers can move
 - Limited by scattering effects
 - $\sim 10^8$ m/s
- Encounter *velocity saturation* when channel short
 - Modern processes, L is short enough to reach this region of operation

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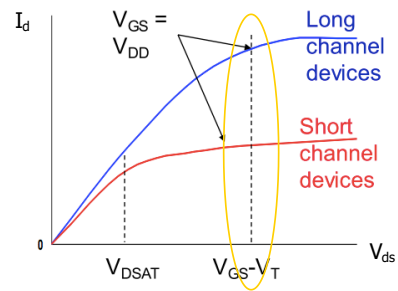
Velocity Saturation (Preclass 1)

- (c) At what voltage do we hit the speed limit 10^5m/s ?
 - $L_{\text{eff}}=25\text{nm}$, $V_{\text{ds}}=1\text{V}$
 - V_{DSAT} = voltage at which velocity (current) saturates

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Velocity Saturation



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Velocity Saturation

- Our current model equation:

$$I_{DS} = \mu_n C_{OX} \left(\frac{W}{L} \right) \left[(V_{GS} - V_{th}) V_{DS} - \frac{V_{DS}^2}{2} \right]$$

- Once velocity saturates:

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Velocity Saturation

- Our current model equation:

$$I_{DS} = \mu_n C_{OX} \left(\frac{W}{L} \right) \left[(V_{GS} - V_{th}) V_{DS} - \frac{V_{DS}^2}{2} \right]$$

- Once velocity saturates:

$$V_{DS} = V_{DSAT} \Rightarrow I_{DS} = \mu_n C_{OX} \left(\frac{W}{L} \right) \left[(V_{GS} - V_{th}) V_{DSAT} - \frac{V_{DSAT}^2}{2} \right]$$

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Velocity Saturation

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$$I_{DS} = \left(\mu_n \frac{V_{DSAT}}{L} \right) C_{OX} W \left[(V_{GS} - V_{th}) - \frac{V_{DSAT}}{2} \right]$$

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Velocity Saturation

- Our current model equation:

$$I_{DS} = \mu_n C_{OX} \left(\frac{W}{L} \right) \left[(V_{GS} - V_{th}) V_{DS} - \frac{V_{DS}^2}{2} \right]$$

- Once velocity saturates:

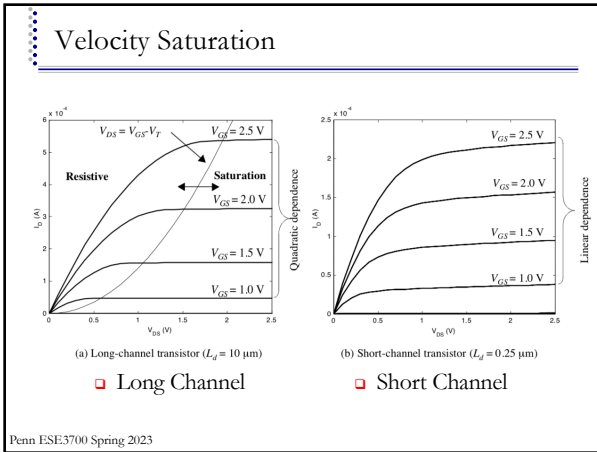
$$V_{DS} = V_{DSAT} \Rightarrow I_{DS} = \mu_n C_{OX} \left(\frac{W}{L} \right) \left[(V_{GS} - V_{th}) V_{DSAT} - \frac{V_{DSAT}^2}{2} \right]$$

$$I_{DS} = \left(\mu_n \frac{V_{DSAT}}{L} \right) C_{OX} W \left[(V_{GS} - V_{th}) - \frac{V_{DSAT}}{2} \right]$$

