

ESE 531: Digital Signal Processing

Lec 13: February 27th, 2018
All-Pass Systems and Min Phase
Decomposition



Lecture Outline

- ❑ Review: Frequency Response of LTI Systems
 - Magnitude Response, Phase Response, Group Delay
 - Examples:
 - Zero on Real Axis
 - 2nd order IIR
 - 3rd order Low Pass
- ❑ Stability and Causality
- ❑ All Pass Systems
- ❑ Minimum Phase Systems



Review: Frequency Response of LTI System

$$Y(e^{j\omega}) = H(e^{j\omega})X(e^{j\omega})$$

- We can define a magnitude response...

$$|Y(e^{j\omega})| = |H(e^{j\omega})| |X(e^{j\omega})|$$

- a phase response...

$$\angle Y(e^{j\omega}) = \angle H(e^{j\omega}) + \angle X(e^{j\omega})$$

- and group delay

$$\text{grd}[H(e^{j\omega})] = -\frac{d}{d\omega} \{\arg[H(e^{j\omega})]\}$$

Review: Group Delay Math

$$H(z) = \frac{b_0 \prod_{k=1}^M (1 - c_k z^{-1})}{a_0 \prod_{k=1}^N (1 - d_k z^{-1})}$$

$$H(e^{j\omega}) = \frac{b_0 \prod_{k=1}^M (1 - c_k e^{-j\omega})}{a_0 \prod_{k=1}^N (1 - d_k e^{-j\omega})}$$

□ Look at each factor:

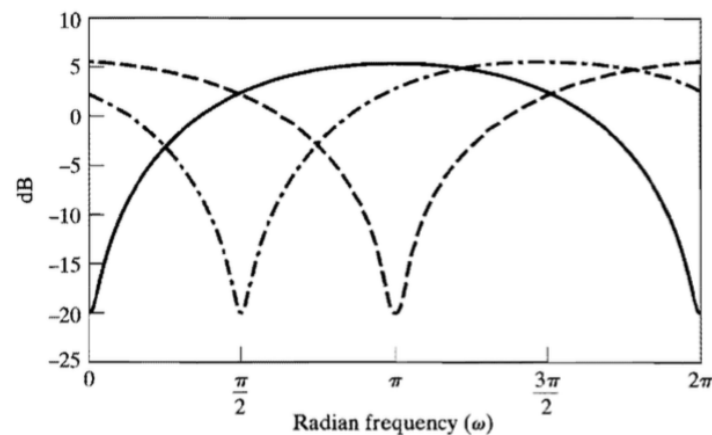
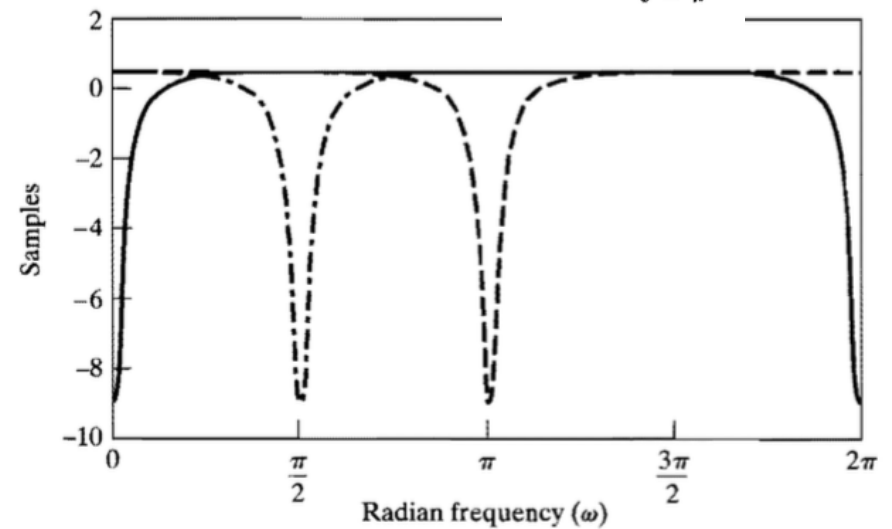
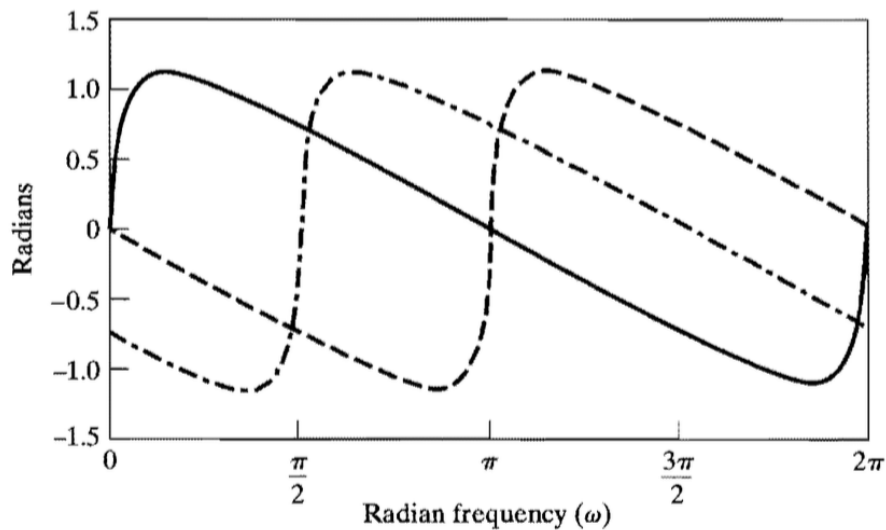
$$\arg[1 - re^{j\theta} e^{-j\omega}] = \tan^{-1} \left(\frac{r \sin(\omega - \theta)}{1 - r \cos(\omega - \theta)} \right)$$

$$\text{grd}[1 - re^{j\theta} e^{-j\omega}] = \frac{r^2 - r \cos(\omega - \theta)}{|1 - re^{j\theta} e^{-j\omega}|^2}$$

Example: Zero on Real Axis

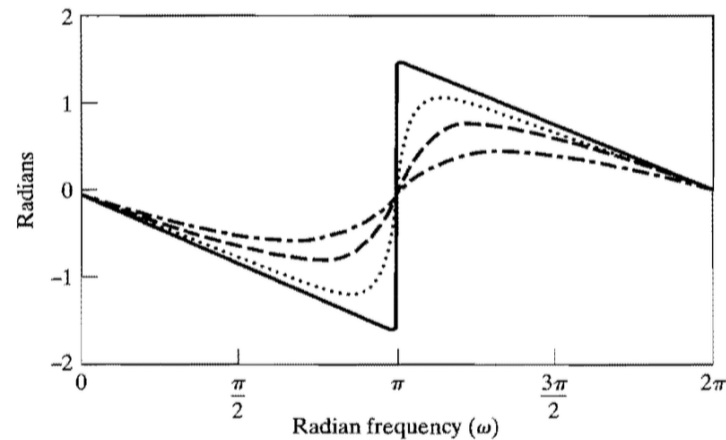
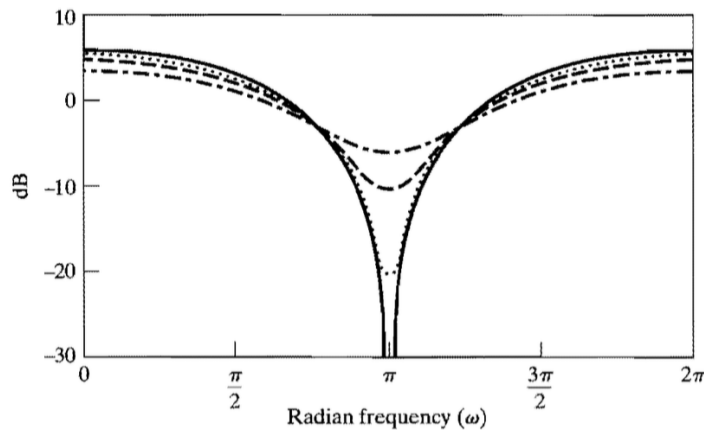
□ For $\theta \neq 0$ $H(e^{j\omega}) = 1 - re^{j\theta}e^{-j\omega}$

— $\theta = 0$
 - - - $\theta = \frac{\pi}{2}$
 - - - $\theta = \pi$



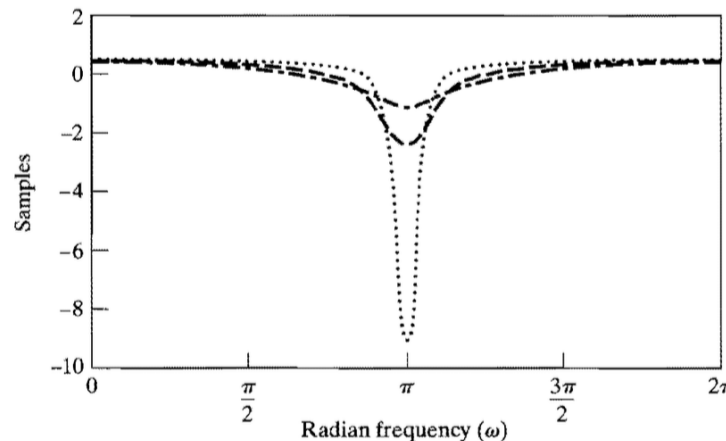
Example: Zero on Real Axis

- For $\theta = \pi$, how zero location affects magnitude, phase and group delay



$$H(e^{j\omega}) = 1 - re^{j\theta} e^{-j\omega}$$

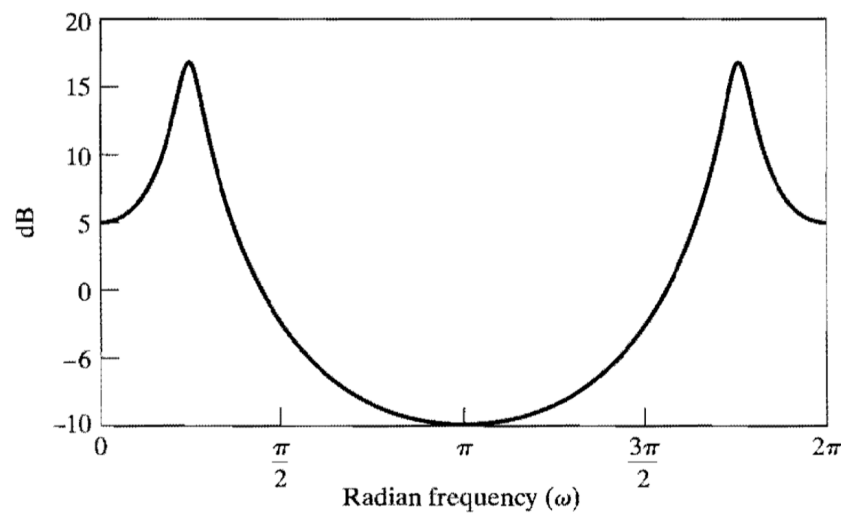
$$H(e^{j\omega}) = 1 + re^{-j\omega}$$



2nd Order IIR with Complex Poles

$$H(z) = \frac{1}{(1 - re^{j\theta}z^{-1})(1 - re^{-j\theta}z^{-1})} \quad r=0.9, \theta = \pi/4$$

