

## ESE 531: Digital Signal Processing

Lec 11: February 21, 2019  
Polyphase Decomposition and Multi-rate  
Filter Banks



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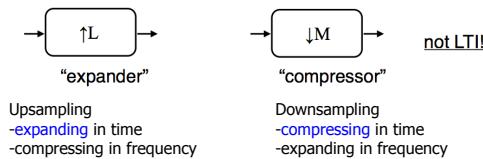
## Lecture Outline

- ❑ Review: Interchanging Operations
- ❑ Polyphase Decomposition
- ❑ Multi-Rate Filter Banks

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## Expander and Compressor



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## Interchanging Operations - Summary

Filter and expander	Expander and expanded filter*
$x[n] \rightarrow H(z) \rightarrow \uparrow L \rightarrow y[n]$	$\equiv x[n] \rightarrow \uparrow L \rightarrow H(z^L) \rightarrow y[n]$
Compressor and filter	Expanded filter* and compressor
$x[n] \rightarrow \downarrow M \rightarrow H(z) \rightarrow y[n]$	$\equiv x[n] \rightarrow H(z^M) \rightarrow \downarrow M \rightarrow y[n]$

\*Expanded filter = expanded impulse response, compressed freq response

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## Polyphase Decomposition

- ❑ The polyphase decomposition of a sequence is obtained by representing it as a superposition of  $M$  subsequences, each consisting of every  $M$ th value of successively delayed versions of the sequence.
- ❑ When this decomposition is applied to a filter impulse response, it can lead to efficient implementation structures for linear filters in several contexts.

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## Polyphase Decomposition

- ❑ We can decompose an impulse response (of our filter) to:

$$h[n] = \sum_{k=0}^{M-1} h_k[n-k]$$

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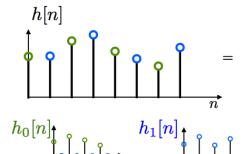
6

## Polyphase Decomposition

- We can decompose an impulse response (of our filter) to:

$$h[n] = \sum_{k=0}^{M-1} h_k[n - k]$$

M=2



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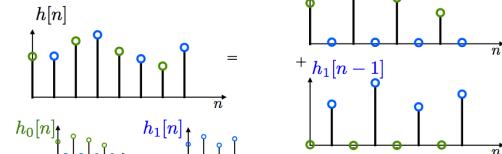
7

## Polyphase Decomposition

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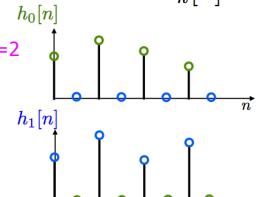
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## Polyphase Decomposition

$$h_k[n] \rightarrow \downarrow M \rightarrow e_k[n]$$

$$e_k[n] = h_k[nM]$$

M=2



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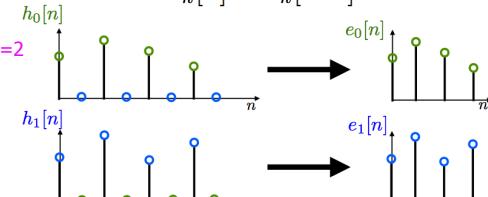
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## Polyphase Decomposition

$$h_k[n] \rightarrow \downarrow M \rightarrow e_k[n]$$

$$e_k[n] = h_k[nM]$$

M=2



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## Polyphase Decomposition

$$e_k[n] \rightarrow \uparrow M \rightarrow h_k[n]$$

recall upsampling  $\Rightarrow$  scaling

$$H_k(z) = E_k(z^M)$$

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## Polyphase Decomposition

$$e_k[n] \rightarrow \uparrow M \rightarrow h_k[n]$$

recall upsampling  $\Rightarrow$  scaling

$$H_k(z) = E_k(z^M)$$

Also, recall:

$$h[n] = \sum_{k=0}^{M-1} h_k[n - k]$$

So,

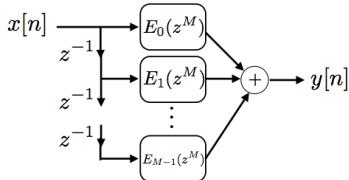
$$H(z) = \sum_{k=0}^{M-1} E_k(z^M) z^{-k}$$

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## Polyphase Decomposition

$$H(z) = \sum_{k=0}^{M-1} E_k(z^M) z^{-k}$$



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## Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \boxed{\downarrow M} \rightarrow w[n] = y[nM]$$

□ Problem:

- Compute all  $y[n]$  and then throw away -- wasted computation!