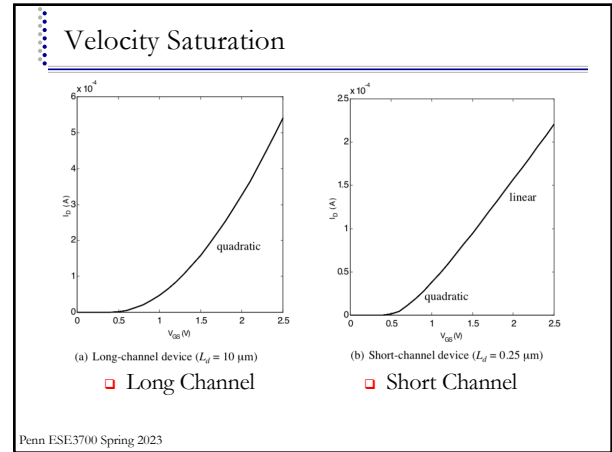
$$I_{DS} \approx v_{sat} C_{OX} W \left[(V_{GS} - V_{th}) - \frac{V_{DSAT}}{2} \right]$$

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Velocity Saturation

- Once velocity saturates we can still increase current with parallelism
 - Effectively make a wider device

$$I_{DS} \approx v_{sat} C_{OX} W \left[(V_{GS} - V_{th}) - \frac{V_{DSAT}}{2} \right]$$

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Threshold

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Short Channel Effects – V_T Reduction

Long Channel Device Short

$V_{T0}(\text{short channel}) = V_{T0} - \Delta V_{T0}$

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Short Channel Effects – V_T Reduction

Long-channel threshold

Threshold as a function of the length (for low V_{DS})

Short

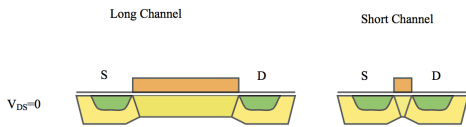
$V_{T0}(\text{short channel}) = V_{T0} - \Delta V_{T0}$

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Short Channel Effects - DIBL

- Drain Induced Barrier Lowering
 - V_T Reduction with Drain Bias

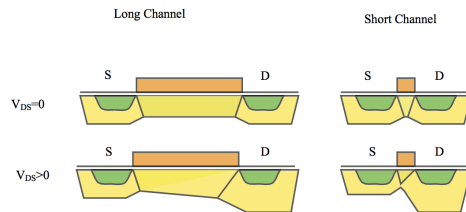


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Short Channel Effects - DIBL

- Drain Induced Barrier Lowering
 - V_T Reduction with Drain Bias

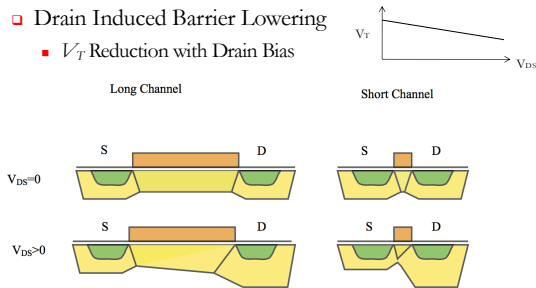


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Short Channel Effects - DIBL

- Drain Induced Barrier Lowering
 - V_T Reduction with Drain Bias



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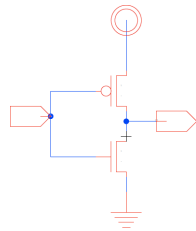
Threshold Reduction Impact



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In a Gate?

- What does it impact most?
 - Which device, has large V_{ds} ?
 - How does this effect operation?
 - Speed of switching?
 - Leakage?



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Capacitance



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Simplified Design Flow

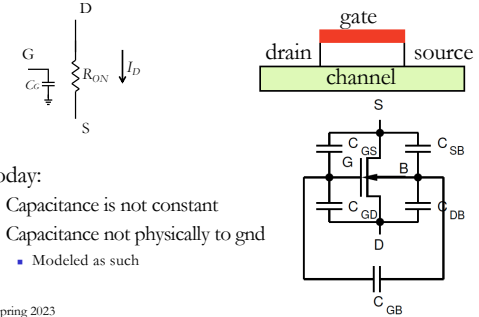
- Design a circuit to perform a function with specified minimum speed and optimized power (minimized with an upper bound)
 - Zero order model to design topology
 - First order model to meet speed spec
 - Rise/fall times, propagation delay, gate capacitance, output stage equivalent resistance
 - Transistor IV curves
 - Iterative SPICE simulation – tweak knobs to optimize for power (switching (dynamic), leakage (static), etc.)