magnitude

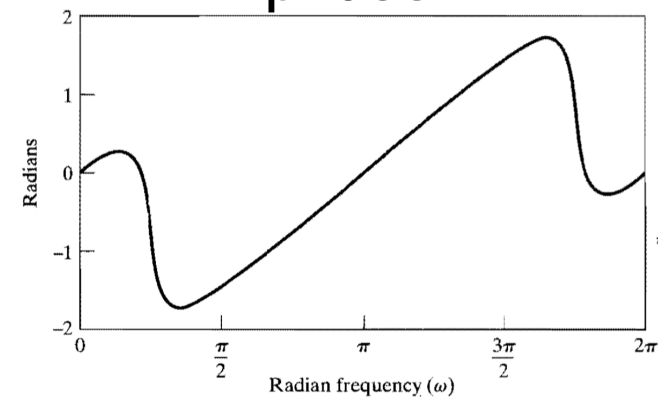
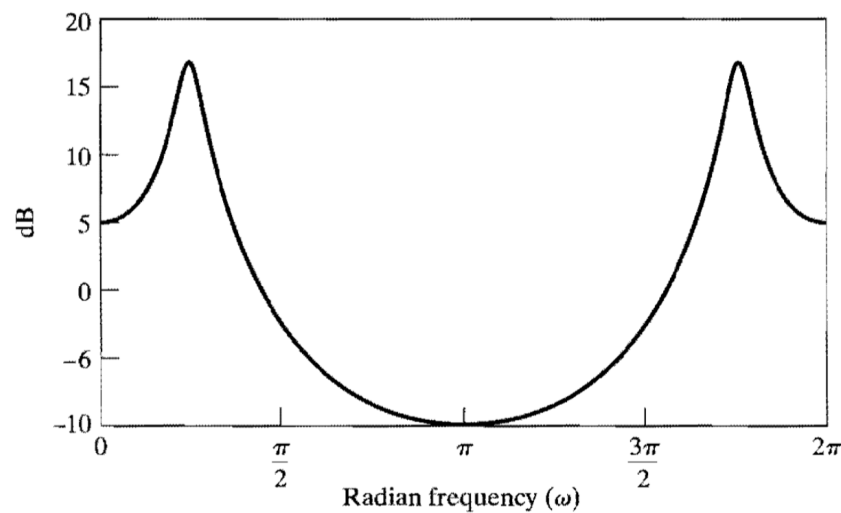


2nd Order IIR with Complex Poles

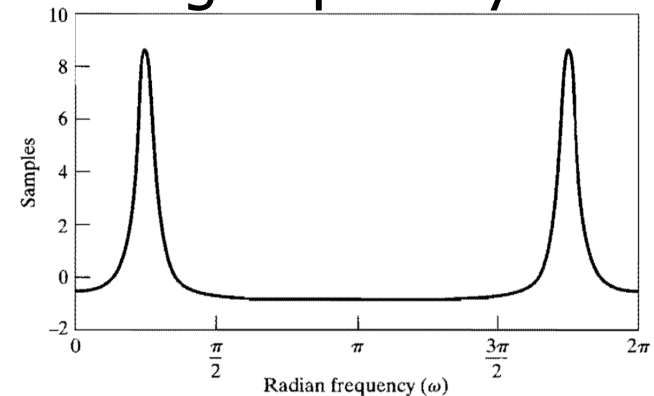
$$H(z) = \frac{1}{(1 - re^{j\theta}z^{-1})(1 - re^{-j\theta}z^{-1})}$$

$r=0.9$, $\theta = \pi/4$
phase

magnitude

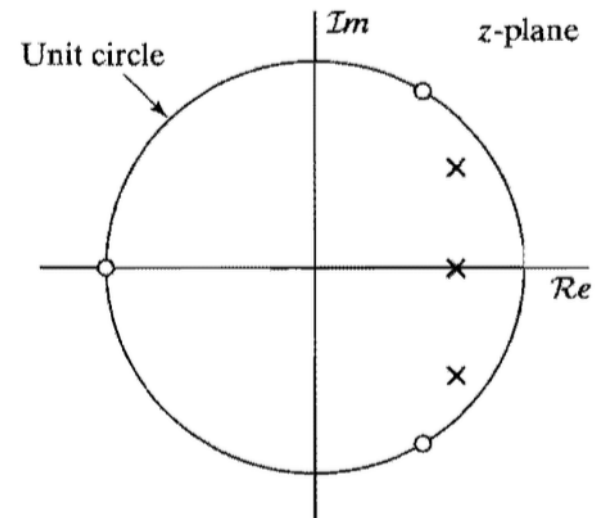


group delay



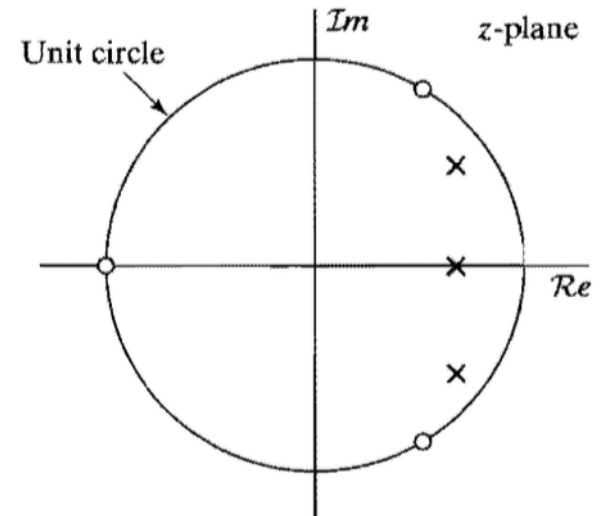
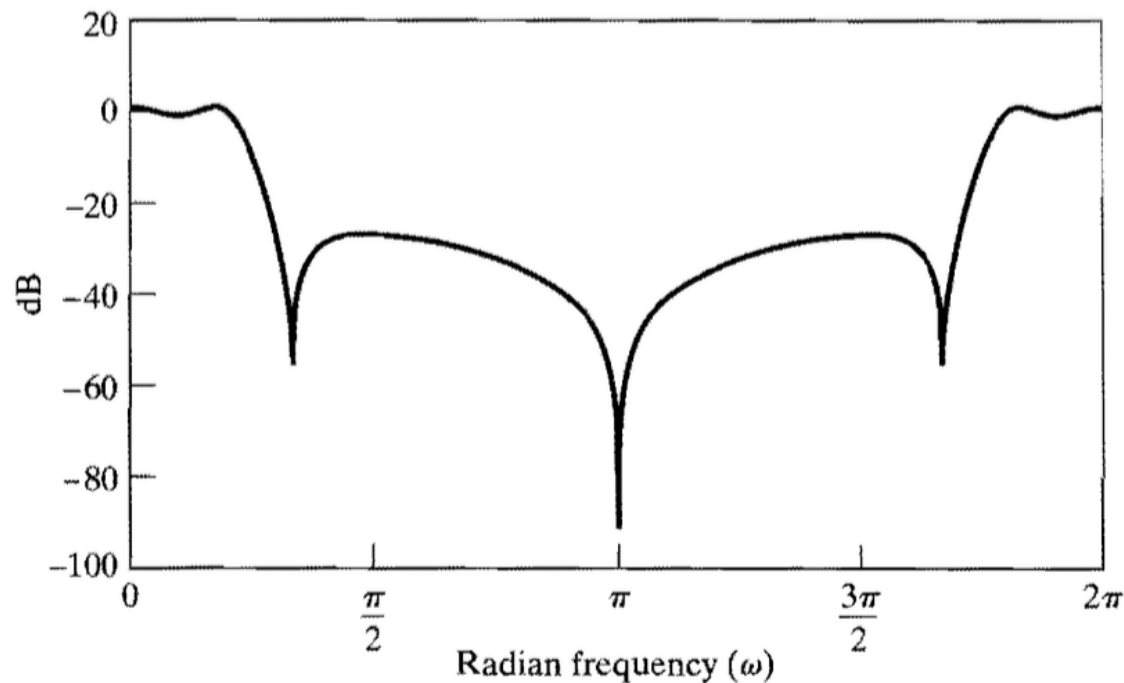
3rd Order IIR Example

$$H(z) = \frac{0.05634(1 + z^{-1})(1 - 1.0166z^{-1} + z^{-2})}{(1 - 0.683z^{-1})(1 - 1.4461z^{-1} + 0.7957z^{-2})}$$



