ESE 531: Digital Signal Processing

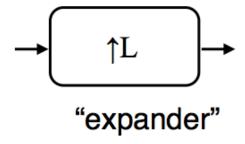
Lec 11: February 21, 2019
Polyphase Decomposition and Multi-rate
Filter Banks



Lecture Outline

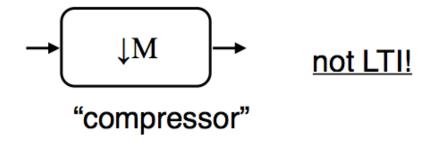
- □ Review: Interchanging Operations
- Polyphase Decomposition
- Multi-Rate Filter Banks

Expander and Compressor



Upsampling

- -expanding in time
- -compressing in frequency



Downsampling

- -compressing in time
- -expanding in frequency

Interchanging Operations - Summary

Filter and expander

Expander and expanded filter*

$$x[n] \rightarrow \underbrace{H(z)} \rightarrow \underbrace{\uparrow L} \rightarrow y[n] \quad \equiv \quad x[n] \rightarrow \underbrace{\uparrow L} \rightarrow \underbrace{H(z^L)} \rightarrow y[n]$$

$$x[n] \longrightarrow \underbrace{\downarrow \mathbf{M}} \longrightarrow \underbrace{H(z)} \longrightarrow y[n] \qquad \equiv \qquad x[n] \longrightarrow \underbrace{H(z^M)} \longrightarrow \underbrace{\downarrow \mathbf{M}} \longrightarrow y[n]$$

Compressor and filter

Expanded filter* and compressor

^{*}Expanded filter = expanded impulse response, compressed freq response

- □ The polyphase decomposition of a sequence is obtained by representing it as a superposition of M subsequences, each consisting of every Mth value of successively delayed versions of the sequence.
- When this decomposition is applied to a filter impulse response, it can lead to efficient implementation structures for linear filters in several contexts.

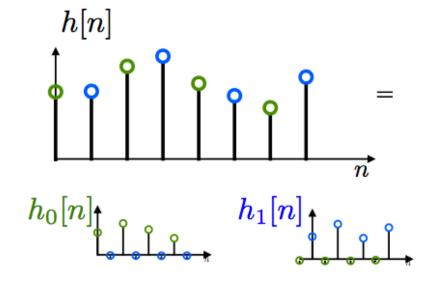
■ We can decompose an impulse response (of our filter) to:

$$h[n] = \sum_{k=0}^{M-1} h_k[n-k]$$

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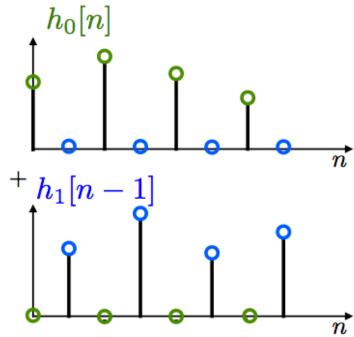
$$h[n] = \sum_{k=0}^{M-1} h_k[n-k]$$

M=2



■ We can decompose an impulse response (of our filter) to:

$$h[n] = \sum_{k=0}^{M-1} h_k[n-k]$$



$$h_k[n] \longrightarrow \downarrow M \longrightarrow e_k[n]$$

$$e_k[n] = h_k[nM]$$

$$h_0[n]$$

$$h_1[n]$$

$$h_1[n]$$

$$h_{k}[n] \longrightarrow \downarrow M \longrightarrow e_{k}[n]$$

$$e_{k}[n] = h_{k}[nM]$$

$$h_{0}[n] \longrightarrow e_{0}[n]$$

$$h_{1}[n] \longrightarrow e_{1}[n]$$

$$e_{1}[n] \longrightarrow e_{1}[n]$$



$$e_k[n] \longrightarrow f_M \longrightarrow h_k[n]$$

recall upsampling ⇒ scaling

$$H_k(z) = E_k(z^M)$$

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recall upsampling ⇒ scaling

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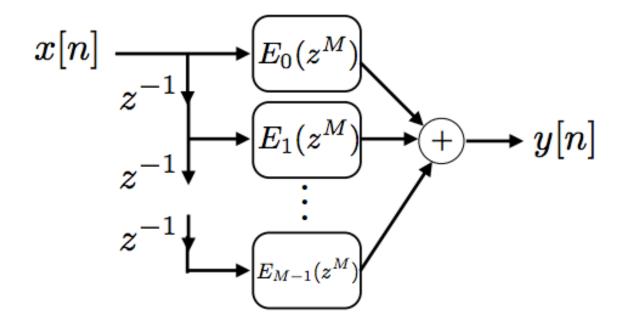
Also, recall:

$$h[n] = \sum_{k=0}^{M-1} h_k[n-k]$$

So,

$$H(z) = \sum_{k=0}^{M-1} E_k(z^M) z^{-k}$$

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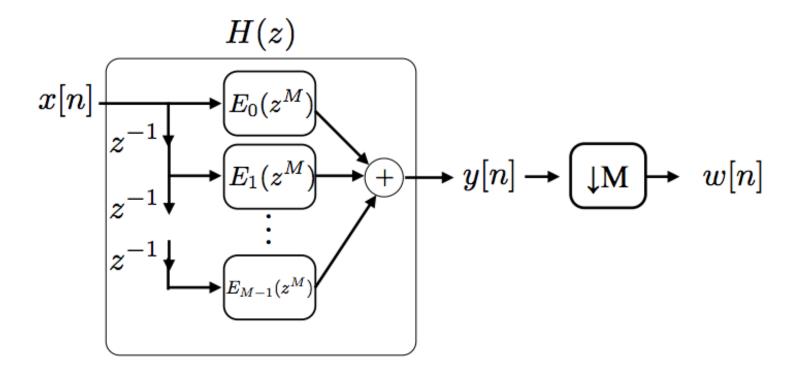
□ Problem:

Compute all y[n] and then throw away -- wasted computation!