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Capacitance

- First order: gate input looks like a capacitor



- Today:
 - Capacitance is not constant
 - Capacitance not physically to gnd
 - Modeled as such

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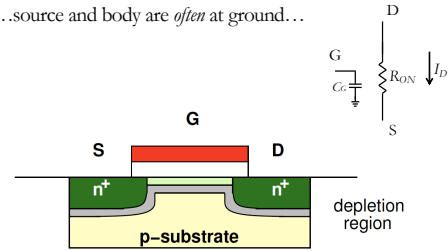
Capacitance Setup



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Capacitance

- Modeled gate with a capacitor to ground
- ...but ground isn't really one of our terminals
 - Don't connect directly to it
 - ...source and body are *often* at ground...

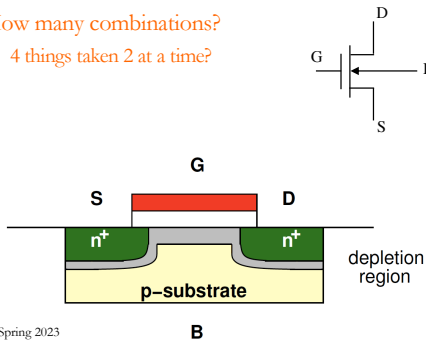


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Capacitance (Preclass 2)

- Four Terminals
- How many combinations?
 - 4 things taken 2 at a time?

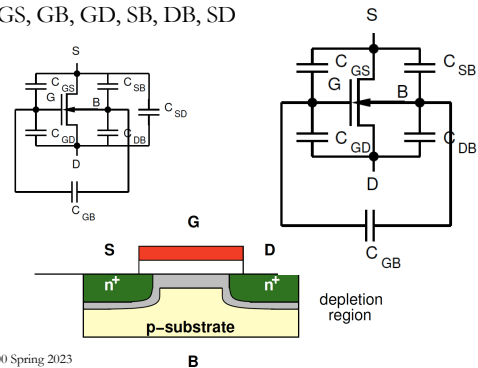


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Capacitances

- GS, GB, GD, SB, DB, SD



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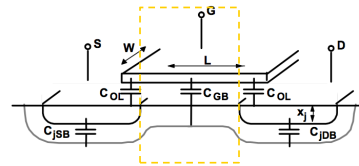
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Capacitance Decomposition



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MOSFET Parasitic Capacitance



- Any two conductors separated by an insulator form a parallel-plate capacitor
- Two types
 - Extrinsic – Outside the box (e.g. junction, overlap)
 - Intrinsic – Inside the box (e.g. gate-to-channel)

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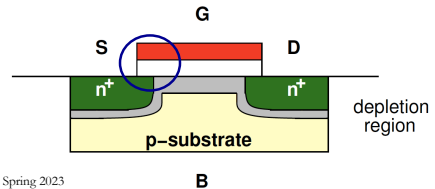
Overlap Capacitance



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Overlap

- What is the capacitive implication of gate/source and gate/drain overlap?

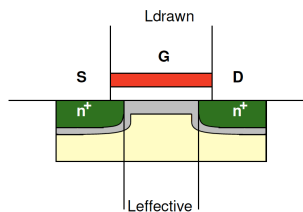


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Overlap

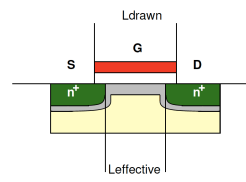
- Length of overlap?



