

ESE 531: Digital Signal Processing

Lec 9: February 13th, 2020
Downsampling/Upsampling and Practical Interpolation



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Lecture Outline

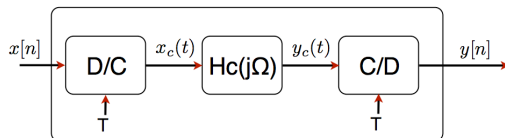
- CT processing of DT signals
- Downsampling
- Upsampling
- Practical Interpolation (time permitting)

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Continuous-Time Processing of Discrete-Time

- Useful to interpret DT systems with no simple interpretation in discrete time

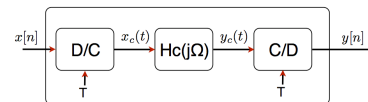


Is the effective $H(e^{j\omega})$ LTI?

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Continuous-Time Processing of Discrete-Time

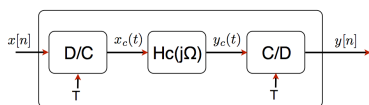


$$X_c(j\Omega) = \begin{cases} TX(e^{j\Omega T}) & |\Omega| < \pi/T \\ 0 & \text{else} \end{cases}$$

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Continuous-Time Processing of Discrete-Time



$$X_c(j\Omega) = \begin{cases} TX(e^{j\Omega T}) & |\Omega| < \pi/T \\ 0 & \text{else} \end{cases}$$

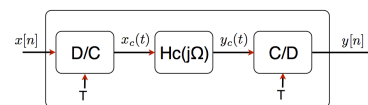
Also bandlimited

$$\rightarrow Y_c(j\Omega) = H_c(j\Omega)X_c(j\Omega)$$

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Continuous-Time Processing of Discrete-Time



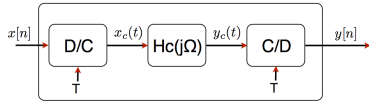
$$Y_c(j\Omega) = H_c(j\Omega)X_c(j\Omega)$$

$$Y(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} Y_c(j(\Omega - k\Omega_s)) \Big|_{\Omega=\omega/T}$$

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Continuous-Time Processing of Discrete-Time



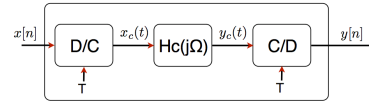
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$$Y(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} Y_c(j\Omega - k\Omega_s) \Big|_{\Omega=\omega/T} = \frac{1}{T} Y_c(j\Omega) \Big|_{\Omega=\omega/T}$$

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Continuous-Time Processing of Discrete-Time

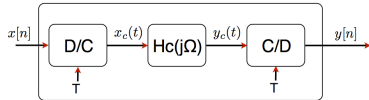


$$Y_c(j\Omega) = H_c(j\Omega)X_c(j\Omega) \quad Y(e^{j\omega}) = \frac{1}{T} Y_c(j\Omega) \Big|_{\Omega=\omega/T}$$

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Continuous-Time Processing of Discrete-Time



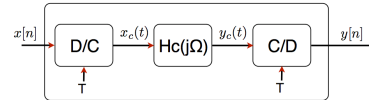
$$Y_c(j\Omega) = H_c(j\Omega)X_c(j\Omega) \quad Y(e^{j\omega}) = \frac{1}{T} Y_c(j\Omega) \Big|_{\Omega=\omega/T}$$

$$Y(e^{j\omega}) = \frac{1}{T} H_c(j\Omega) \Big|_{\Omega=\omega/T} X_c(j\Omega) \Big|_{\Omega=\omega/T}$$

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Continuous-Time Processing of Discrete-Time



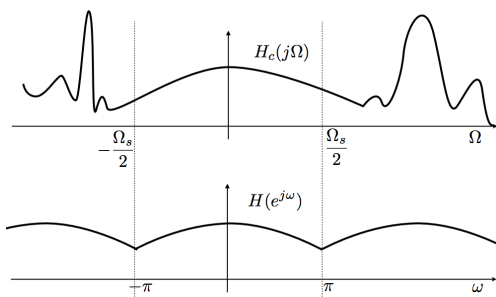
$$Y_c(j\Omega) = H_c(j\Omega)X_c(j\Omega) \quad Y(e^{j\omega}) = \frac{1}{T} Y_c(j\Omega) \Big|_{\Omega=\omega/T}$$

$$Y(e^{j\omega}) = \frac{1}{T} H_c(j\Omega) \Big|_{\Omega=\omega/T} X_c(j\Omega) \Big|_{\Omega=\omega/T} \\ = \frac{1}{T} (H_c(j\Omega) \Big|_{\Omega=\omega/T}) (TX(e^{j\omega})) \quad |\omega| < \pi \\ \boxed{H(e^{j\omega})}$$

