

## ESE 531: Digital Signal Processing

Lec 14: March 14th, 2019  
All-Pass Systems and Min Phase  
Decomposition



Penn ESE 531 Spring 2019 – Khanna  
Adapted from M. Lustig, EECS Berkeley

## Lecture Outline

- Review: Frequency Response of LTI Systems
  - Magnitude Response, Phase Response, Group Delay
  - Examples:
    - Zero on Real Axis
    - 2<sup>nd</sup> order IIR
    - 3<sup>rd</sup> order Low Pass
- Stability and Causality
- All Pass Systems
- Minimum Phase Systems

Penn ESE 531 Spring 2019 – Khanna  
Adapted from M. Lustig, EECS Berkeley

2

## Review: Frequency Response of LTI System

$$Y(e^{j\omega}) = H(e^{j\omega})X(e^{j\omega})$$

- We can define a magnitude response...

$$|Y(e^{j\omega})| = |H(e^{j\omega})| |X(e^{j\omega})|$$

- a phase response...

$$\angle Y(e^{j\omega}) = \angle H(e^{j\omega}) + \angle X(e^{j\omega})$$

- and group delay

$$\text{grd}[H(e^{j\omega})] = -\frac{d}{d\omega} \{\arg[H(e^{j\omega})]\}$$

Penn ESE 531 Spring 2019 – Khanna  
Adapted from M. Lustig, EECS Berkeley

3

## Review: Group Delay Math

$$H(z) = \frac{b_0 \prod_{k=1}^M (1 - c_k z^{-1})}{a_0 \prod_{k=1}^N (1 - d_k z^{-1})} \quad H(e^{j\omega}) = \frac{b_0 \prod_{k=1}^M (1 - c_k e^{-j\omega})}{a_0 \prod_{k=1}^N (1 - d_k e^{-j\omega})}$$

- Look at each factor:

$$\arg[1 - r e^{j\theta} e^{-j\omega}] = \tan^{-1} \left( \frac{r \sin(\omega - \theta)}{1 - r \cos(\omega - \theta)} \right)$$

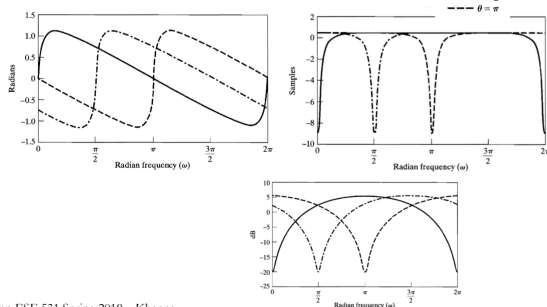
$$\text{grd}[1 - r e^{j\theta} e^{-j\omega}] = \frac{r^2 - r \cos(\omega - \theta)}{|1 - r e^{j\theta} e^{-j\omega}|^2}$$

Penn ESE 531 Spring 2019 – Khanna  
Adapted from M. Lustig, EECS Berkeley

4

## Example: Zero on Real Axis

- For  $\theta \neq 0$   $H(e^{j\omega}) = 1 - r e^{j\theta} e^{-j\omega}$

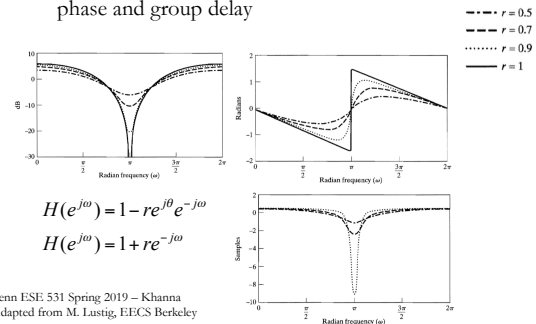


Penn ESE 531 Spring 2019 – Khanna

5

## Example: Zero on Real Axis

- For  $\theta = \pi$ , how zero location affects magnitude, phase and group delay