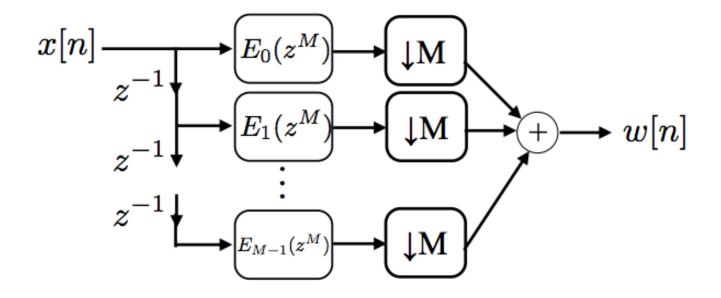
□ Problem:

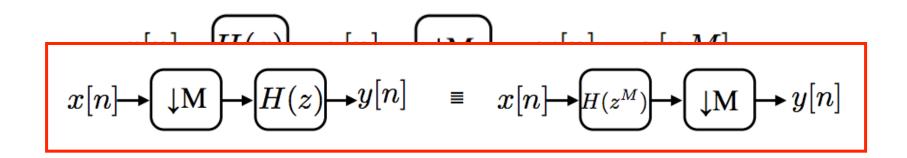
- Compute all y[n] and then throw away -- wasted computation!
- For FIR length $N \rightarrow N$ multiplications/unit time

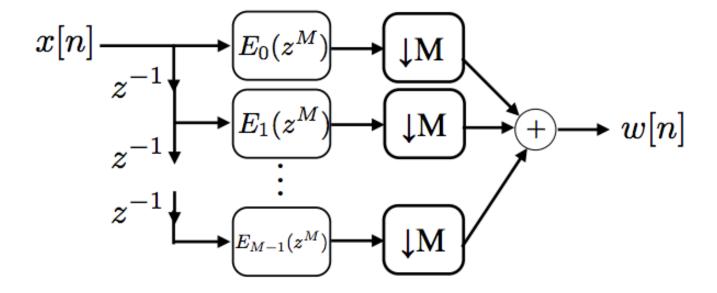
$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$



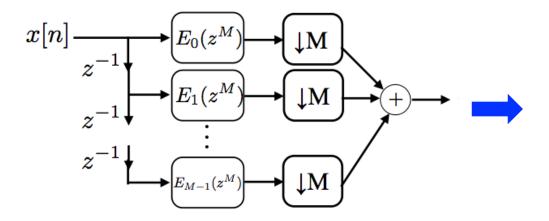
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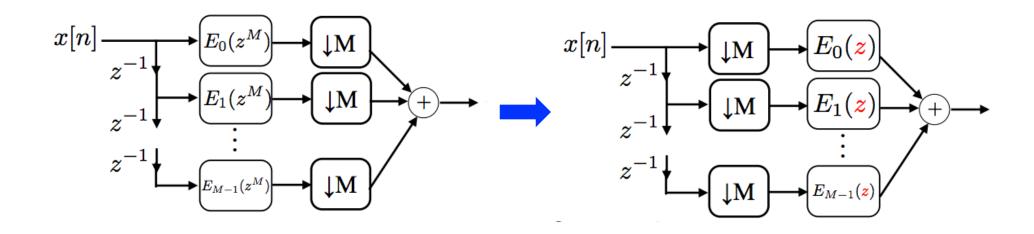




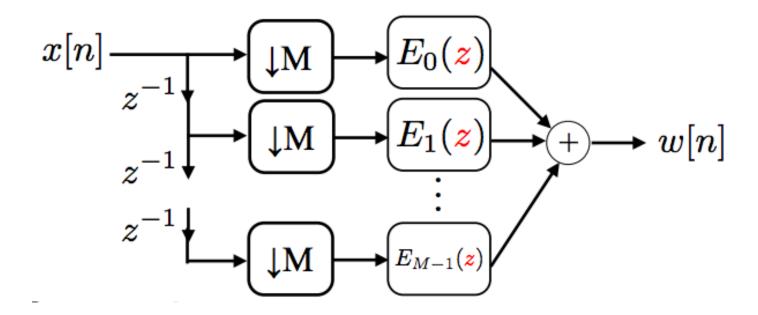
$$x[n] \longrightarrow \underbrace{\downarrow M} \longrightarrow \underbrace{H(z)} \longrightarrow y[n] \equiv x[n] \longrightarrow \underbrace{H(z^M)} \longrightarrow \underbrace{\downarrow M} \longrightarrow y[n]$$