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Example



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Example: Non-integer Delay

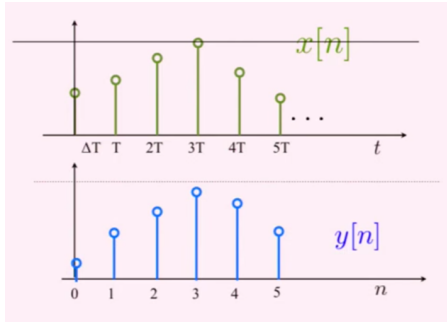
- What is the time domain operation when Δ is non-integer? I.e. $\Delta=1/2$

$$H(e^{j\omega}) = e^{-j\omega\Delta} \quad \begin{matrix} \delta[n] \leftrightarrow 1 \\ \delta[n-n_d] \leftrightarrow e^{-j\omega n_d} \end{matrix}$$

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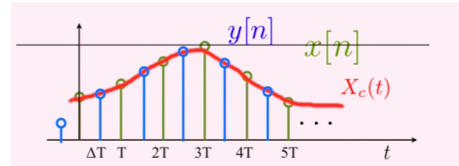
Example: Non-integer Delay



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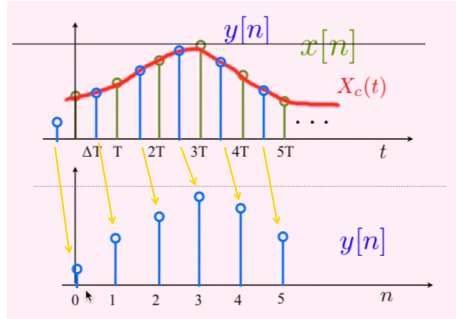
Example: Non-integer Delay



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Example: Non-integer Delay



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Example: Non-integer Delay

- What is the time domain operation when Δ is non-integer? I.e $\Delta=1/2$

$$H(e^{j\omega}) = e^{-j\omega\Delta}$$

Let: $H_c(j\Omega) = e^{-j\Omega\Delta T}$ delay of ΔT in continuous time

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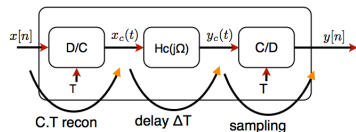
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Example: Non-integer Delay

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Let: $H_c(j\Omega) = e^{-j\Omega\Delta T}$ delay of ΔT in continuous time

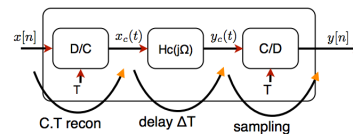


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Example: Non-integer Delay

- The block diagram is for interpretation/analysis only



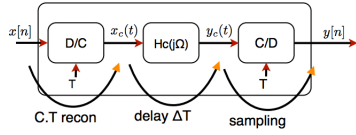
$$x_c(t) = \sum_k x[k] \text{sinc}\left(\frac{t-kT}{T}\right)$$

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Example: Non-integer Delay

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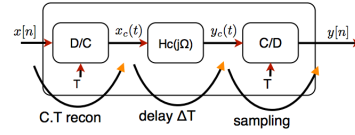
$$x_c(t) = \sum_k x[k] \text{sinc}\left(\frac{t-kT}{T}\right) \quad y_c(t) = x_c(t-T\Delta)$$

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Example: Non-integer Delay

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$$x_c(t) = \sum_k x[k] \text{sinc}\left(\frac{t-kT}{T}\right) \quad y_c(t) = x_c(t-T\Delta)$$