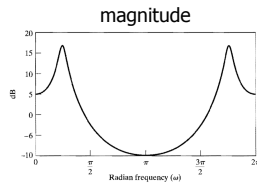


Penn ESE 531 Spring 2019 – Khanna  
Adapted from M. Lustig, EECS Berkeley

6

## 2<sup>nd</sup> Order IIR with Complex Poles

$$H(z) = \frac{1}{(1 - re^{j\theta}z^{-1})(1 - re^{-j\theta}z^{-1})} \quad r=0.9, \theta = \pi/4$$

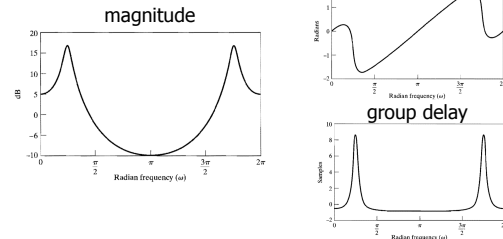


Penn ESE 531 Spring 2019 - Khanna

7

## 2<sup>nd</sup> Order IIR with Complex Poles

$$H(z) = \frac{1}{(1 - re^{j\theta}z^{-1})(1 - re^{-j\theta}z^{-1})} \quad r=0.9, \theta = \pi/4$$

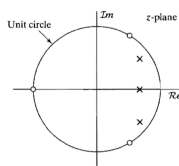


Penn ESE 531 Spring 2019 - Khanna

8

## 3<sup>rd</sup> Order IIR Example

$$H(z) = \frac{0.05634(1 + z^{-1})(1 - 1.0166z^{-1} + z^{-2})}{(1 - 0.683z^{-1})(1 - 1.4461z^{-1} + 0.7957z^{-2})}$$

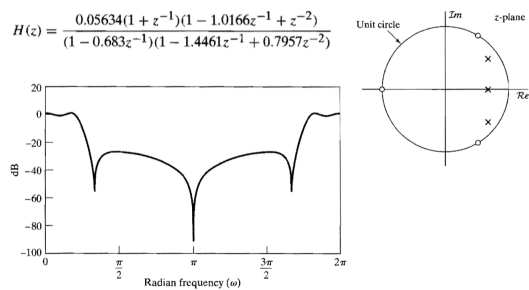


Penn ESE 531 Spring 2019 - Khanna

9

## 3<sup>rd</sup> Order IIR Example

$$H(z) = \frac{0.05634(1 + z^{-1})(1 - 1.0166z^{-1} + z^{-2})}{(1 - 0.683z^{-1})(1 - 1.4461z^{-1} + 0.7957z^{-2})}$$

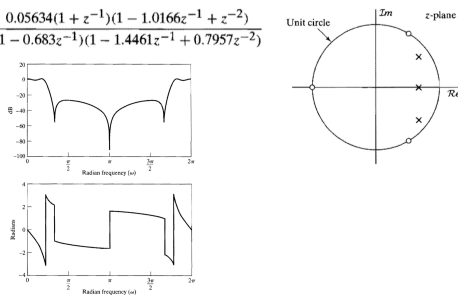


Penn ESE 531 Spring 2019 - Khanna

10

## 3<sup>rd</sup> Order IIR Example

$$H(z) = \frac{0.05634(1 + z^{-1})(1 - 1.0166z^{-1} + z^{-2})}{(1 - 0.683z^{-1})(1 - 1.4461z^{-1} + 0.7957z^{-2})}$$

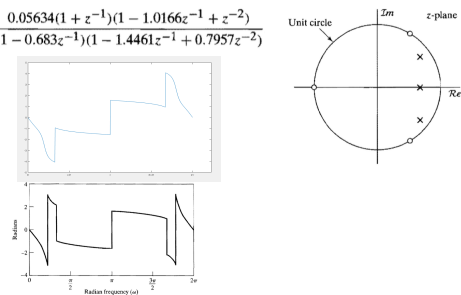


Penn ESE 531 Spring 2019 - Khanna

11

## 3<sup>rd</sup> Order IIR Example

$$H(z) = \frac{0.05634(1 + z^{-1})(1 - 1.0166z^{-1} + z^{-2})}{(1 - 0.683z^{-1})(1 - 1.4461z^{-1} + 0.7957z^{-2})}$$