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Overlap Capacitance



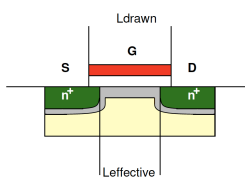
$$C = \epsilon_r \epsilon_0 \frac{A}{d}$$

$$C_o = \epsilon_{ox} \frac{W(L_{drawn} - L_{effective})}{t_{ox}}$$

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Overlap Capacitance



$$C = \epsilon_r \epsilon_0 \frac{A}{d}$$

$$C_{OX} = \frac{\epsilon_{OX}}{t_{OX}}$$

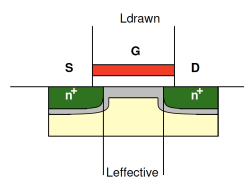
$$C_o = \epsilon_{ox} \frac{W(L_{drawn} - L_{effective})/2}{t_{ox}}$$

$$C_o = \frac{1}{2} C_{ox} W (L_{drawn} - L_{effective})$$

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Overlap Capacitance



$$C = \epsilon_r \epsilon_0 \frac{A}{d}$$

$$C_{OX} = \frac{\epsilon_{OX}}{t_{OX}}$$

$$C_o = \epsilon_{ox} \frac{W(L_{drawn} - L_{effective})/2}{t_{ox}}$$

$$C_o = \frac{1}{2} C_{ox} W (L_{drawn} - L_{effective}) = C_{GSO} = C_{GDO}$$

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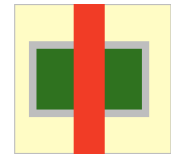
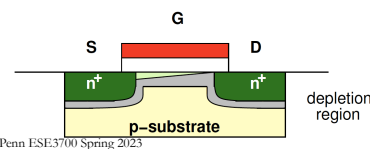
Junction Capacitances



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Junction (diffusion) Capacitance

- n⁺ contacts are formed by doping = diffusion
- Depletion under diffusion region (bottom-plate)
 - Due to reverse biased PN junction
 - Bottom-plate junction capacitance, C_j
- Depletion around perimeter (sidewall) of diffusion region
 - Sidewall junction capacitance, C_{jsw}

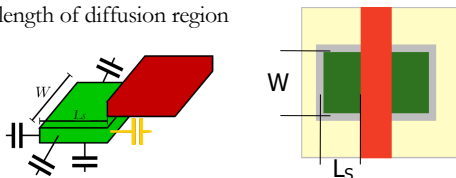


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Junction (Diffusion) Capacitance

- C_j – Bottom-plate junction capacitance (F/Area)
- C_{jsw} – Sidewall junction capacitance (F/Length)
- L_S – length of diffusion region



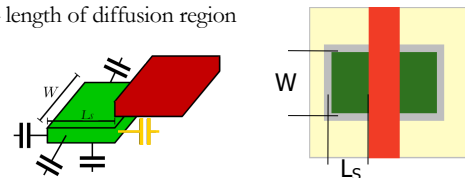
$$C_{diff} = C_j L_S W +$$

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Junction (Diffusion) Capacitance

- C_j – Bottom-plate junction capacitance (F/Area)
- C_{jsw} – Sidewall junction capacitance (F/Length)
- L_S – length of diffusion region



$$C_{diff} = C_j L_S W + C_{jsw} (2L_S + W)$$

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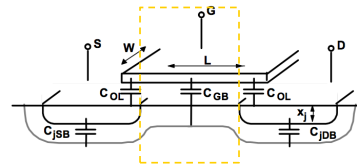
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Gate-to-channel



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MOSFET Parasitic Capacitance



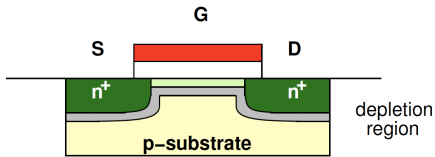
- Any two conductors separated by an insulator form a parallel-plate capacitor
- Two types
 - **Extrinsic** – Outside the box (e.g. junction, overlap)
 - **Intrinsic** – Inside the box (e.g. gate-to-channel)

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Gate-to-Bulk Capacitance

- Looks like parallel plate capacitance
- Two components:
 - What is C_{GC} ? (C_{GCS}, C_{GCD})
 - What is C_{GB} ?