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## Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \boxed{\downarrow M} \rightarrow w[n] = y[nM]$$

□ Problem:

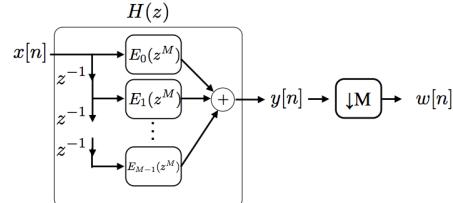
- Compute all  $y[n]$  and then throw away -- wasted computation!
- For FIR length  $N \rightarrow N$  multiplications/unit time

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## Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \boxed{\downarrow M} \rightarrow w[n] = y[nM]$$

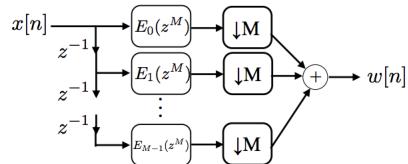


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## Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \boxed{\downarrow M} \rightarrow w[n] = y[nM]$$

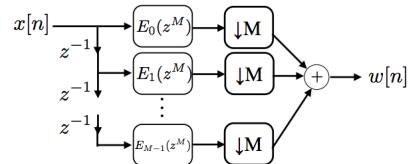


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## Polyphase Implementation of Decimation

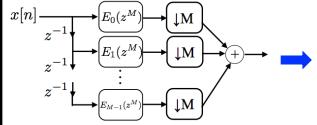
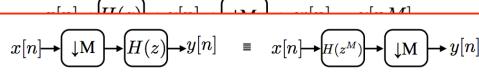
$$x[n] \rightarrow \boxed{\downarrow M} \rightarrow H(z) \rightarrow y[n] \quad \equiv \quad x[n] \rightarrow H(z^M) \rightarrow \boxed{\downarrow M} \rightarrow y[n]$$



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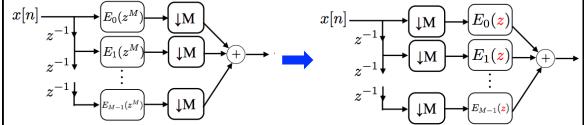
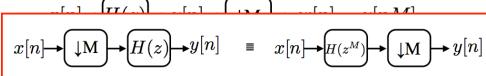
### Polyphase Implementation of Decimation



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### Polyphase Implementation of Decimation

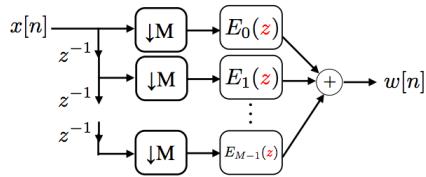


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### Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$



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### Polyphase Implementation of Decimation

$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$

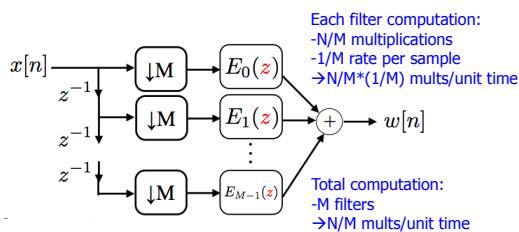
Each filter computation:  
-N/M multiplications  
  
 $x[n] \rightarrow \downarrow M \rightarrow E_0(z) \rightarrow \downarrow M \rightarrow E_1(z) \rightarrow \dots \rightarrow \downarrow M \rightarrow E_{M-1}(z)$   
Total computation:  
-M filters  
→N mults??

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### Polyphase Implementation of Decimation

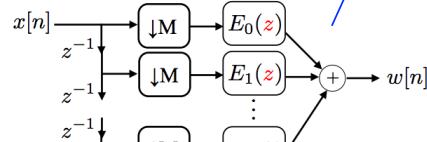
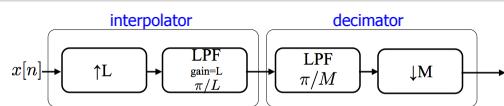
$$x[n] \rightarrow H(z) \rightarrow y[n] \rightarrow \downarrow M \rightarrow w[n] = y[nM]$$



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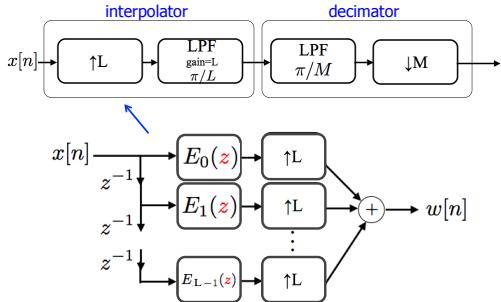
### Polyphase Implementation of Decimator



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### Polyphase Implementation of Interpolation