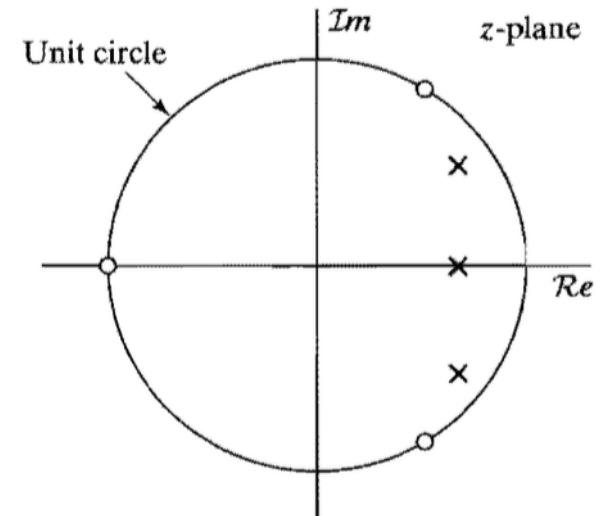
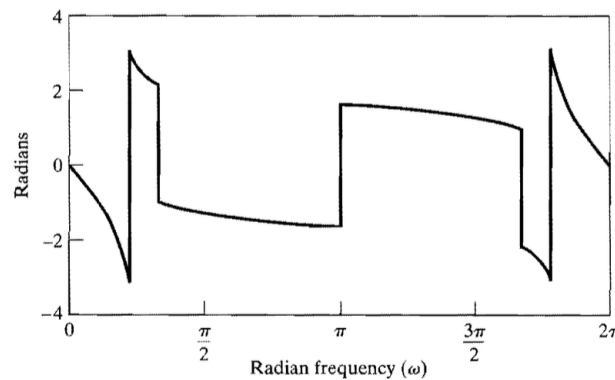
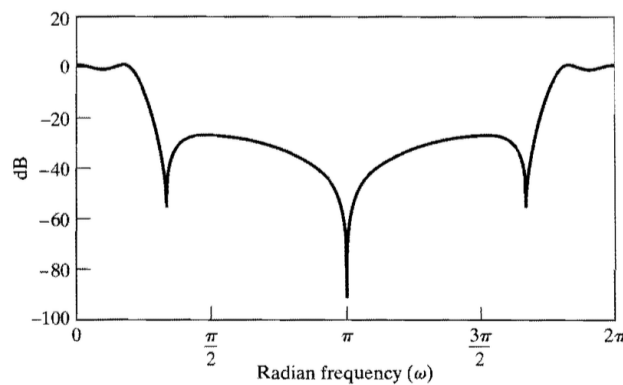
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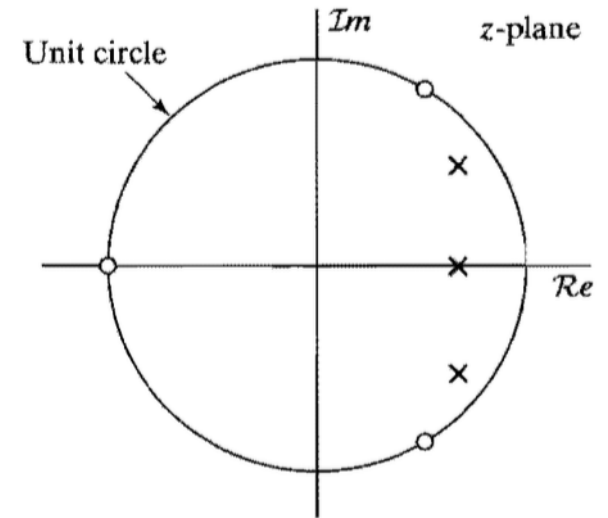
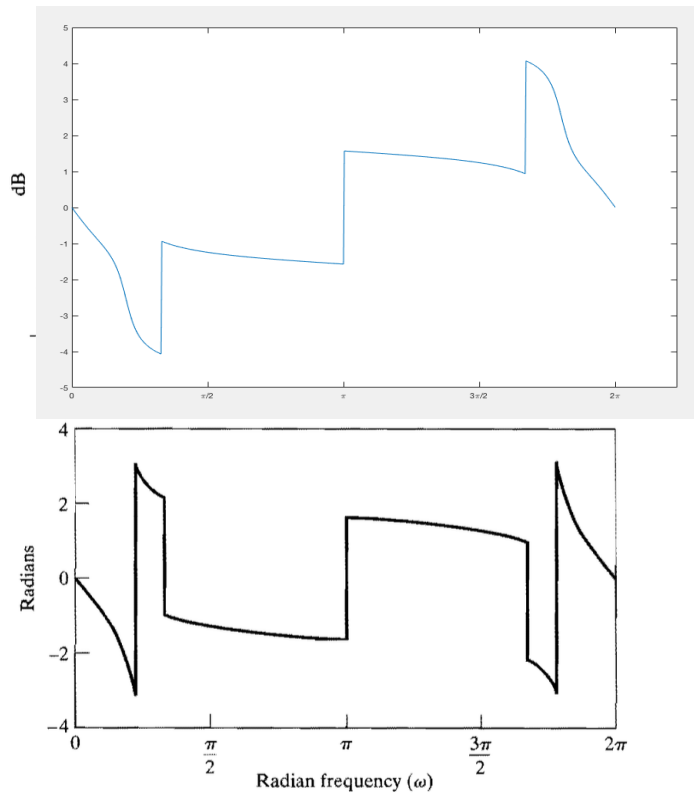
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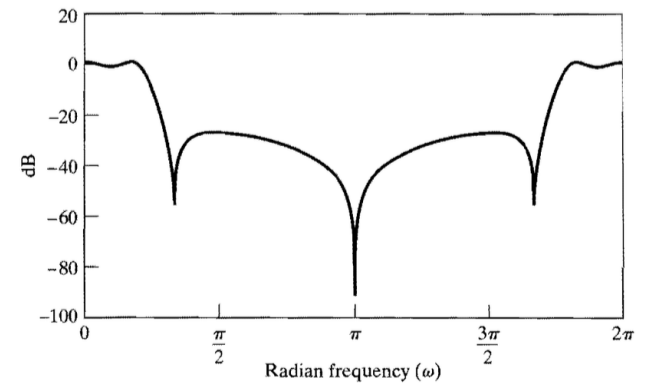
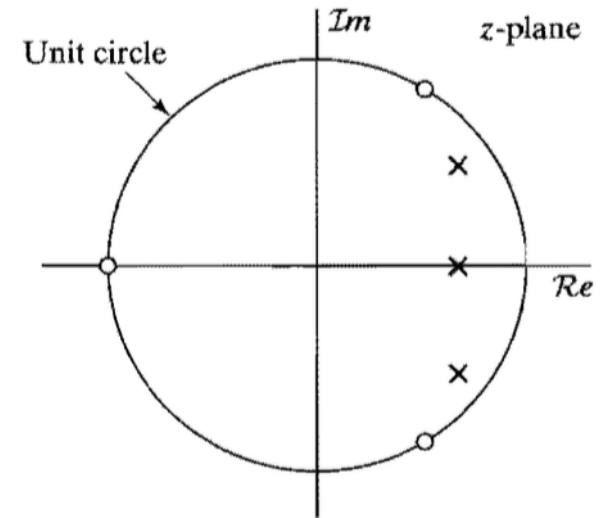
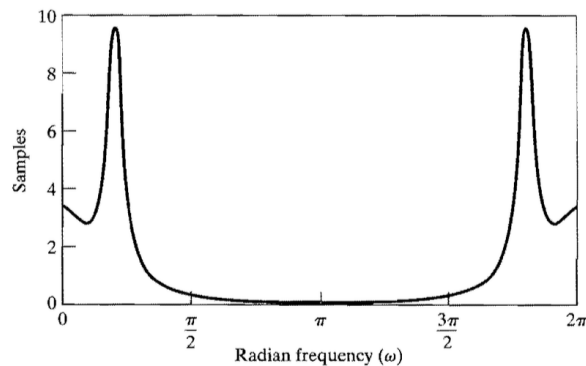
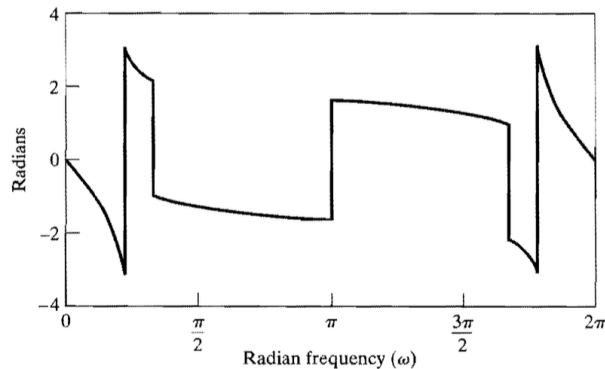
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Review: Group Delay Math

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$$\arg[1 - re^{j\theta} e^{-j\omega}] = \tan^{-1} \left(\frac{r \sin(\omega - \theta)}{1 - r \cos(\omega - \theta)} \right)$$

$$\text{grd}[1 - re^{j\theta} e^{-j\omega}] = \frac{r^2 - r \cos(\omega - \theta)}{|1 - re^{j\theta} e^{-j\omega}|^2}$$

Review: Group Delay Math

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$$H(e^{j\omega}) = \frac{b_0 \prod_{k=1}^M (1 - c_k e^{-j\omega})}{a_0 \prod_{k=1}^N (1 - d_k e^{-j\omega})}$$

arg of products is sum of args

$$\arg[H(e^{j\omega})] = \sum_{k=1}^M \arg[1 - c_k e^{-j\omega}] - \sum_{k=1}^N \arg[1 - d_k e^{-j\omega}]$$

$$\text{grd}[H(e^{j\omega})] = \sum_{k=1}^M \text{grd}[1 - c_k e^{-j\omega}] - \sum_{k=1}^N \text{grd}[1 - d_k e^{-j\omega}]$$

Stability and Causality



LTI System

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

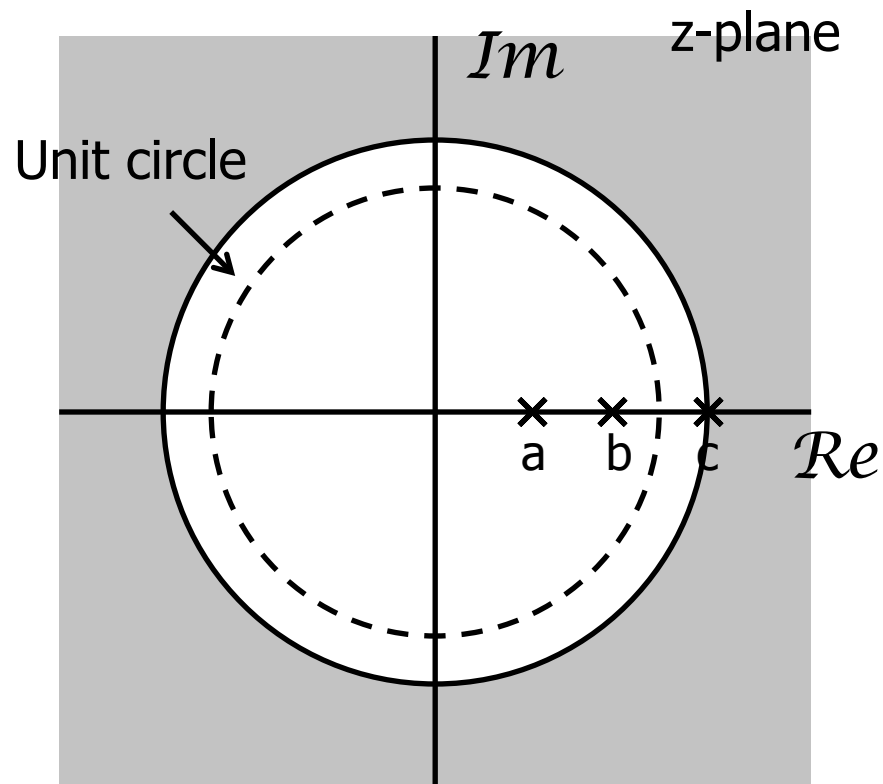
Example: $y[n] = x[n] + 0.1y[n-1]$

$$H(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_M z^{-M}}{a_0 + a_1 z^{-1} + \dots + a_N z^{-N}} = \frac{b_0 \prod_{k=1}^M (1 - c_k z^{-1})}{a_0 \prod_{k=1}^N (1 - d_k z^{-1})}$$

- ❑ Transfer function is not unique without ROC
 - If diff. eq represents LTI and causal system, ROC is region outside all singularities
 - If diff. eq represents LTI and stable system, ROC includes unit circle in z-plane

Example: ROC from Pole-Zero Plot

ROC 1: right-sided



LTI System

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

Example: $y[n] = x[n] + 0.1y[n-1]$

Stable and causal
if all poles inside
unit circle

$$H(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_M z^{-M}}{a_0 + a_1 z^{-1} + \dots + a_N z^{-N}} = \frac{b_0 \prod_{k=1}^M (1 - c_k z^{-1})}{a_0 \prod_{k=1}^N (1 - d_k z^{-1})}$$

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All-Pass Systems



All-Pass Filters

- A system is an all-pass system if

$$|H(e^{j\omega})| = 1, \text{ all } \omega$$

- Its phase response $\theta(\omega)$ may be non-trivial



First Order All-Pass Filter (a real)

$$H(z) = \frac{z^{-1} - a^*}{1 - az^{-1}}$$



First Order All-Pass Filter (a real)

$$H(z) = \frac{z^{-1} - a^*}{1 - az^{-1}}$$

$$\begin{aligned} |H(e^{j\omega})| &= \frac{|e^{-j\omega} - a^*|}{|1 - ae^{-j\omega}|} \\ &= \frac{|e^{-j\omega}(1 - a^*e^{j\omega})|}{|1 - ae^{-j\omega}|} \\ &= \frac{|1 - a^*e^{j\omega}|}{|1 - ae^{-j\omega}|} = 1 \end{aligned}$$