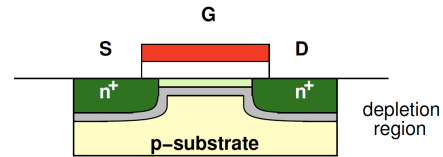


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Gate-to-Channel Capacitance

- Looks like parallel plate capacitance
- Two components: **Case: Strong Inversion (small V_{ds})**
 - C_{GC}
 - C_{GB}



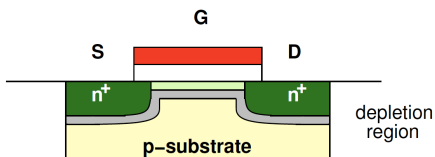
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Gate-to-Channel Capacitance

- Looks like parallel plate capacitance
- Two components: **Case: Strong Inversion**
 - C_{GC}
 - $C_{GCB}=0$

$$C_{GC} = C_{ox}WL_{effective}$$



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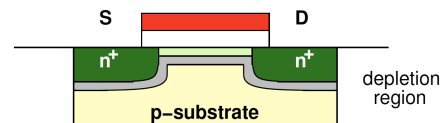
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Gate-to-Channel Capacitance

- Looks like parallel plate capacitance
- Two components: **Case: Strong Inversion**
 - C_{GC} – Split evenly between S and D
 - $C_{GCB}=0$

$$C_{GC} = C_{ox}WL_{effective}$$

$$C_{GCS} = C_{GCD} = \frac{1}{2}C_{ox}WL_{effective}$$



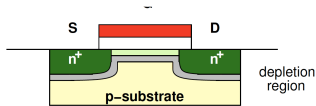
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Gate-to-Source Capacitance

- Channel + Overlap

$$C_{GS} = C_{GCS} + C_{GSO}$$



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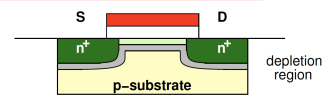
Gate-to-Source Capacitance

- Channel + Overlap

$$C_{GS} = C_{GCS} + C_{GSO}$$

$$C_{GS} = \frac{1}{2} C_{OX} W (L_{drawn} - L_{effective}) + \frac{1}{2} C_{OX} W L_{effective}$$

$$C_{GS} = \frac{1}{2} C_{OX} W L_{drawn}$$



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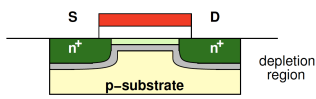
Gate-to-Drain Capacitance

- Channel + Overlap

$$C_{GD} = C_{GCD} + C_{GDO}$$

$$C_{GD} = \frac{1}{2} C_{OX} W (L_{drawn} - L_{effective}) + \frac{1}{2} C_{OX} W L_{effective}$$

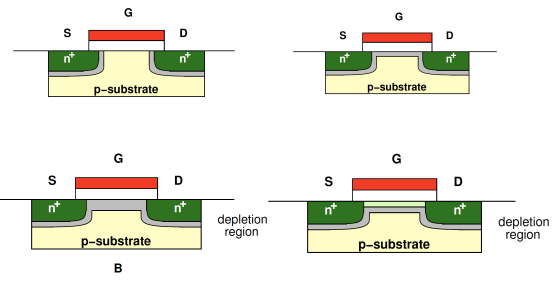
$$C_{GD} = \frac{1}{2} C_{OX} W L_{drawn}$$



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Channel Evolution: Weak Inversion