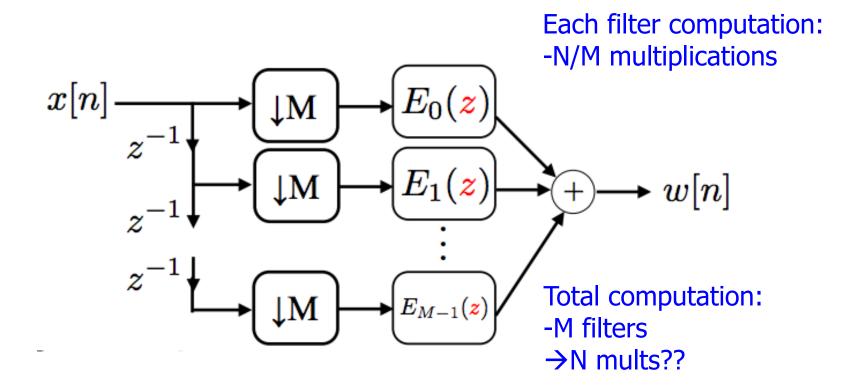
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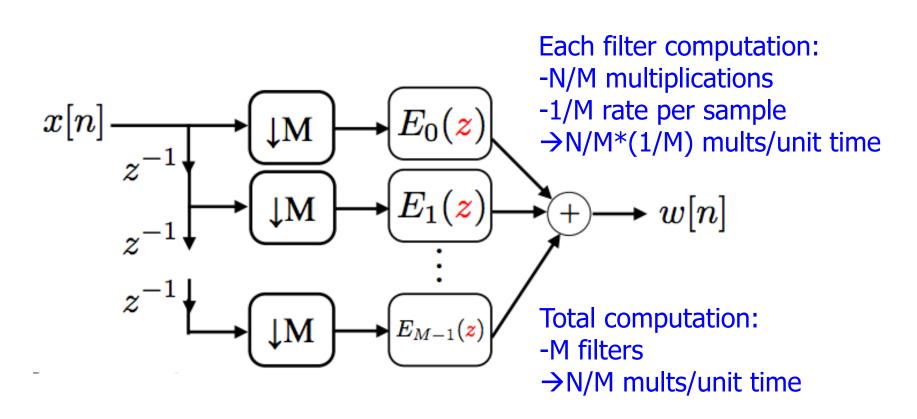


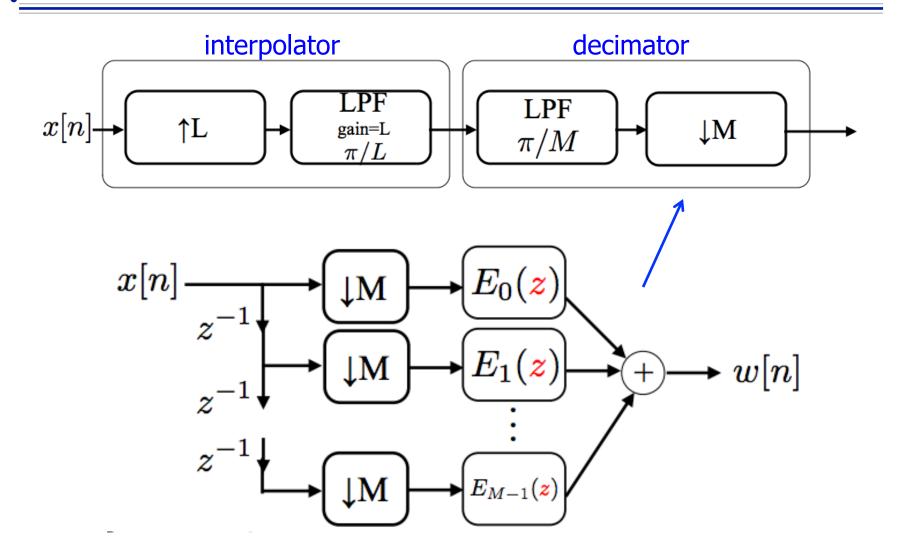
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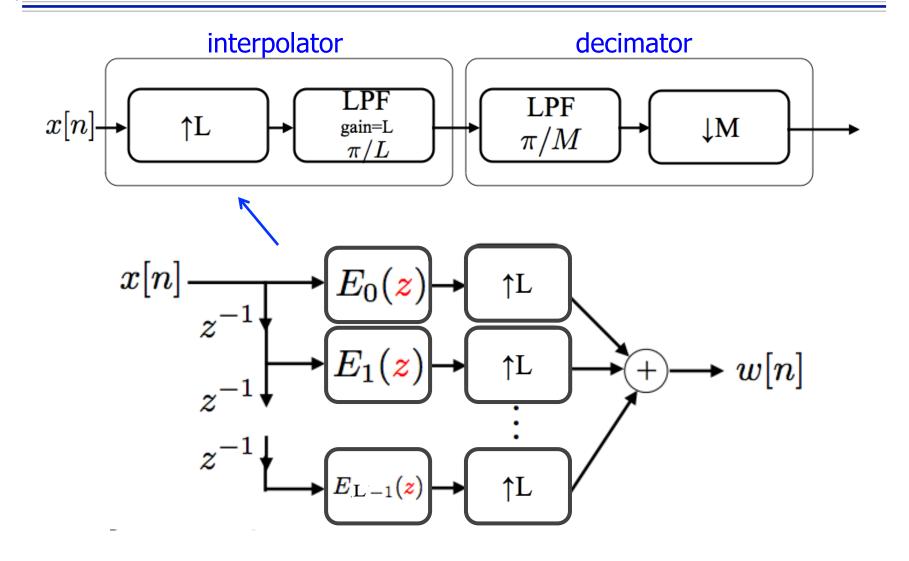
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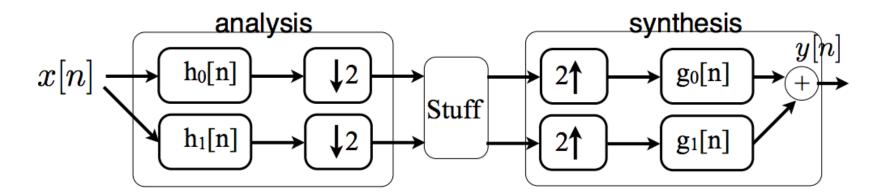
Polyphase Implementation of Interpolation