General All-Pass Filter

- d_k =real pole, e_k =complex poles paired w/ conjugate, e_k^*

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

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- Example:

$$d_k = -\frac{3}{4}$$
$$e_k = 0.8e^{j\pi/4}$$

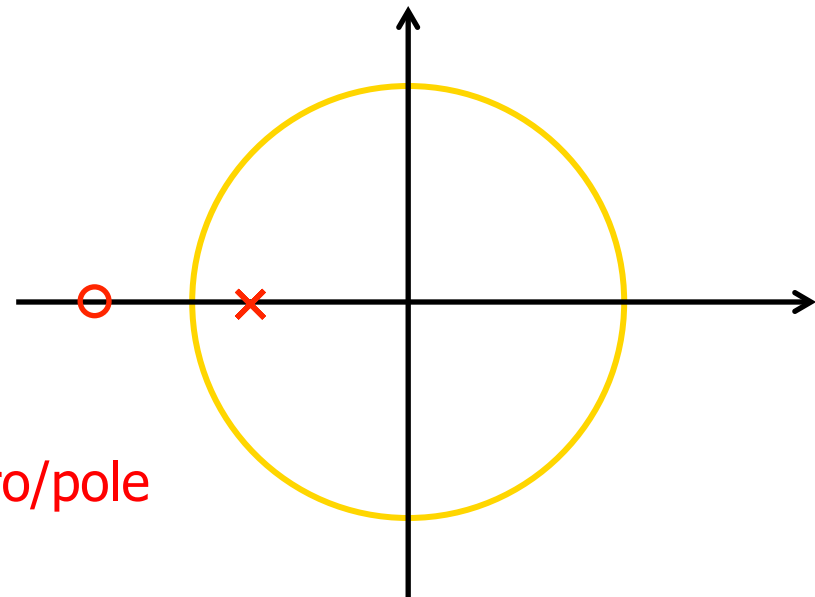
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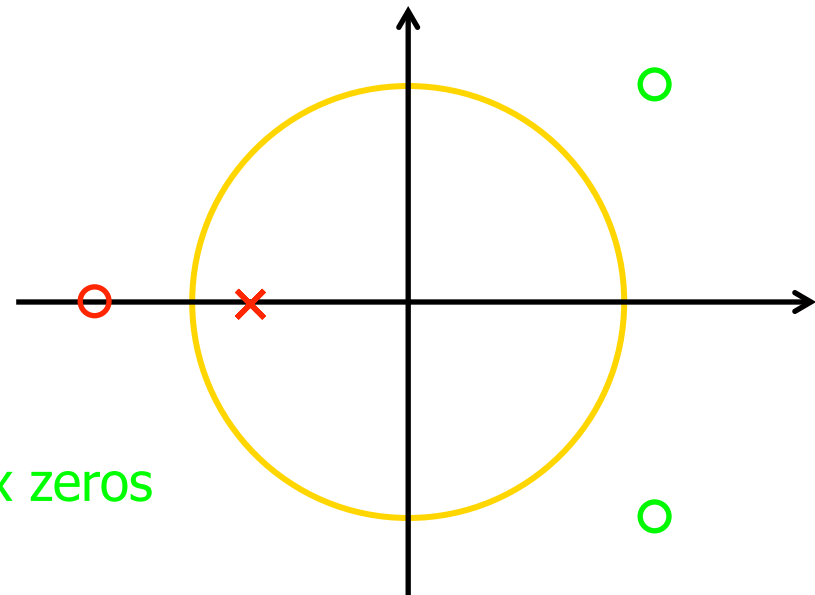
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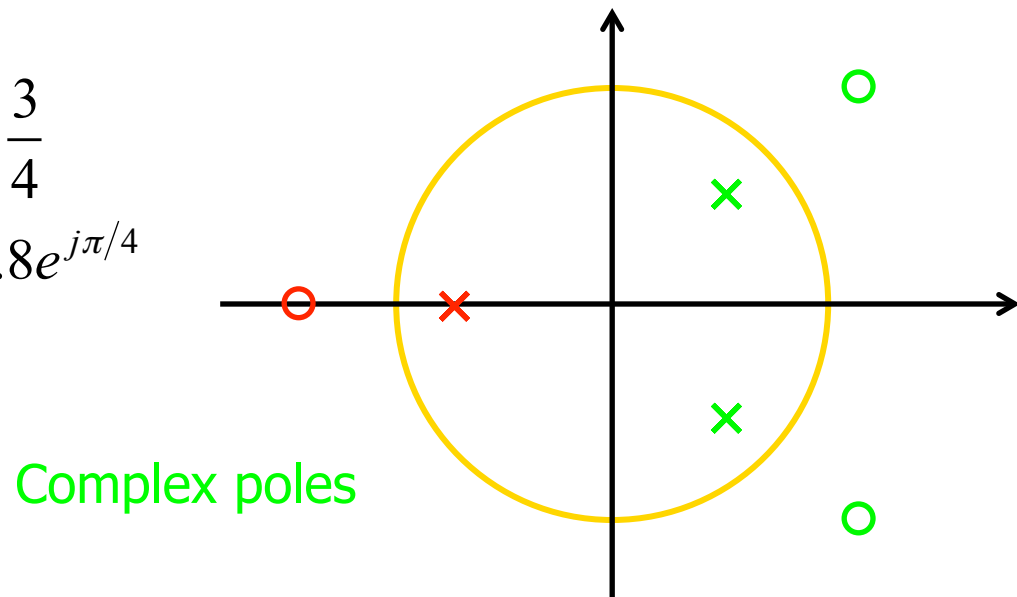
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All Pass Filter Phase Response

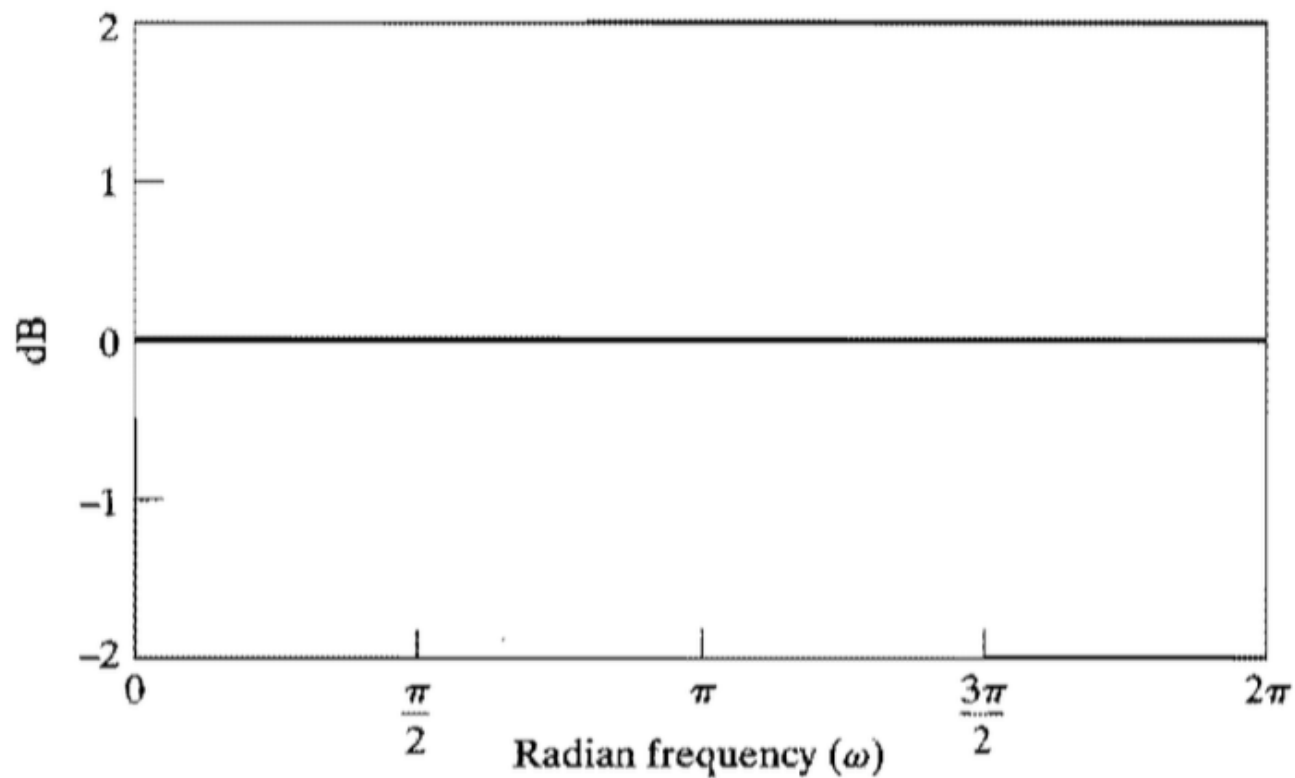
□ First order system

$$H(e^{j\omega}) = \frac{e^{-j\omega} - a^*}{1 - ae^{-j\omega}}$$
$$= \frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}$$

□ phase

First Order Example

□ Magnitude:





All Pass Filter Phase Response

□ First order system
$$H(e^{j\omega}) = \frac{e^{-j\omega} - a^*}{1 - ae^{-j\omega}}$$
$$= \frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}$$

□ phase
$$\arg\left(\frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}\right)$$

All Pass Filter Phase Response

□ First order system
$$H(e^{j\omega}) = \frac{e^{-j\omega} - a^*}{1 - ae^{-j\omega}}$$
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□ phase
$$\arg\left(\frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}\right)$$
$$= \arg\left(\frac{e^{-j\omega}(1 - re^{-j\theta}e^{j\omega})}{1 - re^{j\theta}e^{-j\omega}}\right)$$

All Pass Filter Phase Response

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$$= \arg(e^{-j\omega}) + \arg(1 - re^{-j\theta}e^{j\omega}) - \arg(1 - re^{j\theta}e^{-j\omega})$$

All Pass Filter Phase Response

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$$= -\omega - \arg(1 - re^{j\theta}e^{-j\omega}) - \arg(1 - re^{j\theta}e^{-j\omega})$$

All Pass Filter Phase Response

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□ phase
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$$= -\omega - \arg(1 - re^{j\theta}e^{-j\omega}) - \arg(1 - re^{j\theta}e^{-j\omega})$$
$$= -\omega - 2\arg(1 - re^{j\theta}e^{-j\omega})$$



Group Delay Math

$$\text{grd}[H(e^{j\omega})] = \sum_{k=1}^M \text{grd}[1 - c_k e^{-j\omega}] - \sum_{k=1}^N \text{grd}[1 - d_k e^{-j\omega}]$$