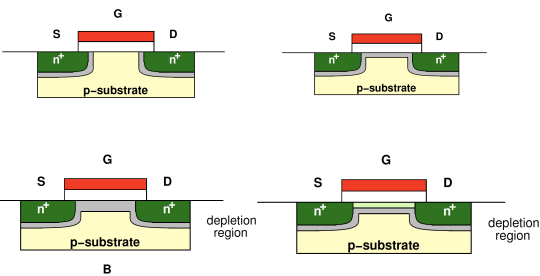


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Channel Evolution: Weak Inversion

- $V_{GS} = 0 \rightarrow C_{GC} = 0, C_{GCB} = W L C_{OX}$

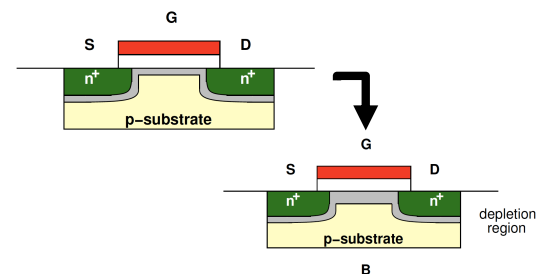


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Channel Evolution: Weak Inversion

- What happens to capacitance here as V_{GS} increases?
 - Capacitor plate distance?

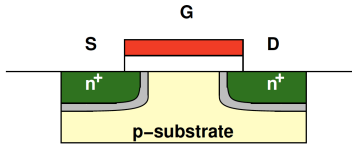


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Channel Evolution: Weak Inversion

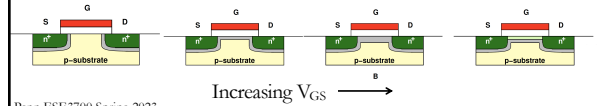
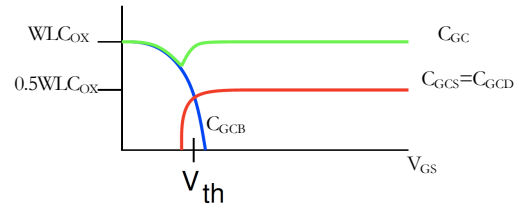
- Capacitance is initially dominated by Gate-to-bulk capacitance ($C_{GCS,D}=0$)
- Gate-to-bulk capacitance drops as V_{GS} increases toward V_{th}



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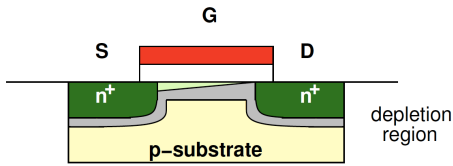
Capacitance vs V_{GS} ($V_{DS}=0$)



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Saturation Capacitance?

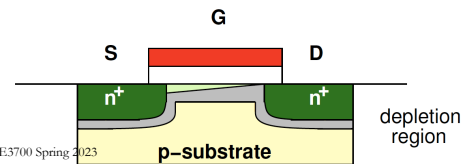


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Saturation Capacitance?

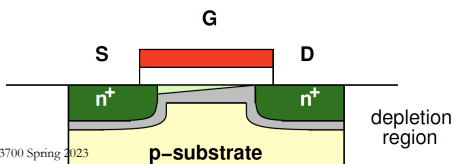
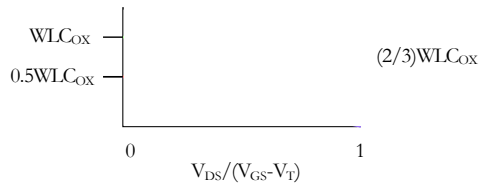
- Source end of channel in inversion
- Voltage at drain end of channel at or below threshold
- Capacitance shifts to source
 - Total capacitance reduced