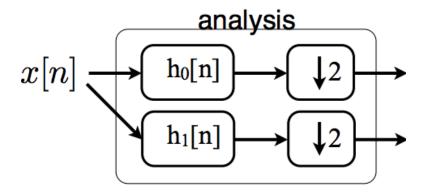
- Use filter banks to operate on a signal differently in different frequency bands
 - To save computation, reduce the rate after filtering

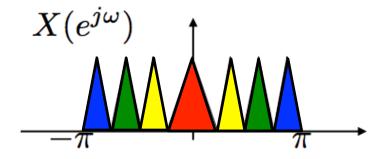


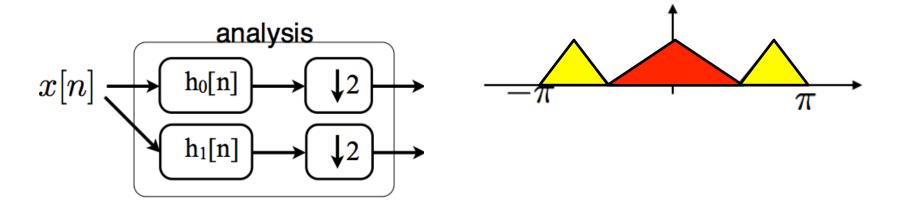
- Use filter banks to operate on a signal differently in different frequency bands
 - To save computation, reduce the rate after filtering
- \bullet h₀[n] is low-pass, h₁[n] is high-pass
 - Often $h_1[n] = e^{j\pi n} h_0[n]$ \leftarrow shift freq resp by π