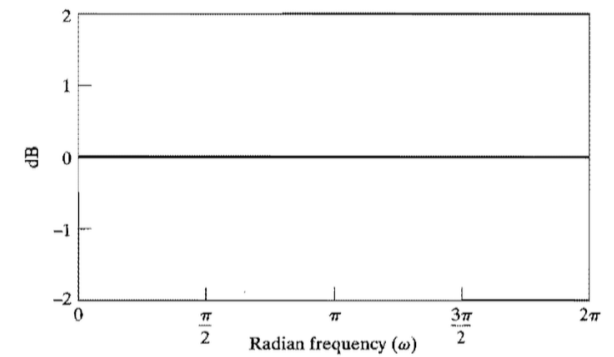
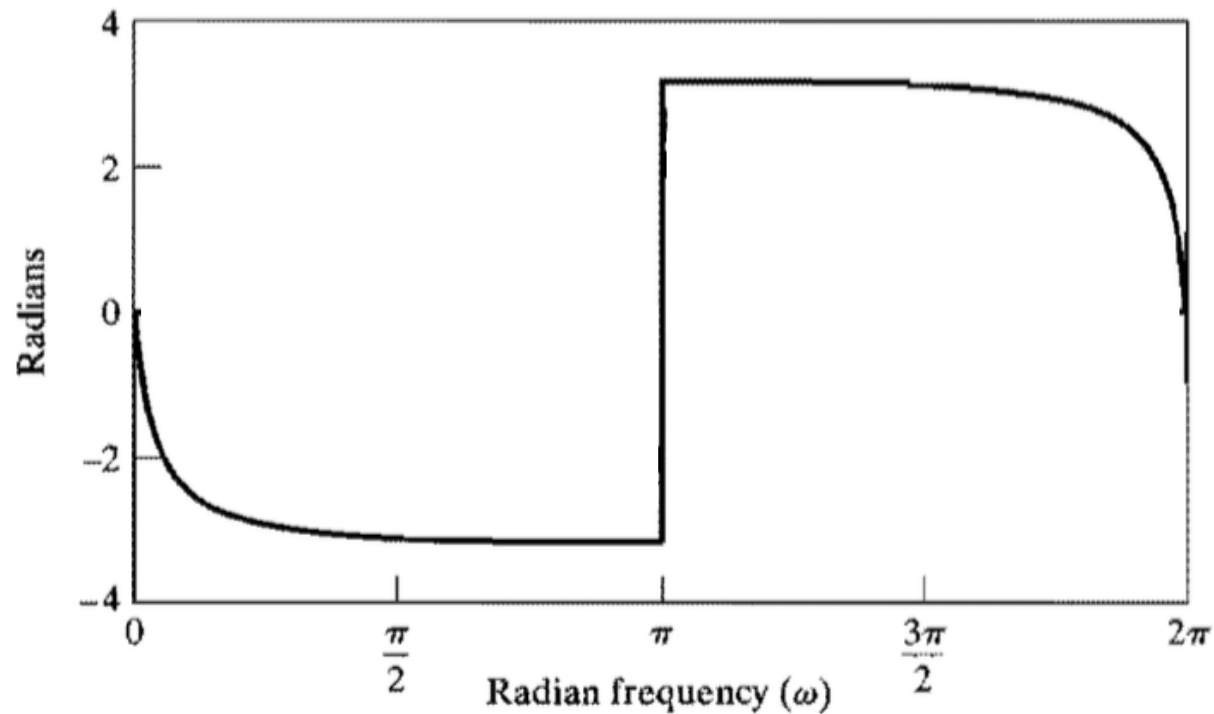
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First Order Example

□ Phase:

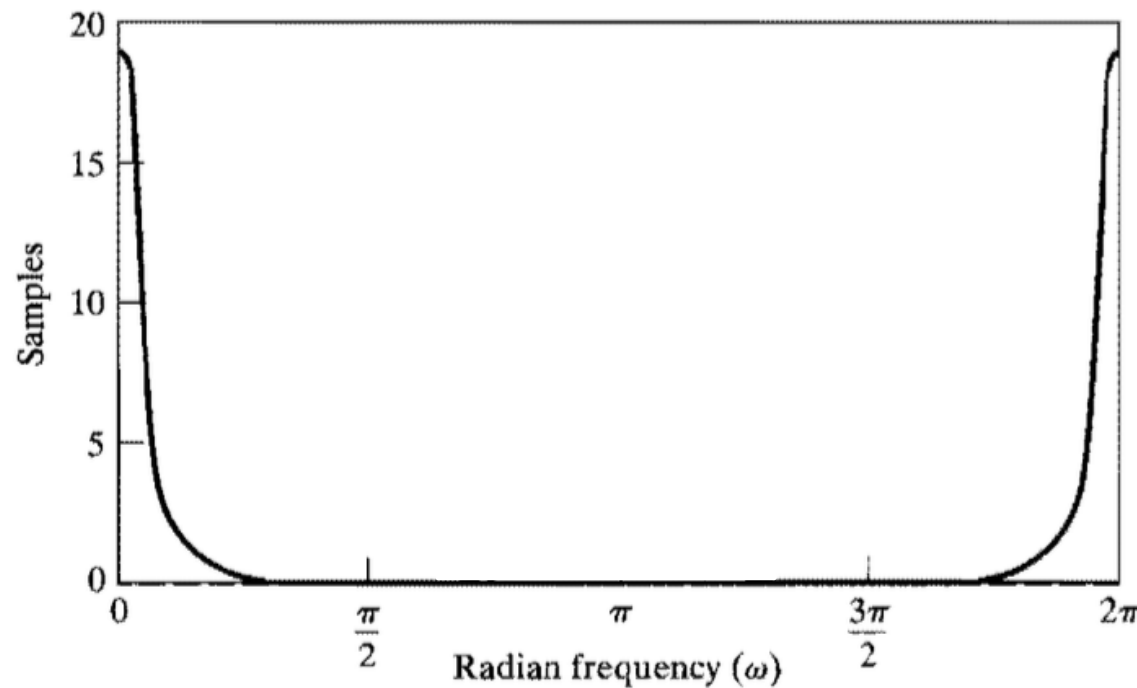
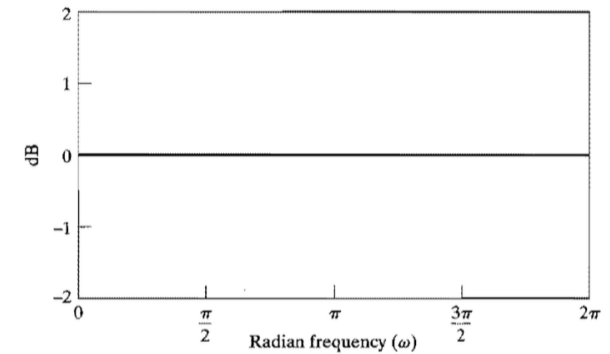
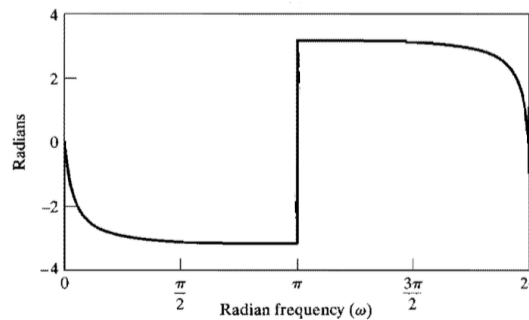


— $z = 0.9$

— $\theta = 0$

First Order Example

□ Group Delay:



— $z = 0.9$
— $\theta = 0$

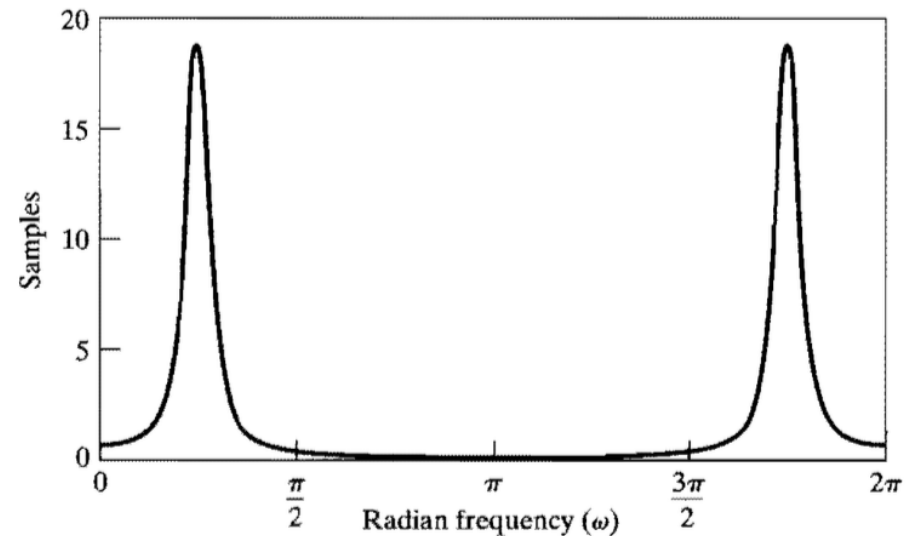
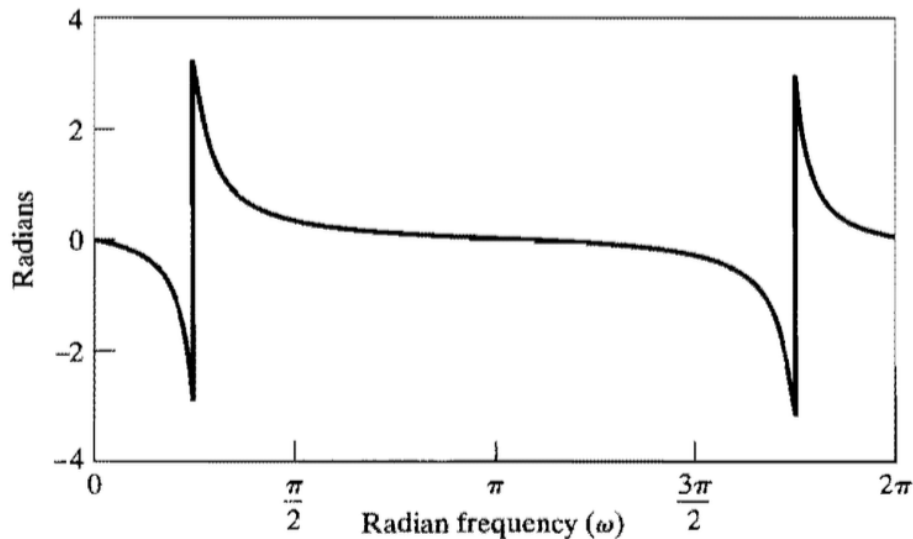
All Pass Filter Phase Response