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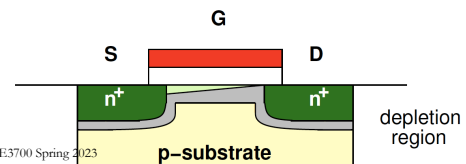
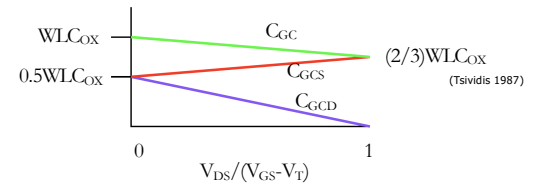
Saturation Capacitance



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Saturation Capacitance

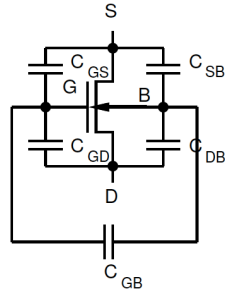


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Capacitance Roundup

- $C_{GS} = C_{GCS} + C_{GSO}$
- $C_{GD} = C_{GCD} + C_{GDO}$
- $C_{GB} = C_{GCB}$
- $C_{SB} = C_{diff}$
- $C_{DB} = C_{diff}$



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First Order Capacitance Summary

Operation Region	C_{GCB}	C_{GCS}	C_{GCD}	C_{GC}	C_G
Subthreshold					
Linear					
Saturation					

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First Order Capacitance Summary

Operation Region	C_{GCB}	C_{GCS}	C_{GCD}	C_{GC}	C_G
Subthreshold					
Linear	0	$C_{ox}WL/2$	$C_{ox}WL/2$		
Saturation					

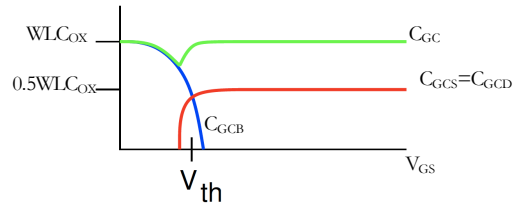
$$C_{GCS} = C_{GCD} = \frac{1}{2} C_{ox} WL_{effective}$$

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First Order Capacitance Summary

Operation Region	C_{GCB}	C_{GCS}	C_{GCD}	C_{GC}	C_G
Subthreshold	$C_{ox}WL$	0	0		
Linear	0	$C_{ox}WL/2$	$C_{ox}WL/2$		
Saturation					

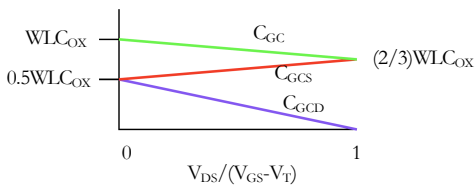


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First Order Capacitance Summary

Operation Region	C_{GCB}	C_{GCS}	C_{GCD}	C_{GC}	C_G
Subthreshold	$C_{ox}WL$	0	0		
Linear	0	$C_{ox}WL/2$	$C_{ox}WL/2$		
Saturation	0	$(2/3)C_{ox}WL$	0		



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First Order Capacitance Summary

Operation Region	C_{GCB}	C_{GCS}	C_{GCD}	C_{GC}	C_G
Subthreshold	$C_{ox}WL$	0	0	$C_{ox}WL$	
Linear	0	$C_{ox}WL/2$	$C_{ox}WL/2$	$C_{ox}WL$	
Saturation	0	$(2/3)C_{ox}WL$	0	$(2/3)C_{ox}WL$	

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First Order Capacitance Summary

Operation Region	C_{GCB}	C_{GCS}	C_{GCD}	C_{GC}	C_G
Subthreshold	$C_{ox}WL$	0	0	$C_{ox}WL$	$C_{ox}WL+2C_o$
Linear	0	$C_{ox}WL/2$	$C_{ox}WL/2$	$C_{ox}WL$	$C_{ox}WL+2C_o$
Saturation	0	$(2/3)C_{ox}WL$	0	$(2/3)C_{ox}WL$	$(2/3)C_{ox}WL+2C_o$

$$C_o = \frac{1}{2} C_{ox} W (L_{drawn} - L_{effective}) = C_{GSO} = C_{GDO}$$

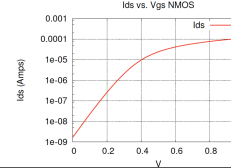
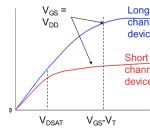
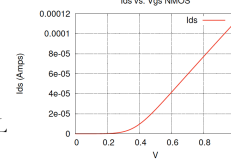
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Big Idea