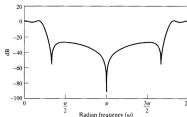
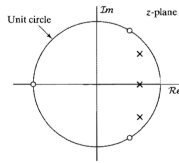
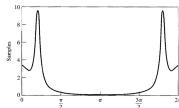
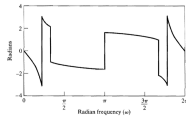


Penn ESE 531 Spring 2019 - Khanna

12

### 3<sup>rd</sup> Order IIR Example

$$H(z) = \frac{0.05634(1+z^{-1})(1-1.0166z^{-1}+z^{-2})}{(1-0.683z^{-1})(1-1.4461z^{-1}+0.7957z^{-2})}$$



Penn ESE 531 Spring 2019 - Khanna

13

### Stability and Causality



Penn ESE 531 Spring 2019 - Khanna

### LTI System

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

Example:  $y[n] = x[n] + 0.1y[n-1]$

$$H(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_M z^{-M}}{a_0 + a_1 z^{-1} + \dots + a_N z^{-N}} = \frac{b_0 \prod_{k=1}^M (1 - c_k z^{-1})}{a_0 \prod_{k=1}^N (1 - d_k z^{-1})}$$

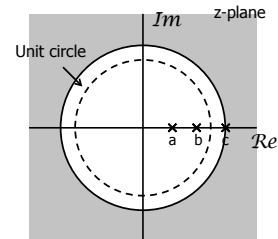
- Transfer function is not unique without ROC
  - If diff. eq represents LTI and causal system, ROC is region outside all singularities
  - If diff. eq represents LTI and stable system, ROC includes unit circle in z-plane

Penn ESE 531 Spring 2019 - Khanna

15

### Example: ROC from Pole-Zero Plot

#### ROC 1: right-sided



Penn ESE 531 Spring 2019 - Khanna

16

### LTI System

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

Example:  $y[n] = x[n] + 0.1y[n-1]$

Stable and causal  
if all poles inside  
unit circle

$$H(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_M z^{-M}}{a_0 + a_1 z^{-1} + \dots + a_N z^{-N}} = \frac{b_0 \prod_{k=1}^M (1 - c_k z^{-1})}{a_0 \prod_{k=1}^N (1 - d_k z^{-1})}$$

- Transfer function is not unique without ROC
  - If diff. eq represents LTI and causal system, ROC is region outside all singularities
  - If diff. eq represents LTI and stable system, ROC includes unit circle in z-plane

Penn ESE 531 Spring 2019 - Khanna

17

### All-Pass Systems



Penn ESE 531 Spring 2019 - Khanna

## All-Pass Filters

- A system is an all-pass system if

$$|H(e^{j\omega})| = 1, \text{ all } \omega$$

- Its phase response  $\theta(\omega)$  may be non-trivial

## First Order All-Pass Filter ( $a$ real)

$$H(z) = \frac{z^{-1} - a^*}{1 - az^{-1}}$$

## First Order All-Pass Filter ( $a$ real)

$$H(z) = \frac{z^{-1} - a^*}{1 - az^{-1}}$$

$$\begin{aligned} |H(e^{j\omega})| &= \left| \frac{e^{-j\omega} - a^*}{1 - ae^{-j\omega}} \right| \\ &= \left| \frac{e^{-j\omega}(1 - a^*e^{j\omega})}{1 - ae^{-j\omega}} \right| \\ &= \frac{|1 - a^*e^{j\omega}|}{|1 - ae^{-j\omega}|} = 1 \end{aligned}$$