- Second order system with poles at $z = re^{j\theta}, re^{-j\theta}$

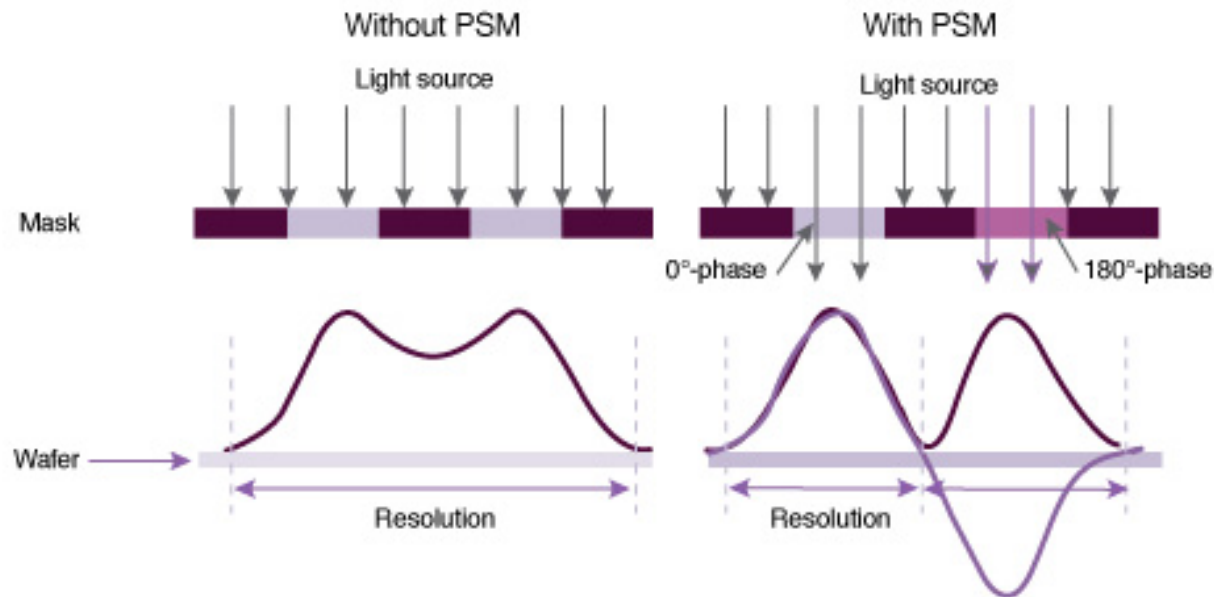
$$\angle \left[\frac{(e^{-j\omega} - re^{-j\theta})(e^{-j\omega} - re^{j\theta})}{(1 - re^{j\theta}e^{-j\omega})(1 - re^{-j\theta}e^{-j\omega})} \right] = -2\omega - 2 \arctan \left[\frac{r \sin(\omega - \theta)}{1 - r \cos(\omega - \theta)} \right] \\ - 2 \arctan \left[\frac{r \sin(\omega + \theta)}{1 - r \cos(\omega + \theta)} \right].$$

Second Order Example

□ Poles at $z = 0.9e^{\pm j\pi/4}$ (zeros at conjugates)



Phase Shift Masking

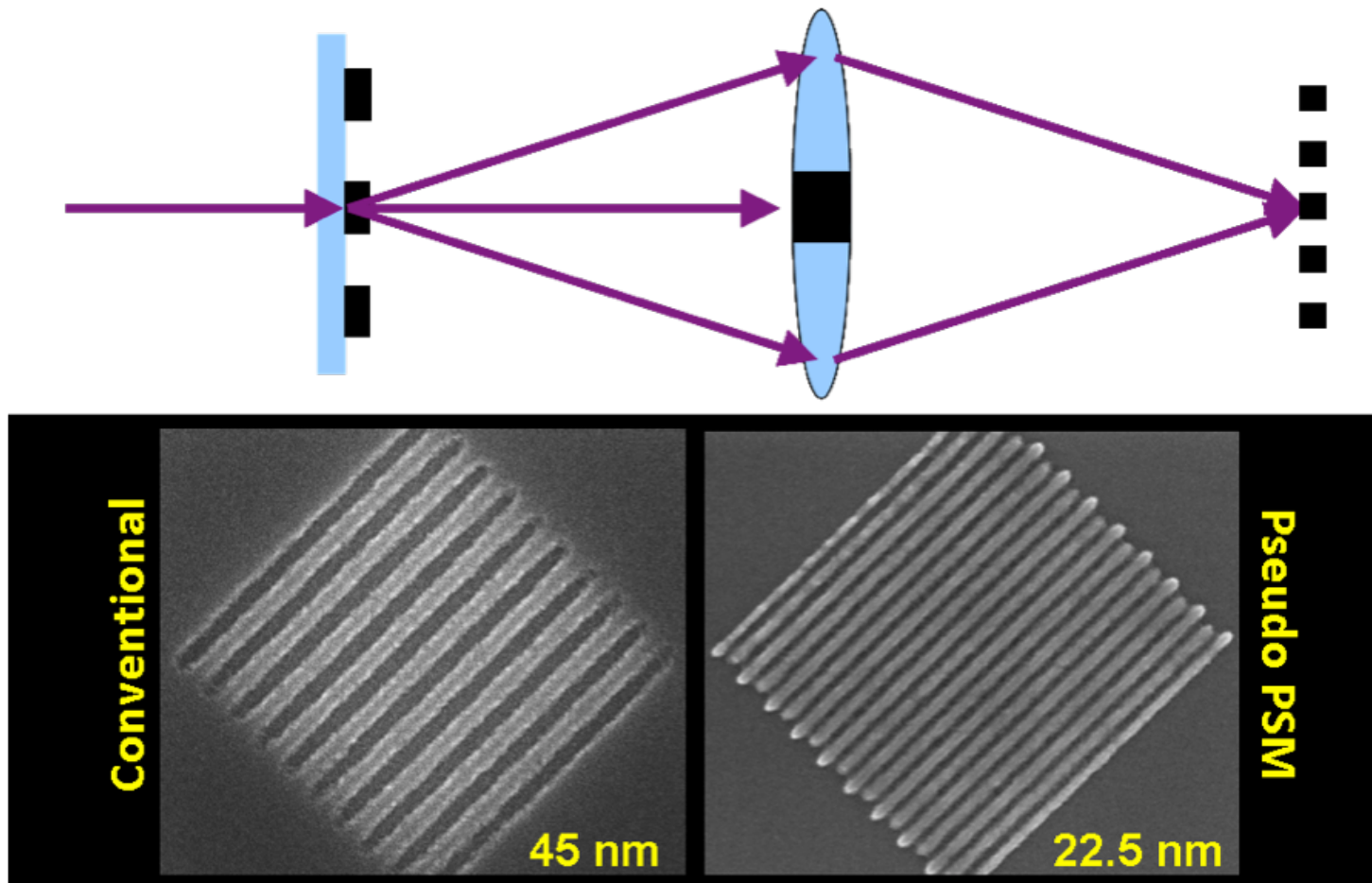


Today's chips use $\lambda = 193\text{nm}$

Source

<http://www.synopsys.com/Tools/Manufacturing/MaskSynthesis/PSMCreate/Pages/default.aspx>

Phase Shift Masking



Patrick Naulleau, et al., "The SEMATECH Berkeley MET pushing EUV development beyond 22nm half pitch", Proc. SPIE 7636, Extreme Ultraviolet (EUV) Lithography, 76361J (March 22, 2010)



All-Pass Properties

□ Claim: For a stable, causal ($r < 1$) all-pass system:

- $\arg[H_{ap}(e^{j\omega})] \leq 0$

- Unwrapped phase always non-positive and decreasing

- $\text{grd}[H_{ap}(e^{j\omega})] > 0$

- Group delay always positive

Pg 309 in text book

- Intuition

- delay is positive $\rightarrow$ system is causal

- Phase negative $\rightarrow$ phase lag

Minimum-Phase Systems



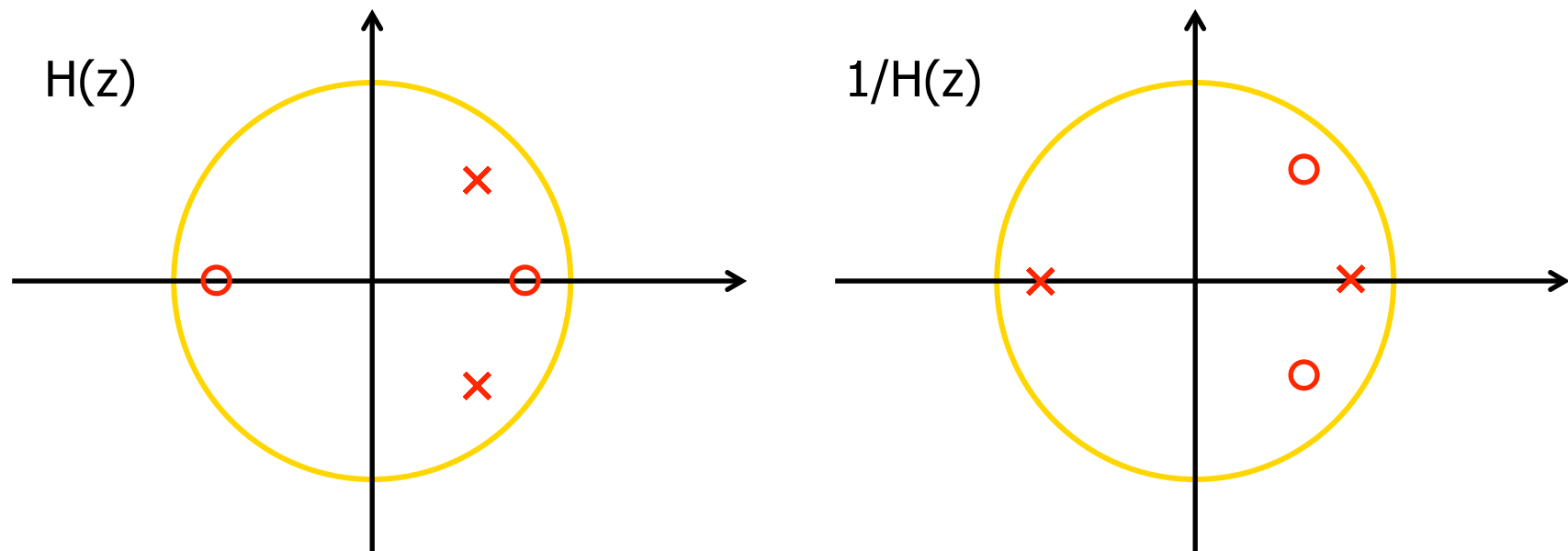


Minimum-Phase Systems

- Definition: A stable and causal system $H(z)$ (i.e. poles inside unit circle) whose inverse $1/H(z)$ is also stable and causal (i.e. zeros inside unit circle)
 - All poles and zeros inside unit circle

Minimum-Phase Systems

- Definition: A stable and causal system $H(z)$ (i.e. poles inside unit circle) whose inverse $1/H(z)$ is also stable and causal (i.e. zeros inside unit circle)
 - All poles and zeros inside unit circle