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### Multi-Rate Filter Banks

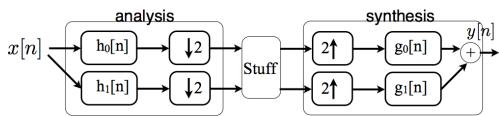
- ❑ Use filter banks to operate on a signal differently in different frequency bands
  - To save computation, reduce the rate after filtering

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### Multi-Rate Filter Banks

- ❑ Use filter banks to operate on a signal differently in different frequency bands
  - To save computation, reduce the rate after filtering
- ❑  $h_0[n]$  is low-pass,  $h_1[n]$  is high-pass
  - Often  $h_1[n] = e^{j\pi n} h_0[n]$  ← shift freq resp by  $\pi$

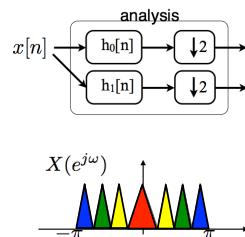


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### Multi-Rate Filter Banks

- ❑ Assume  $h_0, h_1$  are ideal low/high pass with  $\omega_c = \pi/2$

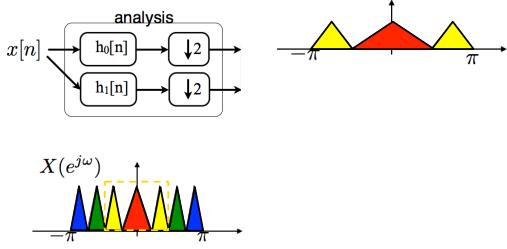


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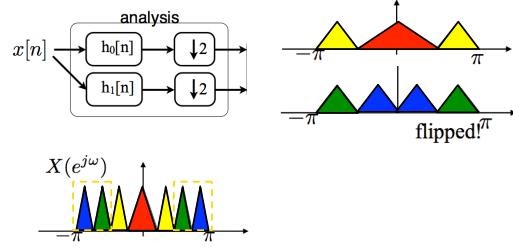


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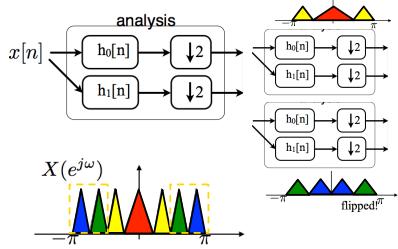


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## Multi-Rate Filter Banks

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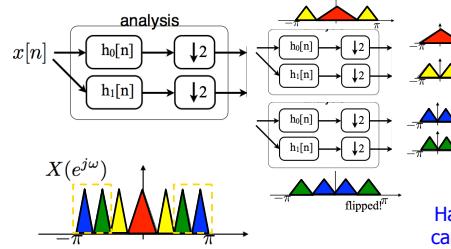


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## Multi-Rate Filter Banks

- Assume  $h_0, h_1$  are ideal low/high pass with  $\omega_c = \pi/2$



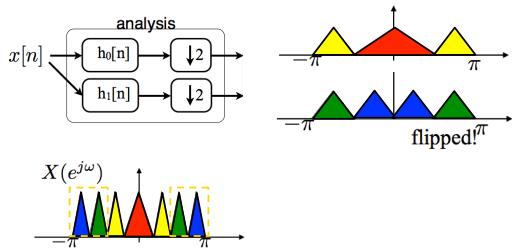
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Have to be  
careful with  
order!

## Multi-Rate Filter Banks

- Assume  $h_0, h_1$  are ideal low/high pass with  $\omega_c = \pi/2$

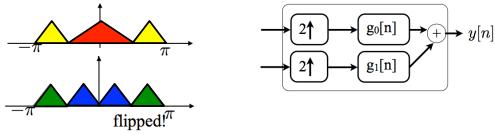


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## Multi-Rate Filter Banks

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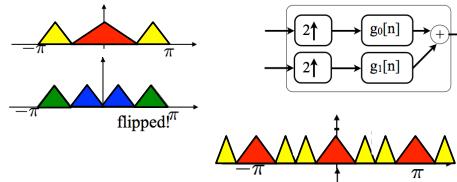


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## Multi-Rate Filter Banks

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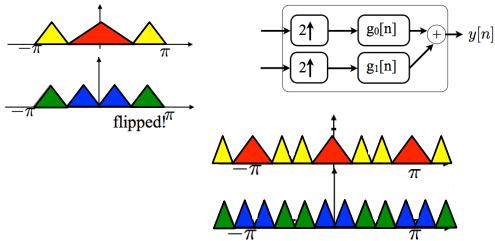