## General All-Pass Filter

- $d_k$ =real pole,  $e_k$ =complex poles paired w/ conjugate,  $e_k^*$

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

## General All-Pass Filter

- $d_k$ =real pole,  $e_k$ =complex poles paired w/ conjugate,  $e_k^*$

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

## General All-Pass Filter

- $d_k$ =real pole,  $e_k$ =complex poles paired w/ conjugate,  $e_k^*$

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

## General All-Pass Filter

- $d_k$ =real pole,  $e_k$ =complex poles paired w/ conjugate,  $e_k^*$

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

Penn ESE 531 Spring 2019 - Khanna

25

## General All-Pass Filter

- $d_k$ =real pole,  $e_k$ =complex poles paired w/ conjugate,  $e_k^*$

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

- Example:

$$d_k = -\frac{3}{4}$$

$$e_k = 0.8e^{j\pi/4}$$

Penn ESE 531 Spring 2019 - Khanna

26

## General All-Pass Filter

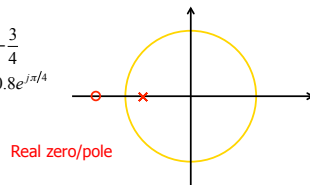
- $d_k$ =real pole,  $e_k$ =complex poles paired w/ conjugate,  $e_k^*$

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

- Example:

$$d_k = -\frac{3}{4}$$

$$e_k = 0.8e^{j\pi/4}$$



Penn ESE 531 Spring 2019 - Khanna

27

## General All-Pass Filter

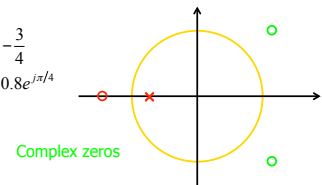
- $d_k$ =real pole,  $e_k$ =complex poles paired w/ conjugate,  $e_k^*$

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

- Example:

$$d_k = -\frac{3}{4}$$

$$e_k = 0.8e^{j\pi/4}$$



Penn ESE 531 Spring 2019 - Khanna

28

## General All-Pass Filter

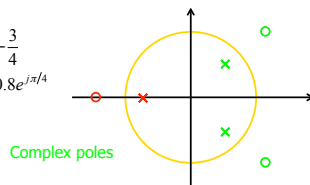
- $d_k$ =real pole,  $e_k$ =complex poles paired w/ conjugate,  $e_k^*$

$$H_{ap}(z) = A \prod_{k=1}^{M_r} \frac{z^{-1} - d_k}{1 - d_k z^{-1}} \prod_{k=1}^{M_c} \frac{(z^{-1} - e_k^*)(z^{-1} - e_k)}{(1 - e_k z^{-1})(1 - e_k^* z^{-1})}$$

- Example:

$$d_k = -\frac{3}{4}$$

$$e_k = 0.8e^{j\pi/4}$$



Penn ESE 531 Spring 2019 - Khanna

29

## All Pass Filter Phase Response

- First order system  $H(e^{j\omega}) = \frac{e^{-j\omega} - a^*}{1 - ae^{-j\omega}} = \frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}$

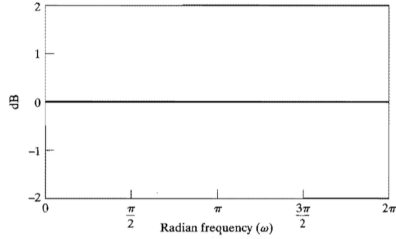
- phase

Penn ESE 531 Spring 2019 - Khanna

30

## First Order Example

- Magnitude:



Penn ESE 531 Spring 2019 - Khanna

31

## All Pass Filter Phase Response

- First order system  $H(e^{j\omega}) = \frac{e^{-j\omega} - a^*}{1 - ae^{-j\omega}}$   

$$= \frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}$$
- phase  $\arg\left(\frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}\right)$