All-Pass Min-Phase Decomposition

- Any stable, causal system can be decomposed to:

$$H(z) = H_{\min}(z) \cdot H_{ap}(z)$$

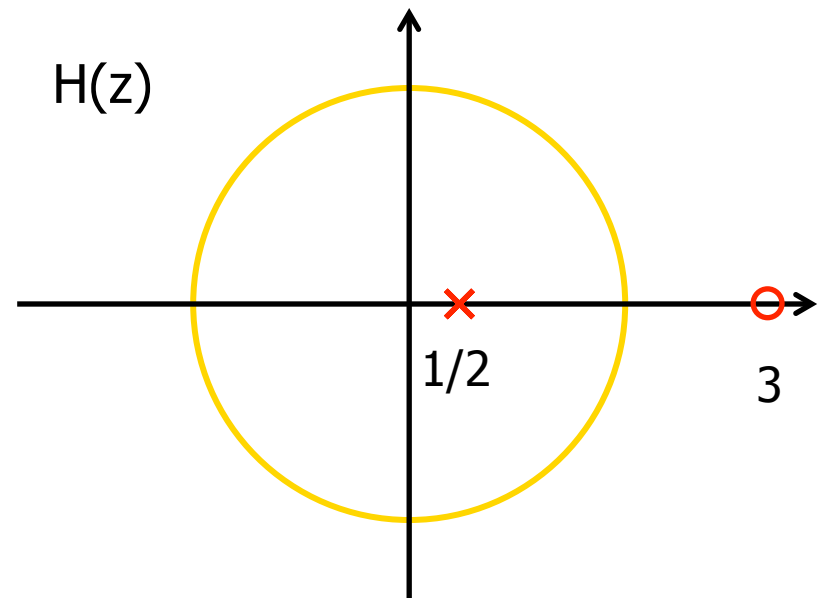
- Approach:

- (1) First construct H_{ap} with all zeros outside unit circle
- (2) Compute

$$H_{\min}(z) = \frac{H(z)}{H_{ap}(z)}$$

Min-Phase Decomposition Example

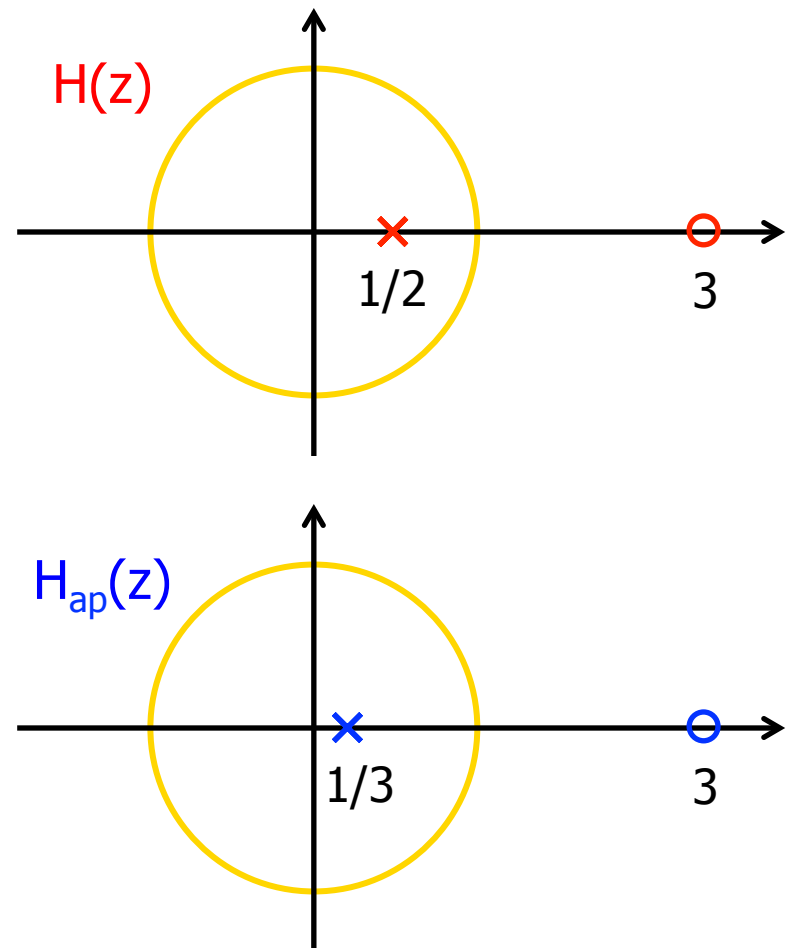
$$H(z) = \frac{1 - 3z^{-1}}{1 - \frac{1}{2}z^{-1}}$$



Min-Phase Decomposition Example

$$H(z) = \frac{1 - 3z^{-1}}{1 - \frac{1}{2}z^{-1}}$$

□ Set $H_{ap}(z) = \frac{z^{-1} - \frac{1}{3}}{1 - \frac{1}{3}z^{-1}}$

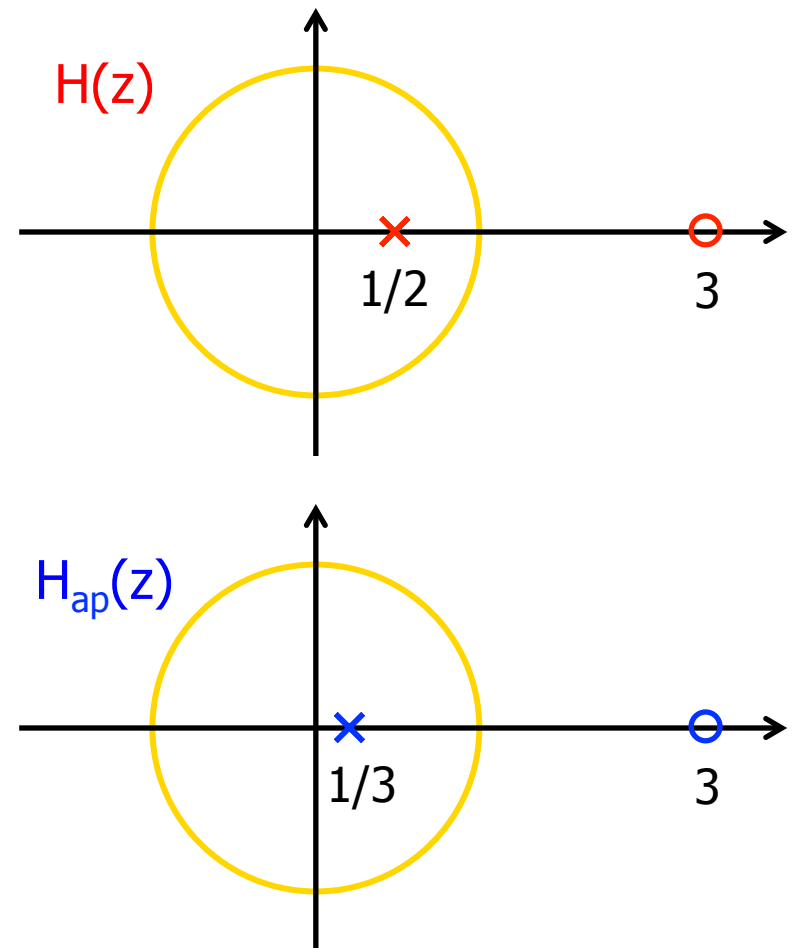


Min-Phase Decomposition Example

$$H(z) = \frac{1 - 3z^{-1}}{1 - \frac{1}{2}z^{-1}}$$

□ Set $H_{ap}(z) = \frac{z^{-1} - \frac{1}{3}}{1 - \frac{1}{3}z^{-1}}$

$$H_{\min}(z) = \frac{1 - 3z^{-1}}{1 - \frac{1}{2}z^{-1}} \left(\frac{z^{-1} - \frac{1}{3}}{1 - \frac{1}{3}z^{-1}} \right)^{-1}$$

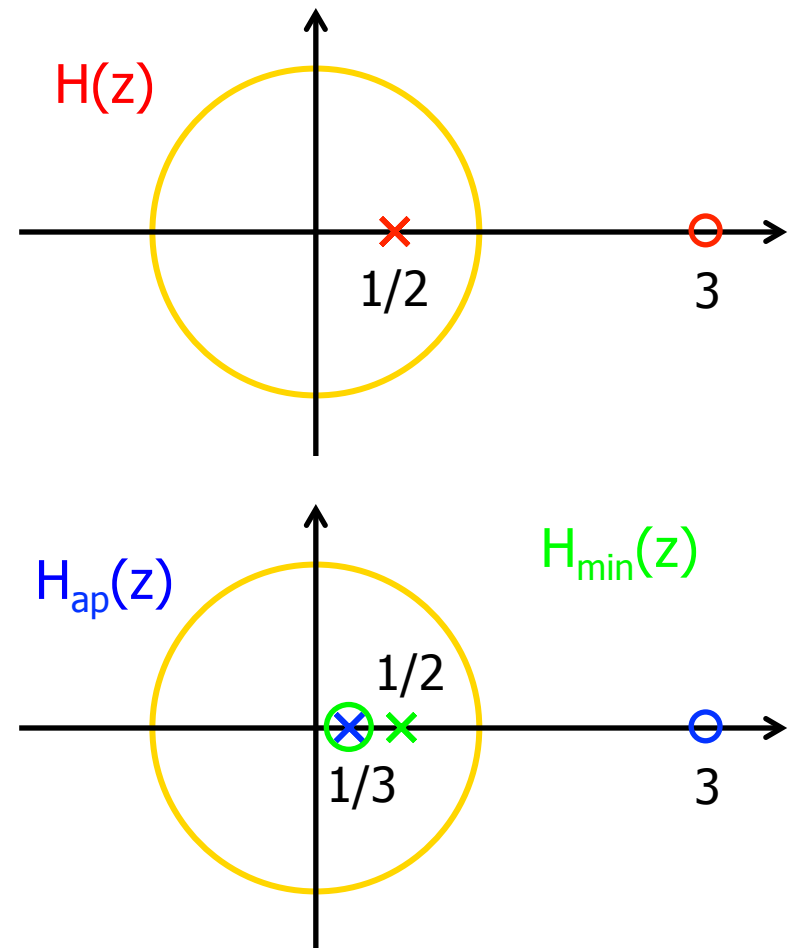


Min-Phase Decomposition Example

$$H(z) = \frac{1 - 3z^{-1}}{1 - \frac{1}{2}z^{-1}}$$

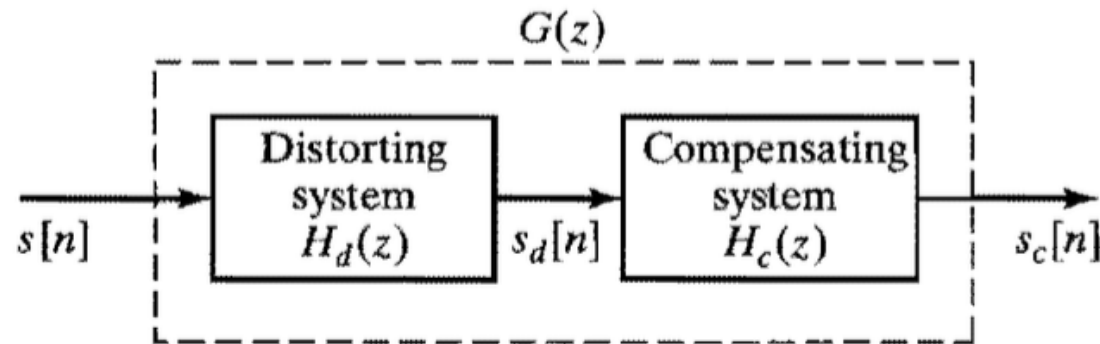
□ Set $H_{ap}(z) = \frac{z^{-1} - \frac{1}{3}}{1 - \frac{1}{3}z^{-1}}$

$$H_{\min}(z) = -3 \frac{1 - \frac{1}{3}z^{-1}}{1 - \frac{1}{2}z^{-1}}$$



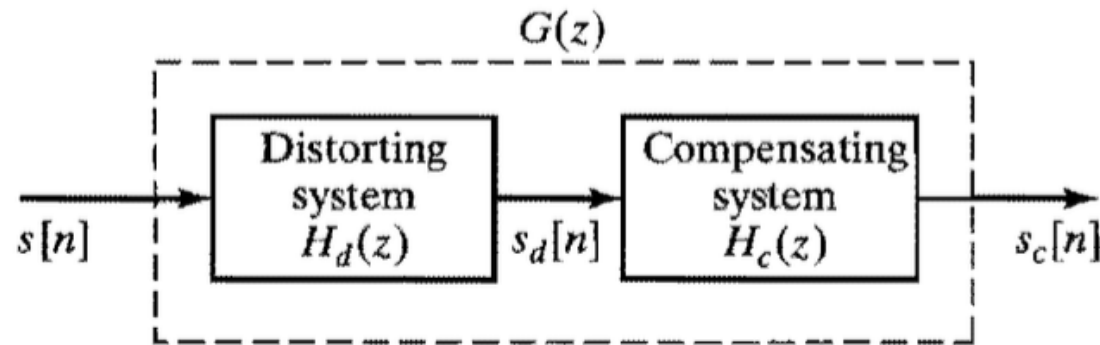
Min-Phase Decomposition Purpose

- Have some distortion that we want to compensate for:



Min-Phase Decomposition Purpose

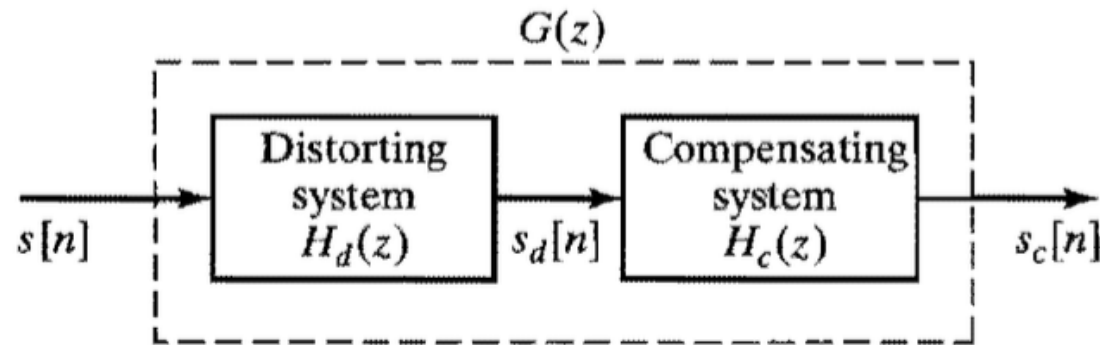
- Have some distortion that we want to compensate for:



- If $H_d(z)$ is min phase, easy:
 - $H_c(z) = 1/H_d(z)$ ← also stable and causal

Min-Phase Decomposition Purpose

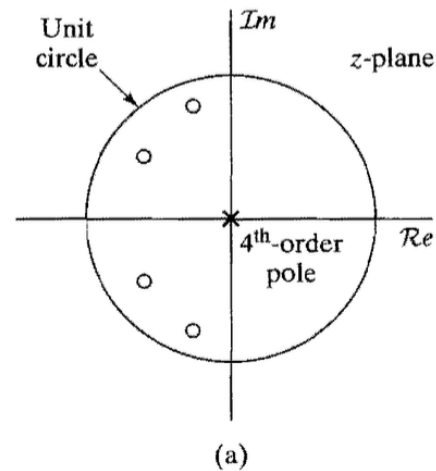
- Have some distortion that we want to compensate for:



- If $H_d(z)$ is min phase, easy:
 - $H_c(z) = 1/H_d(z)$ ← also stable and causal
- Else, decompose $H_d(z) = H_{d,\min}(z) H_{d,\text{ap}}(z)$
 - $H_c(z) = 1/H_{d,\min}(z) \rightarrow H_d(z)H_c(z) = H_{d,\text{ap}}(z)$
 - Compensate for magnitude distortion

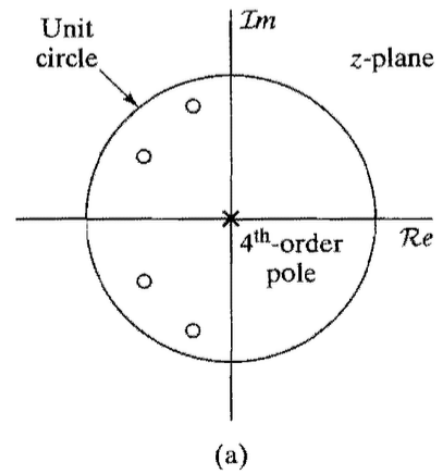
Minimum Energy-Delay Property

Min phase

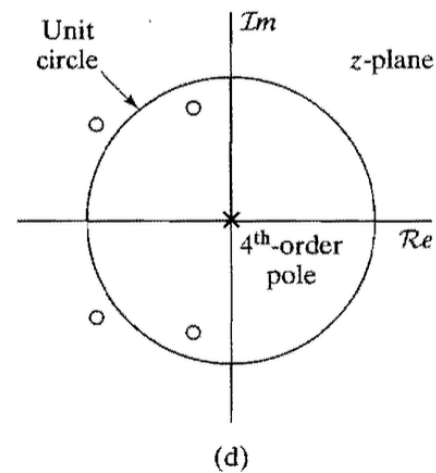
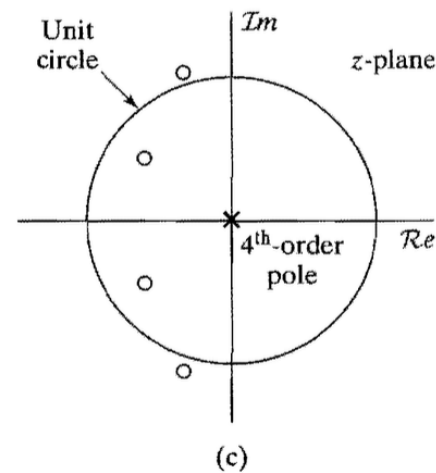
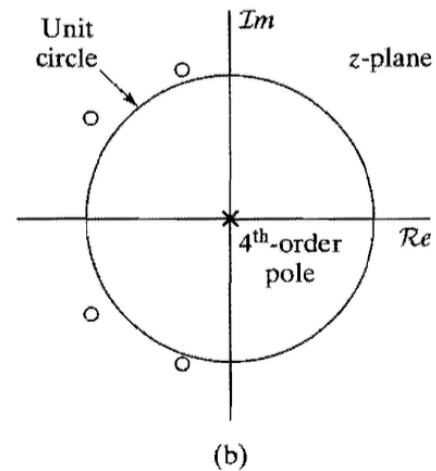


Minimum Energy-Delay Property

Min phase

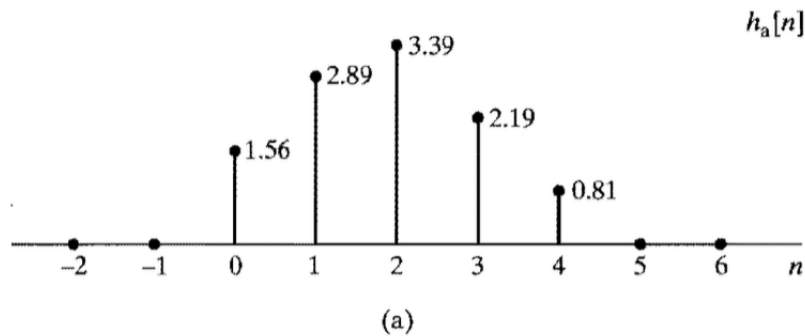


Max phase

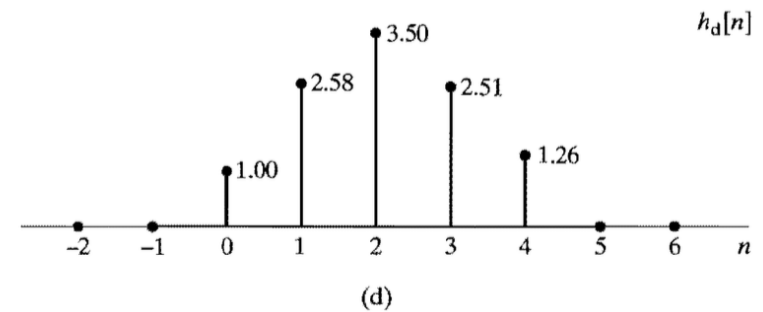
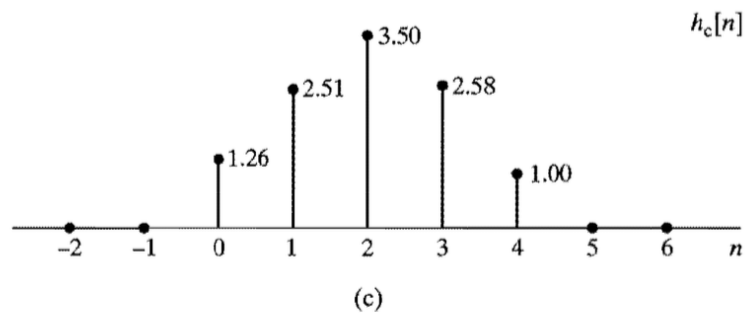
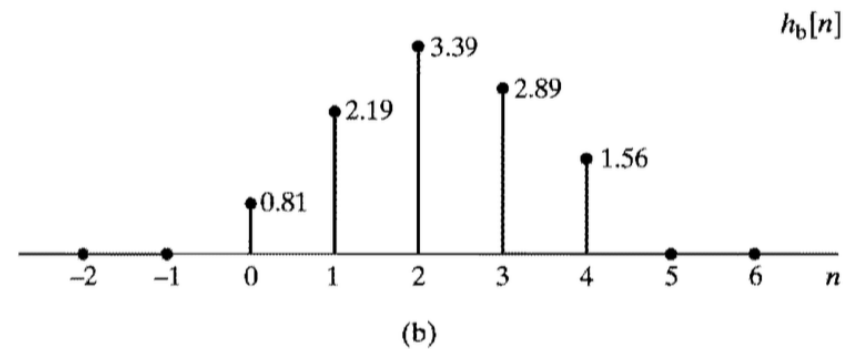


Minimum Energy-Delay Property

Min phase



Max phase





Big Ideas

- ❑ Frequency Response of LTI Systems
 - Magnitude Response, Phase Response, Group Delay
- ❑ LTI Stability and Causality
 - If all poles inside unit circle
- ❑ All Pass Systems
 - Used for delay compensation
- ❑ Minimum Phase Systems
 - Can compensate for magnitude distortion
 - Minimum energy-delay property



Midterm Exam

□ Midterm – 3/13

- During class
 - Starts at exactly 4:30pm, ends at exactly 5:50pm (80 minutes)
- Location DRLB A8
- Old exam posted on previous year's website
- Covers Lec 1- 13
- Closed book, one page cheat sheet allowed
- Calculators allowed, no smart phones
- Review Session TBD (likely 3/11)
- Tania office hours moved to Monday (3/12)



Admin

- HW 5

- Due Friday 3/2