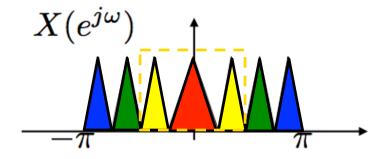


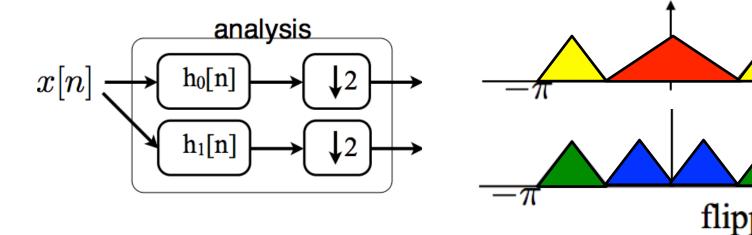


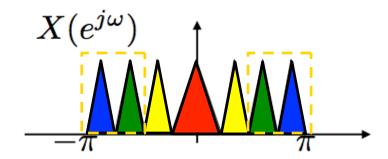


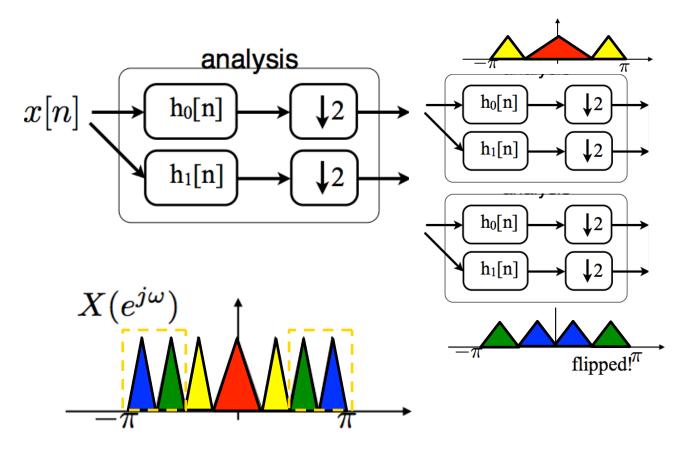


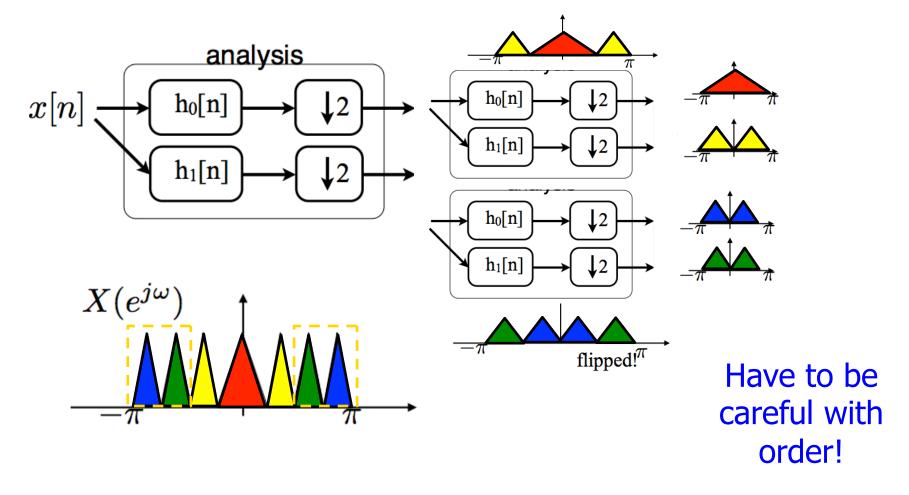


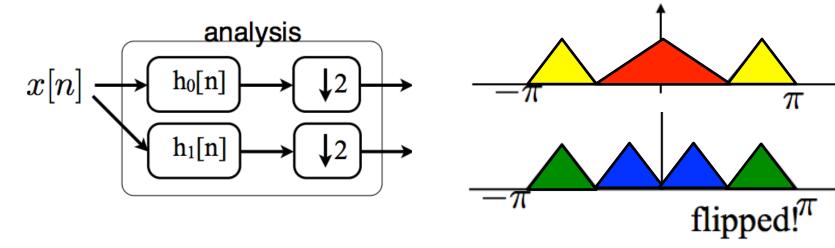


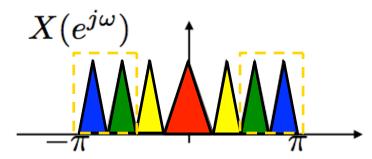


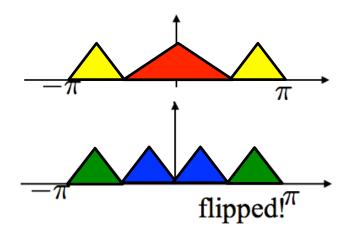


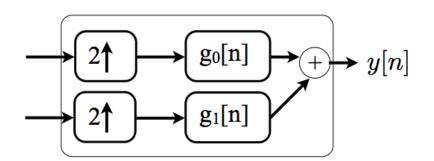


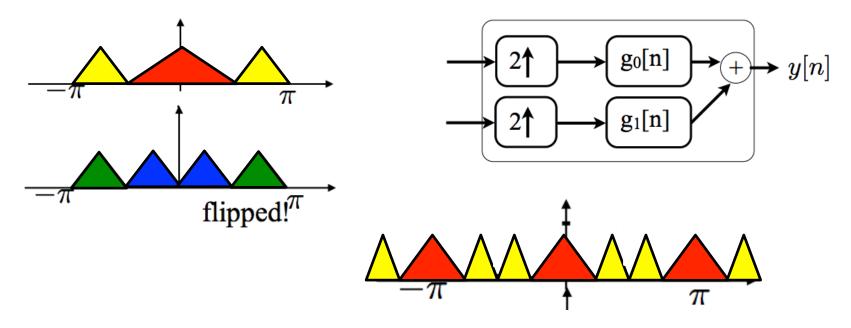