Penn ESE 531 Spring 2019 - Khanna

32

## All Pass Filter Phase Response

- First order system  $H(e^{j\omega}) = \frac{e^{-j\omega} - a^*}{1 - ae^{-j\omega}}$   

$$= \frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}$$
- phase  $\arg\left(\frac{e^{-j\omega} - re^{-j\theta}}{1 - re^{j\theta}e^{-j\omega}}\right)$   

$$= \arg\left(\frac{e^{-j\omega}(1 - re^{-j\theta}e^{j\omega})}{1 - re^{j\theta}e^{-j\omega}}\right)$$
  

$$= \arg(e^{-j\omega}) + \arg(1 - re^{-j\theta}e^{j\omega}) - \arg(1 - re^{j\theta}e^{-j\omega})$$
  

$$= -\omega - \arg(1 - re^{j\theta}e^{-j\omega}) - \arg(1 - re^{j\theta}e^{-j\omega})$$
  

$$= -\omega - 2\arg(1 - re^{j\theta}e^{-j\omega})$$

Penn ESE 531 Spring 2019 - Khanna

33

## Group Delay Math

$$\text{grd}[H(e^{j\omega})] = \sum_{k=1}^M \text{grd}[1 - c_k e^{-j\omega}] - \sum_{k=1}^N \text{grd}[1 - d_k e^{-j\omega}]$$

- Look at each factor:

$$\arg[1 - re^{j\theta}e^{-j\omega}] = \tan^{-1}\left(\frac{r \sin(\omega - \theta)}{1 - r \cos(\omega - \theta)}\right)$$

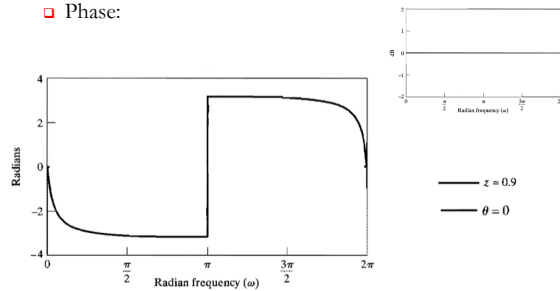
$$\text{grd}[1 - re^{j\theta}e^{-j\omega}] = \frac{r^2 - r \cos(\omega - \theta)}{|1 - re^{j\theta}e^{-j\omega}|^2}$$

Penn ESE 531 Spring 2019 - Khanna  
Adapted from M. Lustig, EECS Berkeley

34

## First Order Example

- Phase:

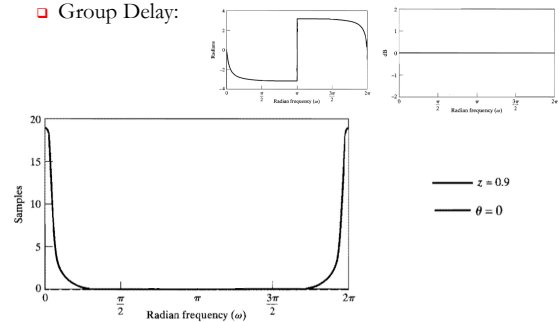


Penn ESE 531 Spring 2019 - Khanna

35

## First Order Example

- Group Delay:



Penn ESE 531 Spring 2019 - Khanna

36

## All Pass Filter Phase Response

- Second order system with poles at  $z = re^{j\theta}, re^{-j\theta}$

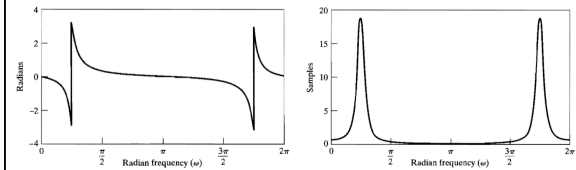
$$\angle \left[ \frac{(e^{-j\omega} - re^{-j\theta})(e^{-j\omega} - re^{j\theta})}{(1 - re^{j\theta}e^{-j\omega})(1 - re^{-j\theta}e^{-j\omega})} \right] = -2\omega - 2 \arctan \left[ \frac{r \sin(\omega - \theta)}{1 - r \cos(\omega - \theta)} \right] - 2 \arctan \left[ \frac{r \sin(\omega + \theta)}{1 - r \cos(\omega + \theta)} \right]$$