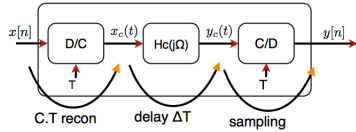
$$y[n] = y_c(nT)$$

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Example: Non-integer Delay

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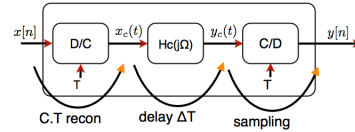
$$y[n] = y_c(nT) = x_c(nT - T \cdot \Delta) \quad x_c(t) = \sum_k x[k] \text{sinc}\left(\frac{t-kT}{T}\right)$$

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Example: Non-integer Delay

- The block diagram is for interpretation/analysis only



$$y[n] = y_c(nT) = x_c(nT - T \cdot \Delta) \quad x_c(t) = \sum_k x[k] \text{sinc}\left(\frac{t-kT}{T}\right)$$

$$x_c(nT - T \cdot \Delta) = \sum_k x[k] \text{sinc}\left(\frac{nT - T \cdot \Delta - kT}{T}\right) = \sum_k x[k] \text{sinc}(n - \Delta - k)$$

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Example: Non-integer Delay

- Delay system has an impulse response of a sinc with a continuous time delay

$$y[n] = \sum_k x[k] \text{sinc}(n - \Delta - k) \\ = x[n] * \text{sinc}(n - \Delta)$$

$$\Rightarrow h[n] = \text{sinc}(n - \Delta)$$

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Example: Non-integer Delay

- What is the time domain operation when Δ is non-integer? I.e. $\Delta = 1/2$

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Example: Non-integer Delay

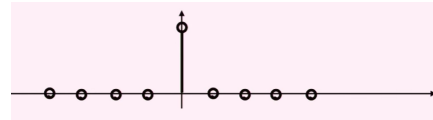
- My delay system has an impulse response of a sinc with a continuous time delay

$$h[n] = \text{sinc}(n - \Delta)$$

Example: Non-integer Delay

- My delay system has an impulse response of a sinc with a continuous time delay

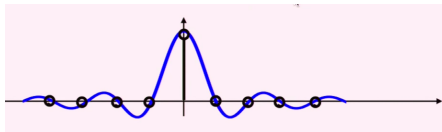
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Example: Non-integer Delay

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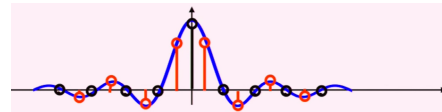
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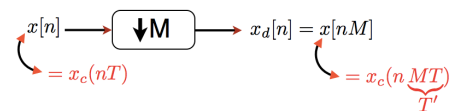
$$h[n] = \text{sinc}(n - \Delta)$$



Downsampling

Downsampling

- Definition: Reducing the sampling rate by an integer number ($M > 1$)



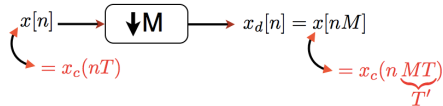
Downsampling

- Similar to C/D conversion
 - Need to worry about aliasing
 - Use anti-aliasing filter to mitigate effects

Downsampling

- Similar to C/D conversion
 - Need to worry about aliasing
 - Use anti-aliasing filter to mitigate effects
- If your discrete time signal is finely sampled (i.e. oversampled) almost like a CT signal
 - Downsampling is just like sampling (C/D conversion)

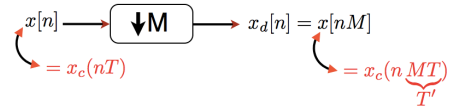
Downsampling



The DTFT:

$$X(e^{j\omega}) = \frac{1}{T} \sum_k X_c \left(j \left(\underbrace{\frac{\omega}{T}}_{\Omega} - \underbrace{\frac{2\pi}{T}}_{\Omega_s} k \right) \right)$$

Downsampling



The DTFT:

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$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right)$$

Downsampling

The DTFT:

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$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right)$$

- Want to relate $X_d(e^{j\omega})$ to $X(e^{j\omega})$ not $X_c(j\Omega)$
- Separate sum into two sums—fine sum and coarse sum (i.e. like counting minutes within hours)

Downsampling

The DTFT:

$$X(e^{j\omega}) = \frac{1}{T} \sum_k X_c \left(j \left(\underbrace{\frac{\omega}{T}}_{\Omega} - \underbrace{\frac{2\pi}{T}}_{\Omega_s} k \right) \right)$$

$$X_d(e^{j\omega}) = \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right)$$