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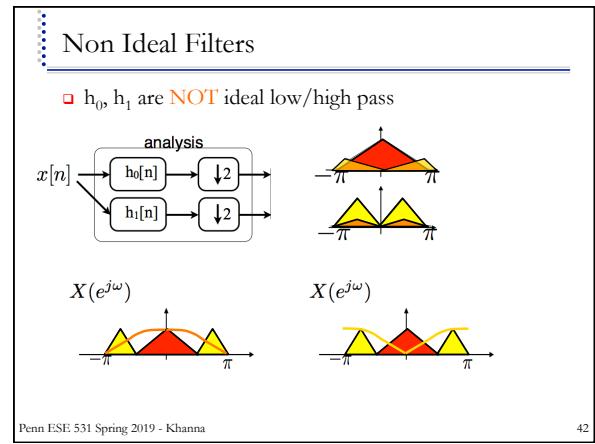
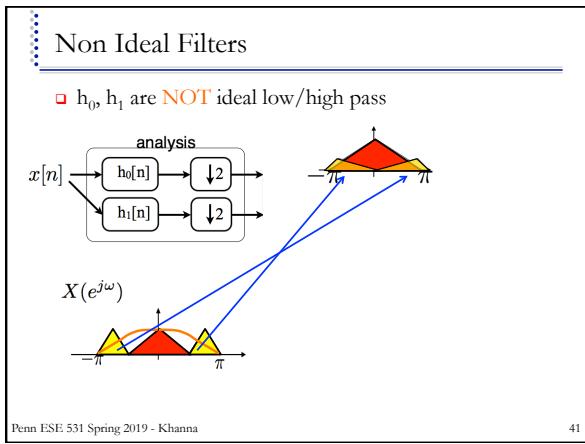
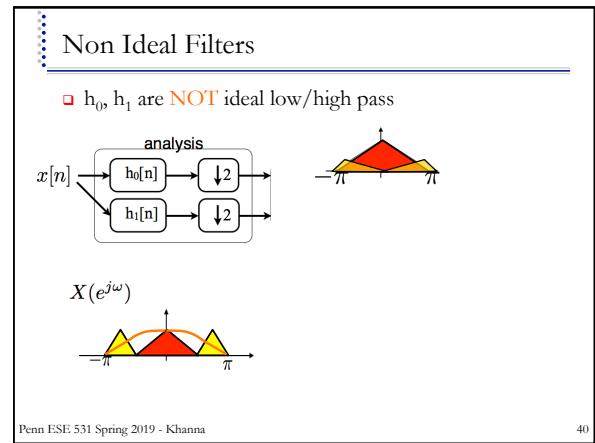
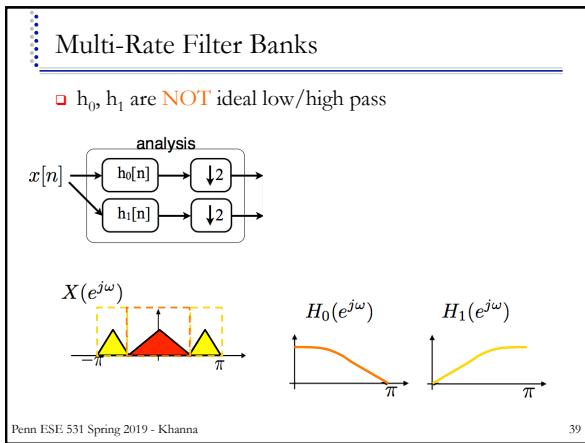
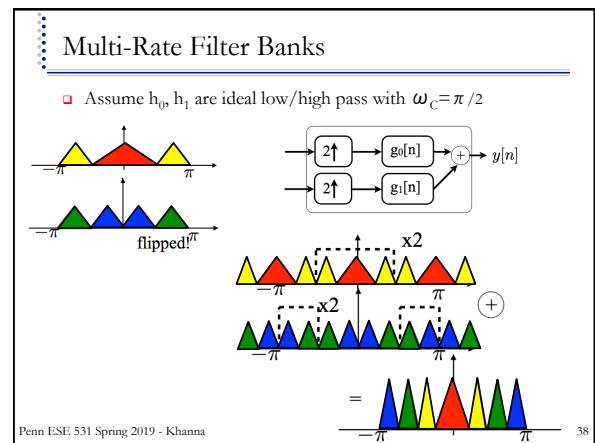
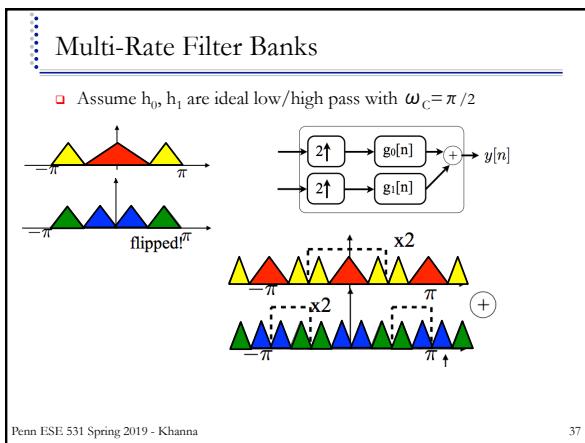
## Multi-Rate Filter Banks