Penn ESE 531 Spring 2019 - Khanna

37

## Second Order Example

- Poles at  $z = 0.9e^{\pm j\pi/4}$  (zeros at conjugates)



Penn ESE 531 Spring 2019 - Khanna

38

## All-Pass Properties

- Claim: For a stable, causal ( $r < 1$ ) all-pass system:
  - $\arg[H_{ap}(e^{j\omega})] \leq 0$ 
    - Unwrapped phase always non-positive and decreasing
  - $\text{grd}[H_{ap}(e^{j\omega})] > 0$ 
    - Group delay always positive [Pg 309 in text book](#)
- Intuition
  - delay is positive  $\rightarrow$  system is causal
  - Phase negative  $\rightarrow$  phase lag

Penn ESE 531 Spring 2019 - Khanna

39

## Minimum-Phase Systems



## Minimum-Phase Systems

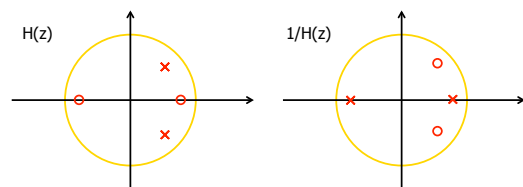
- Definition: A stable and causal system  $H(z)$  (i.e. **poles inside unit circle**) whose inverse  $1/H(z)$  is also stable and causal (i.e. **zeros inside unit circle**)

Penn ESE 531 Spring 2019 - Khanna

41

## Minimum-Phase Systems

- Definition: A stable and causal system  $H(z)$  (i.e. **poles inside unit circle**) whose inverse  $1/H(z)$  is also stable and causal (i.e. **zeros inside unit circle**)
  - All poles and zeros inside unit circle**



Penn ESE 531 Spring 2019 - Khanna

42

## All-Pass Min-Phase Decomposition

- Any stable, causal system can be decomposed to:

$$H(z) = H_{\min}(z) \cdot H_{ap}(z)$$

- Approach:

- (1) First construct  $H_{ap}$  with all zeros outside unit circle
- (2) Compute

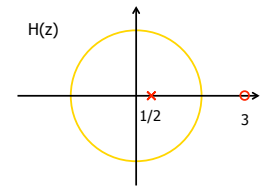
$$H_{\min}(z) = \frac{H(z)}{H_{ap}(z)}$$

Penn ESE 531 Spring 2019 - Khanna

43

## Min-Phase Decomposition Example

$$H(z) = \frac{1-3z^{-1}}{1-\frac{1}{2}z^{-1}}$$



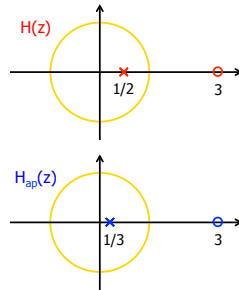
Penn ESE 531 Spring 2019 - Khanna

44

## Min-Phase Decomposition Example

$$H(z) = \frac{1-3z^{-1}}{1-\frac{1}{2}z^{-1}}$$

- Set  $H_{ap}(z) = \frac{z^{-1} - \frac{1}{3}}{1 - \frac{1}{3}z^{-1}}$



Penn ESE 531 Spring 2019 - Khanna

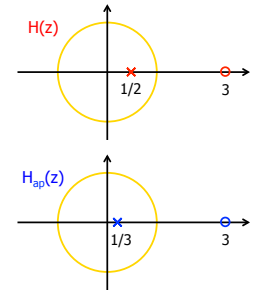
45

## Min-Phase Decomposition Example

$$H(z) = \frac{1-3z^{-1}}{1-\frac{1}{2}z^{-1}}$$

- Set  $H_{ap}(z) = \frac{z^{-1} - \frac{1}{3}}{1 - \frac{1}{3}z^{-1}}$

$$H_{\min}(z) = \frac{1-3z^{-1}}{1-\frac{1}{2}z^{-1}} \left( \frac{z^{-1} - \frac{1}{3}}{1 - \frac{1}{3}z^{-1}} \right)^{-1}$$