- $k = rM + i$
 - $i = 0, 1, \dots, M-1$
 - $r = -\infty, \dots, \infty$

Downsampling

$$\begin{aligned}
 X_d(e^{j\omega}) &= \frac{1}{MT} \sum_k X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} k \right) \right) \\
 &= \frac{1}{M} \sum_{i=0}^{M-1} \underbrace{\frac{1}{T} \sum_{r=-\infty}^{\infty} X_c \left(j \left(\frac{\omega}{MT} - \frac{2\pi}{MT} i - \frac{2\pi}{T} r \right) \right)}_{X(e^{j(\frac{\omega}{M} - \frac{2\pi}{M} i)})} \\
 X(e^{j\omega}) &= \frac{1}{T} \sum_k X_c \left(j \left(\frac{\omega}{T} - \frac{2\pi}{T} k \right) \right) \\
 X_d(e^{j\omega}) &= \frac{1}{M} \sum_{i=0}^{M-1} X(e^{j(\frac{\omega}{M} - \frac{2\pi}{M} i)})
 \end{aligned}$$

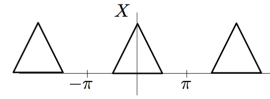
scale by 1/M stretch by M replicate

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Example: M=2

$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X(e^{j(\frac{\omega}{M} - \frac{2\pi}{M} i)})$$

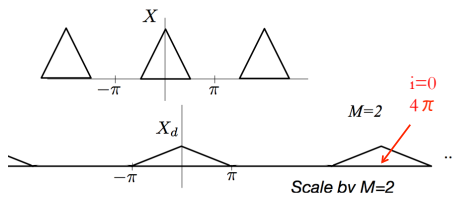


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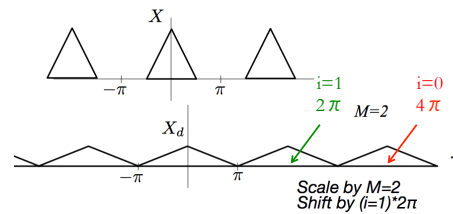


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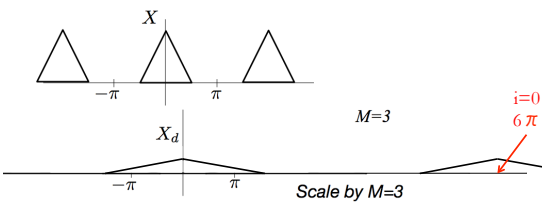


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Example: M=3

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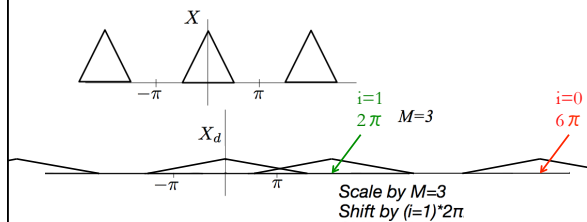


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Example: M=3

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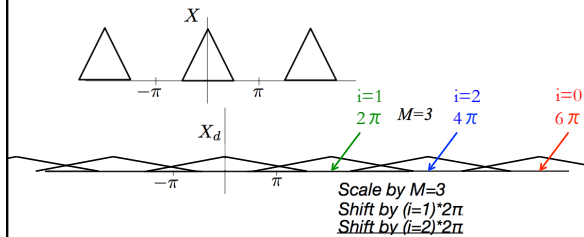


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Example: M=3

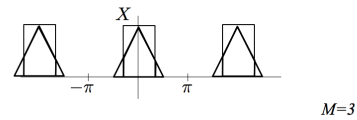
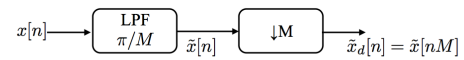
$$X_d(e^{j\omega}) = \frac{1}{M} \sum_{i=0}^{M-1} X(e^{j(\frac{\omega}{M} - \frac{2\pi}{M}i)})$$