3+ Regions of operation for MOSFET

- Subthreshold
- Linear
 - Pinch Off
- Saturation
- Velocity Saturation, DIBL
 - Short channel

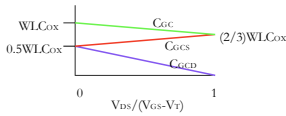
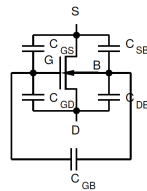
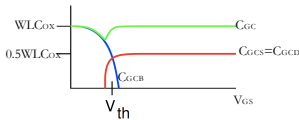


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Big Idea

- Capacitance
 - To every terminal
 - Voltage dependent



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Admin

- HW3 out now – due 2/10 (Friday)
 - Takes time! Learning curve for how to debug
 - Don't forget the demo/video of SPICE workflow
- Monday Lecture Cancelled 2/13
- Midterm 1 Postponed to Wednesday 2/15
 - 1:30pm-3:30pm (Tentative) in LRSM 112B
 - See Ed Discussion
 - Midterm 1 Review session 2/8
 - See Ed Discussion

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Acknowledgement

- Prof. André DeHon (University of Pennsylvania)
- Prof. Tania Li (University of Pennsylvania)

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One Implication (Optional)

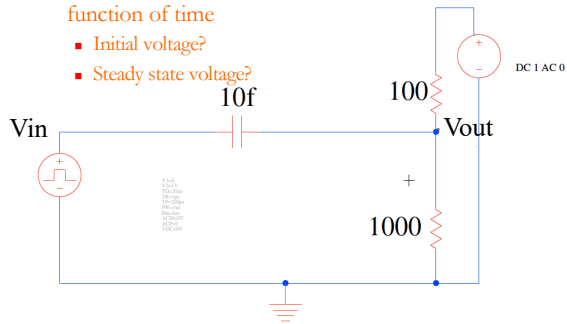
Feedback Capacitance C_{gd}



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Step Response? (Preclass 3)

- V_{in} steps from 0 to 1, what does V_{out} look like as a function of time
 - Initial voltage?
 - Steady state voltage?

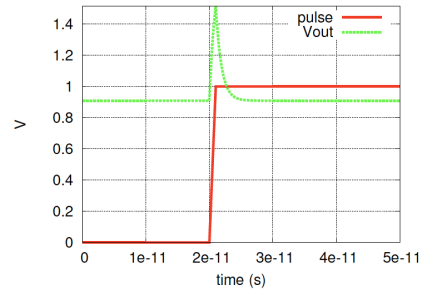


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Step Response

Voltage peaking!

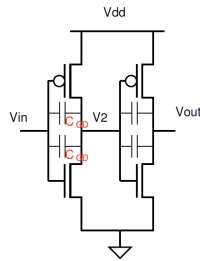


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Impact of C_{GD}

- What does C_{GD} do to the switching response here?
 - V_2
 - V_{out}

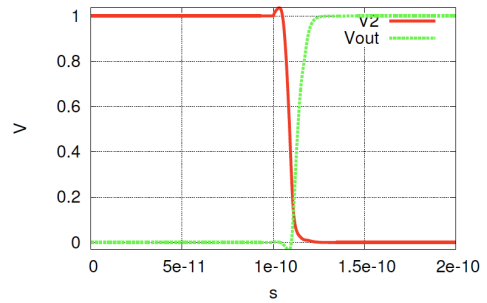


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Impact of C_{GD}

*** spice deck for cell flat_inv[sch] from library test



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