Penn ESE 531 Spring 2019 - Khanna

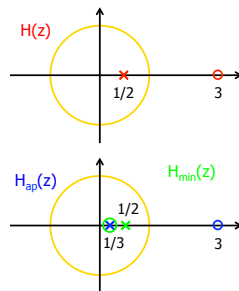
46

## Min-Phase Decomposition Example

$$H(z) = \frac{1-3z^{-1}}{1-\frac{1}{2}z^{-1}}$$

- Set  $H_{ap}(z) = \frac{z^{-1} - \frac{1}{3}}{1 - \frac{1}{3}z^{-1}}$

$$H_{\min}(z) = -3 \frac{1 - \frac{1}{2}z^{-1}}{1 - \frac{1}{3}z^{-1}}$$

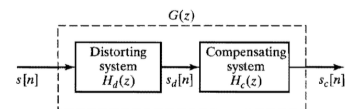


Penn ESE 531 Spring 2019 - Khanna

47

## Min-Phase Decomposition Purpose

- Have some distortion that we want to compensate for:

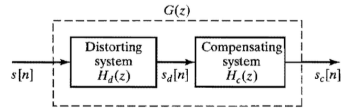


Penn ESE 531 Spring 2019 - Khanna

48

## Min-Phase Decomposition Purpose

- Have some distortion that we want to compensate for:



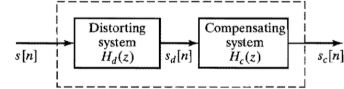
- If  $H_d(z)$  is min phase, easy:
  - $H_c(z) = 1/H_d(z)$  ← also stable and causal

Penn ESE 531 Spring 2019 - Khanna

49

## Min-Phase Decomposition Purpose

- Have some distortion that we want to compensate for:

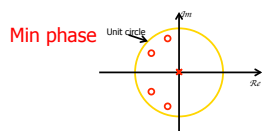


- If  $H_d(z)$  is min phase, easy:
  - $H_c(z) = 1/H_d(z)$  ← also stable and causal
- Else, decompose  $H_d(z) = H_{d,min}(z) H_{d,ap}(z)$ 
  - $H_c(z) = 1/H_{d,min}(z) \rightarrow H_d(z)H_c(z) = H_{d,ap}(z)$ 
    - Compensate for magnitude distortion