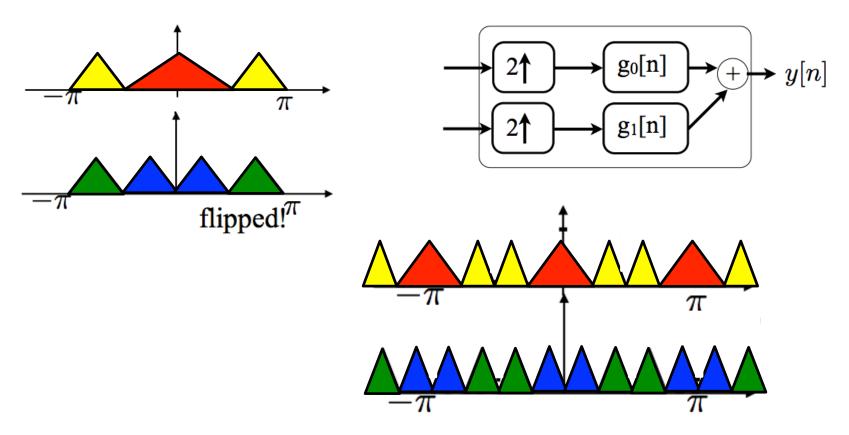






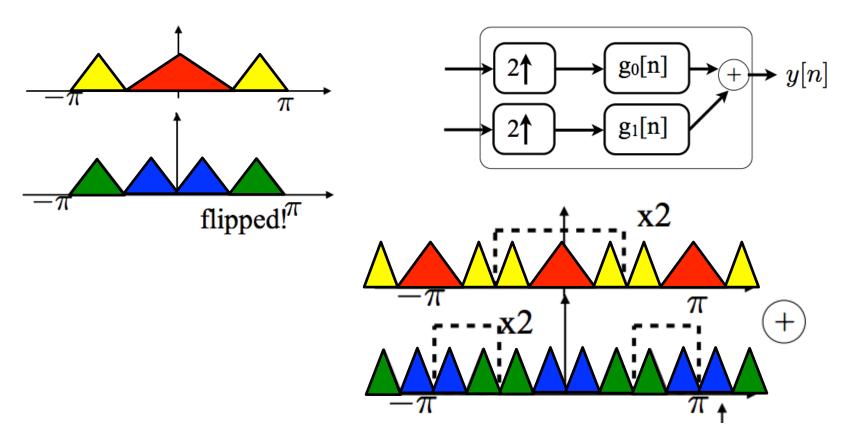






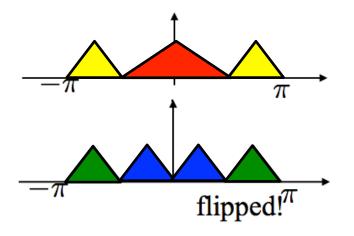
Multi-Rate Filter Banks

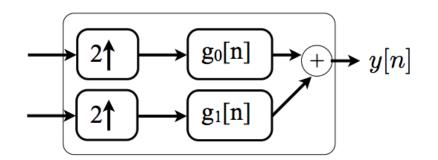
□ Assume h_0 , h_1 are ideal low/high pass with $\omega_C = \pi/2$

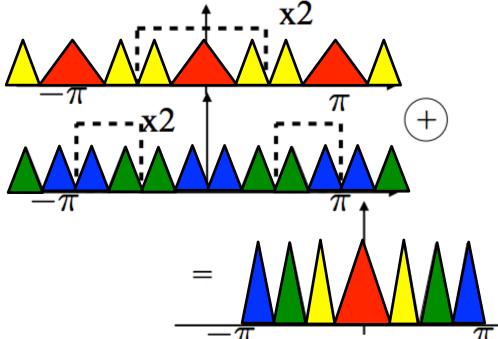


Multi-Rate Filter Banks

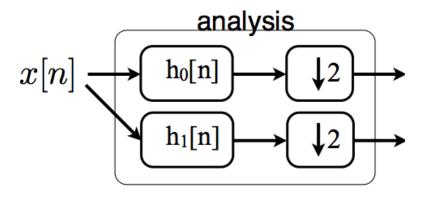
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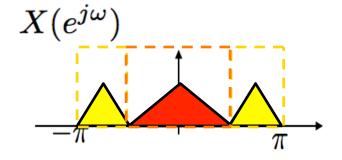