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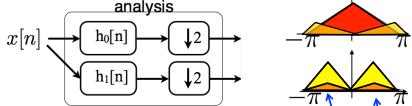
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### Non Ideal Filters

◻  $h_0, h_1$  are NOT ideal low/high pass



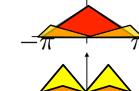
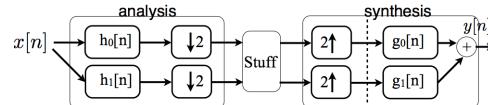
$X(e^{j\omega})$



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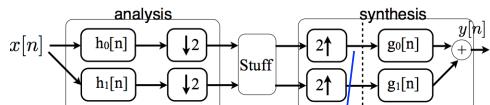
### Non Ideal Filters



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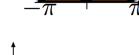
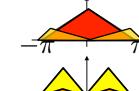
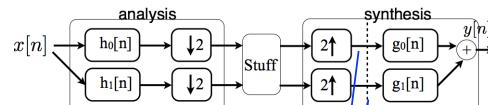
### Non Ideal Filters



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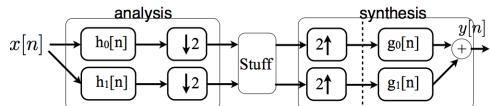
### Non Ideal Filters



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### Perfect Reconstruction non-Ideal Filters



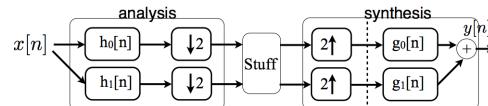
$$Y(e^{j\omega}) = \frac{1}{2} [G_0(e^{j\omega})H_0(e^{j\omega}) + G_1(e^{j\omega})H_1(e^{j\omega})] X(e^{j\omega}) \\ + \frac{1}{2} [G_0(e^{j\omega})H_0(e^{j(\omega-\pi)}) + G_1(e^{j\omega})H_1(e^{j(\omega-\pi)})] X(e^{j(\omega-\pi)})$$

↑ need to cancel!    ↑ aliasing

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### Quadrature Mirror Filters



Quadrature mirror filters

$$H_1(e^{j\omega}) = H_0(e^{j(\omega-\pi)}) \\ G_0(e^{j\omega}) = 2H_0(e^{j\omega}) \\ G_1(e^{j\omega}) = -2H_1(e^{j\omega})$$

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## Perfect Reconstruction non-Ideal Filters

$$Y(e^{j\omega}) = \frac{1}{2} [G_0(e^{j\omega})H_0(e^{j\omega}) + G_1(e^{j\omega})H_1(e^{j\omega})] X(e^{j\omega}) + \frac{1}{2} [G_0(e^{j\omega})H_0(e^{j(\omega-\pi)}) + G_1(e^{j\omega})H_1(e^{j(\omega-\pi)})] X(e^{j(\omega-\pi)})$$

↑    ↓  
need to cancel!                              aliasing

$$\begin{aligned} H_1(e^{j\omega}) &= H_0(e^{j(\omega-\pi)}) \\ G_0(e^{j\omega}) &= 2H_0(e^{j\omega}) \\ G_1(e^{j\omega}) &= -2H_1(e^{j\omega}) \end{aligned}$$

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## Big Ideas

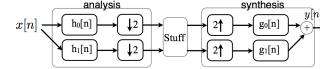
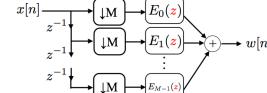
- Interchanging Operations

- Polyphase Decomposition

- Multi-Rate Filter Banks

$$x[n] \xrightarrow{H(z)} \downarrow L \xrightarrow{y[n]} \quad \approx \quad x[n] \xrightarrow{\text{TL}} \downarrow L \xrightarrow{H(z^L)} \xrightarrow{y[n]}$$

$$x[n] \xrightarrow{\downarrow M} \xrightarrow{H(z)} \xrightarrow{y[n]} \quad \approx \quad x[n] \xrightarrow{H(z^M)} \xrightarrow{\downarrow M} \xrightarrow{y[n]}$$



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## Admin

- HW 4 due Sunday

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