Penn ESE 531 Spring 2019 - Khanna

50

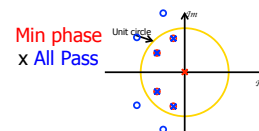
## Minimum Phase to Max Phase



Penn ESE 531 Spring 2018 - Khanna

51

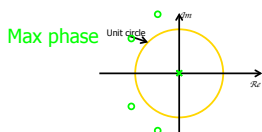
## Minimum Phase to Max Phase



Penn ESE 531 Spring 2018 - Khanna

52

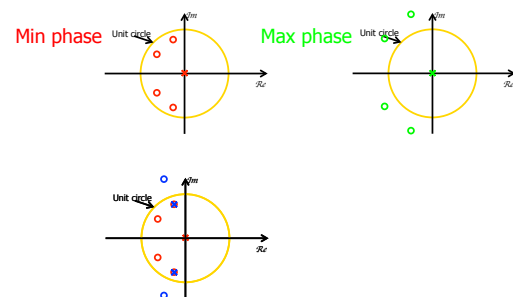
## Minimum Phase to Max Phase



Penn ESE 531 Spring 2018 - Khanna

53

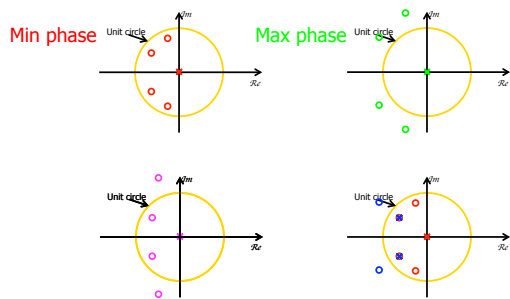
## Minimum Phase



Penn ESE 531 Spring 2018 - Khanna

54

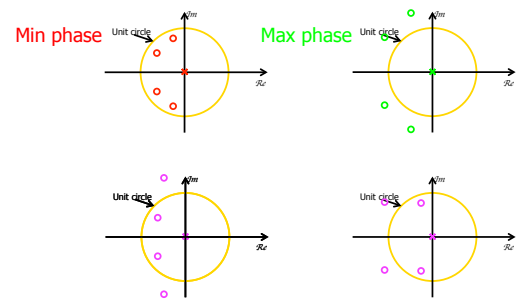
## Minimum Phase



Penn ESE 531 Spring 2018 - Khanna

55

## Minimum Phase



Penn ESE 531 Spring 2018 - Khanna

56

## Minimum Phase Lag Property

- All pass properties
  - $\arg[H_{ap}(e^{j\omega})] \leq 0$
  - $\text{grd}[H_{ap}(e^{j\omega})] > 0$
- $\arg[H_{\max}(e^{j\omega})] = \arg[H_{\min}(e^{j\omega}) * H_{ap}(e^{j\omega})]$   
 $= \arg[H_{\min}(e^{j\omega})] + \arg[H_{ap}(e^{j\omega})]$   
 $= \leq 0 + \leq 0$

Penn ESE 531 Spring 2018 - Khanna

57

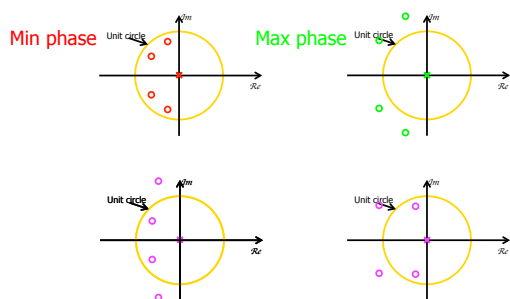
## Minimum Group Delay Property

- All pass properties
  - $\arg[H_{ap}(e^{j\omega})] \leq 0$
  - $\text{grd}[H_{ap}(e^{j\omega})] > 0$
- $\text{grd}[H_{\max}(e^{j\omega})] = \text{grd}[H_{\min}(e^{j\omega}) * H_{ap}(e^{j\omega})]$   
 $= \text{grd}[H_{\min}(e^{j\omega})] + \text{grd}[H_{ap}(e^{j\omega})]$   
 $= \geq 0 + \geq 0$

Penn ESE 531 Spring 2018 - Khanna

58

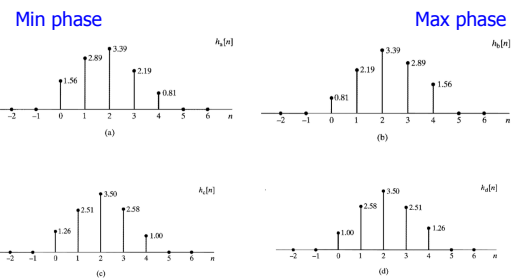
## Minimum Phase



Penn ESE 531 Spring 2018 - Khanna

59

## Minimum Energy-Delay Property



Penn ESE 531 Spring 2018 - Khanna

60

## Big Ideas

- Frequency Response of LTI Systems
  - Magnitude Response, Phase Response, Group Delay
- LTI Stability and Causality
  - If all poles inside unit circle
- All Pass Systems
  - Used for delay compensation
- Minimum Phase Systems
  - Can compensate for magnitude distortion
  - Minimum energy-delay property

Penn ESE 531 Spring 2019 – Khanna  
Adapted from M. Lustig, EECS Berkeley

61

## Admin

- HW 6
  - Out now
  - Due Sunday 3/24

Penn ESE 531 Spring 2019 - Khanna

62