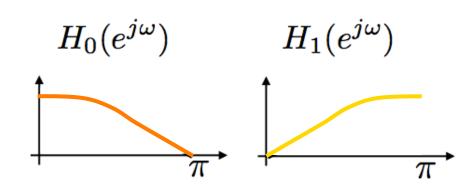


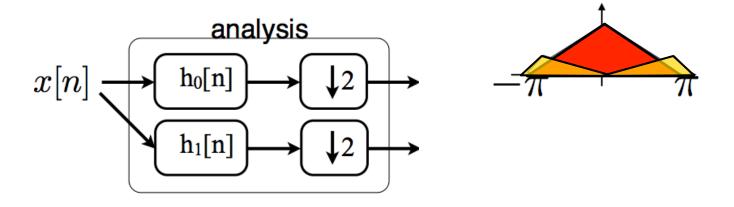


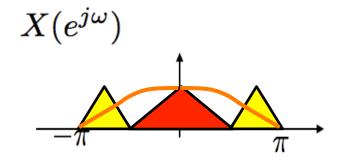
Multi-Rate Filter Banks

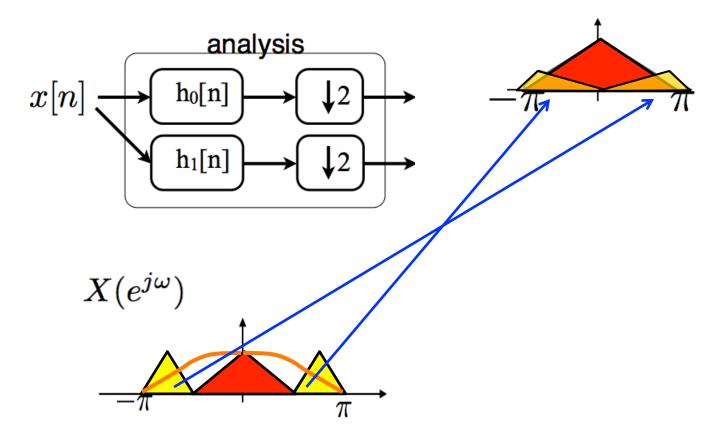


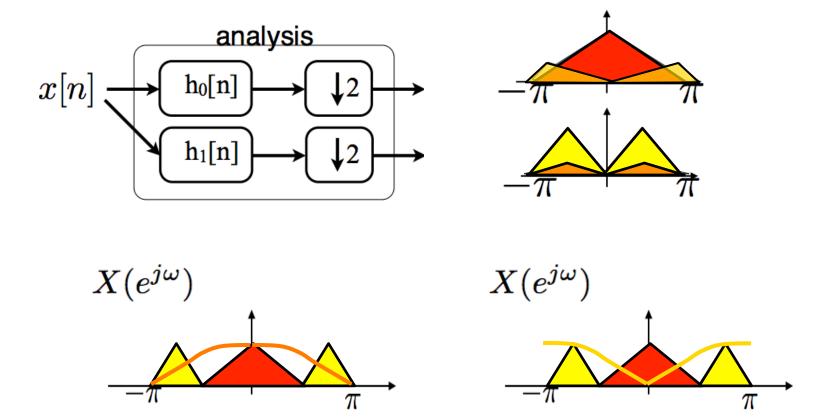


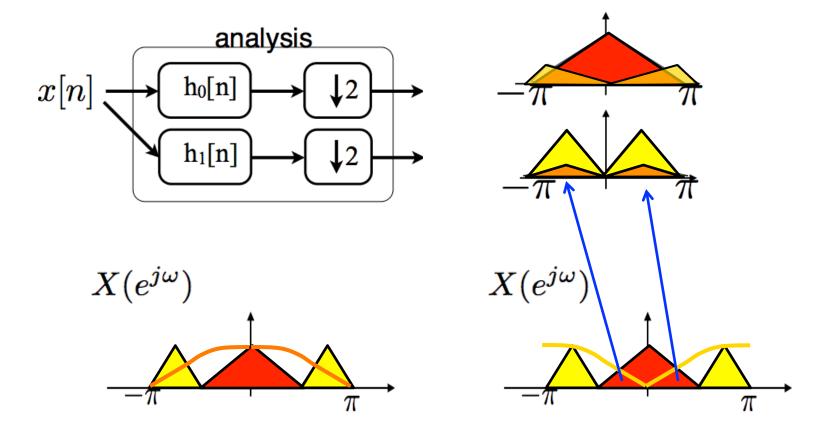


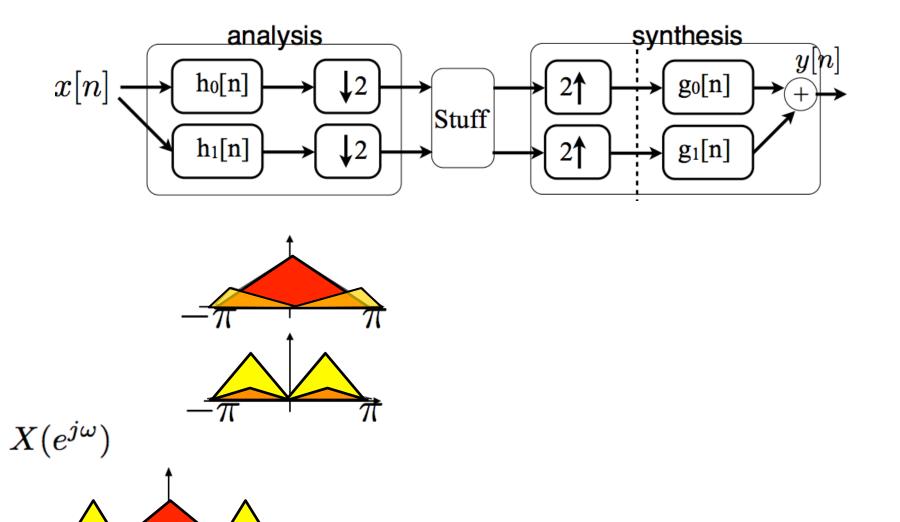




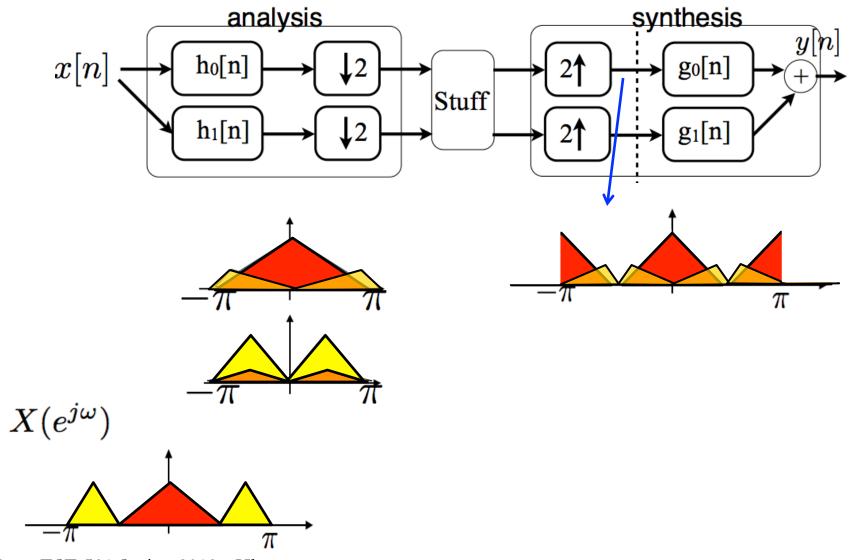




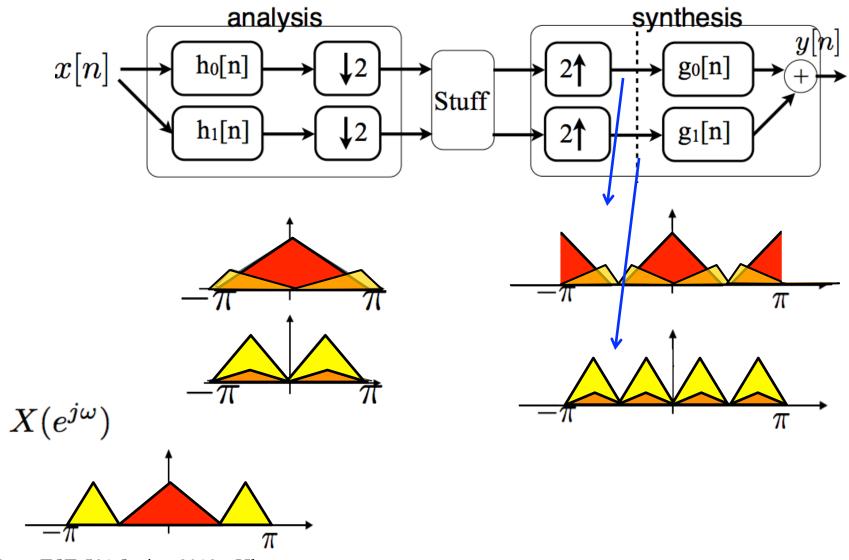




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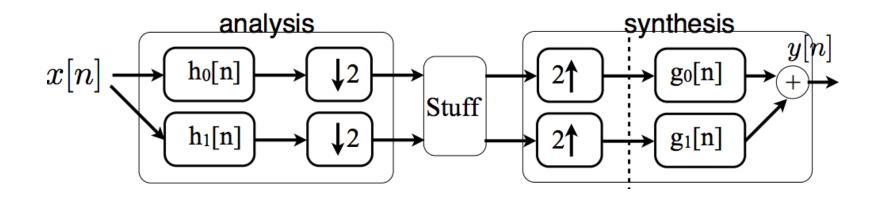


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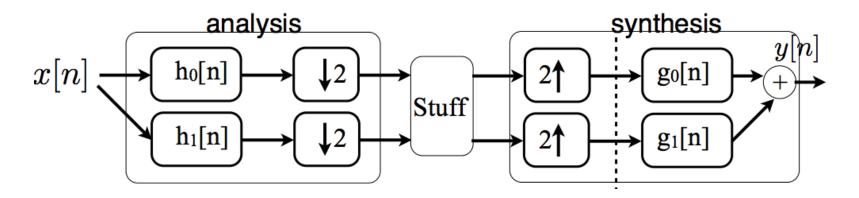
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Perfect Reconstruction non-Ideal Filters



$$Y(e^{j\omega}) = \frac{1}{2} \left[G_0(e^{j\omega}) H_0(e^{j\omega}) + G_1(e^{j\omega}) H_1(e^{j\omega}) \right] X(e^{j\omega}) \\ + \frac{1}{2} \left[G_0(e^{j\omega}) H_0(e^{j(\omega-\pi)}) + G_1(e^{j\omega}) H_1(e^{j(\omega-\pi)}) \right] X(e^{j(\omega-\pi)}) \\ \uparrow \qquad \qquad \downarrow \qquad \qquad \uparrow \qquad \qquad \downarrow \qquad \qquad$$

Quadrature Mirror Filters



Quadrature mirror filters

$$H_1(e^{j\omega}) = H_0(e^{j(\omega-\pi)})$$

 $G_0(e^{j\omega}) = 2H_0(e^{j\omega})$
 $G_1(e^{j\omega}) = -2H_1(e^{j\omega})$



Perfect Reconstruction non-Ideal Filters

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$$G_0(e^{j\omega}) = 2H_0(e^{j\omega})$$

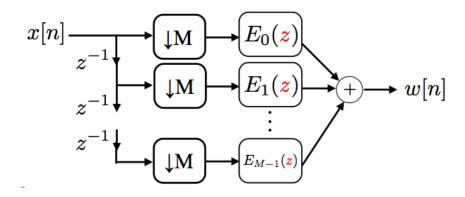
$$G_1(e^{j\omega}) = -2H_1(e^{j\omega})$$

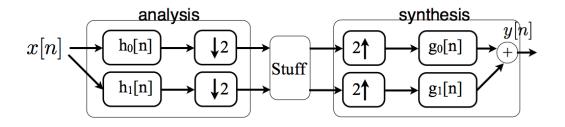
Big Ideas

- Interchanging Operations
- Polyphase Decomposition
- Multi-Rate Filter Banks

$$x[n] \longrightarrow \overbrace{\hspace{1cm}} \hspace{1cm} \downarrow \hspace{1cm} \uparrow \hspace{1cm} L \hspace{1cm} \longrightarrow \hspace{1cm} \downarrow \hspace{$$







Admin

■ HW 4 due Sunday