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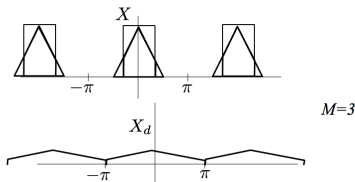
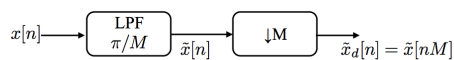
Example: M=3



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Example: M=3



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Upsampling



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Upsampling

- Definition: Increasing the sampling rate by an integer number

$$x[n] = x_c(nT)$$

$$x_i[n] = x_c(nT') \quad \text{where } T' = \frac{T}{L} \quad L \text{ integer}$$

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Upsampling

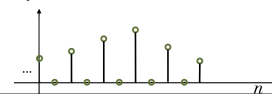
- Definition: Increasing the sampling rate by an integer number

$$x[n] = x_c(nT)$$

$$x_i[n] = x_c(nT') \quad \text{where } T' = \frac{T}{L} \quad L \text{ integer}$$

Obtain $x_i[n]$ from $x[n]$ in two steps:

$$(1) \text{ Generate: } x_e[n] = \begin{cases} x[n/L] & n = 0, \pm L, \pm 2L, \dots \\ 0 & \text{otherwise} \end{cases}$$

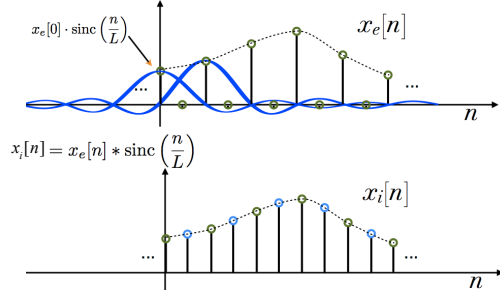


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Upsampling

(2) Obtain $x_i[n]$ from $x_e[n]$ by bandlimited interpolation:



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Upsampling

- Much like D/C converter
- Upsample by A LOT $\rightarrow$ almost continuous
- Intuition:
 - Recall our D/C model: $x[n] \rightarrow x_s(t) \rightarrow x_c(t)$
 - Approximate " $x_s(t)$ " by placing zeros between samples
 - Convolve with a sinc to obtain " $x_c(t)$ "

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Upsampling

$$x_e[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n - kL]$$

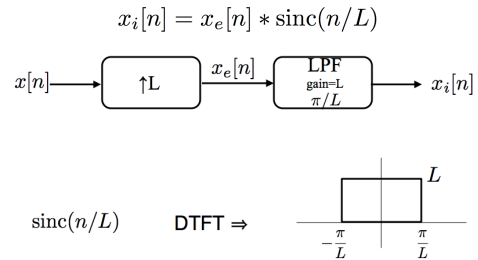
$$x_i[n] = x_e[n] * \text{sinc}(n/L)$$

$$x_i[n] = \sum_{k=-\infty}^{\infty} x[k] \text{sinc}\left(\frac{n - kL}{L}\right)$$

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Frequency Domain Interpretation



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Frequency Domain Interpretation

$$X_e(e^{j\omega}) = \sum_{n=-\infty}^{\infty} \underbrace{x_e[n]}_{\neq 0 \text{ only for } n=mL \text{ (integer } m)} e^{-j\omega n}$$

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Frequency Domain Interpretation

$$X_e(e^{j\omega}) = \sum_{n=-\infty}^{\infty} \underbrace{x_e[n]}_{\neq 0 \text{ only for } n=mL \text{ (integer } m)} e^{-j\omega n}$$

$$= \sum_{m=-\infty}^{\infty} \underbrace{x_e[mL]} e^{-j\omega mL}$$

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Frequency Domain Interpretation

$$\begin{aligned}
 & x[n] \rightarrow \uparrow L \rightarrow x_e[n] \rightarrow \text{LPF}_{\text{gain}=L, \pi/L} \rightarrow x_i[n] \\
 & X_e(e^{j\omega}) = \sum_{n=-\infty}^{\infty} \underbrace{x_e[n]}_{\substack{\neq 0 \text{ only for } n=mL \\ \text{(integer } m)}} e^{-j\omega n} \\
 & = \sum_{m=-\infty}^{\infty} \underbrace{x_e[mL]}_{=x[m]} e^{-j\omega mL} = X(e^{j\omega L})
 \end{aligned}$$

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Frequency Domain Interpretation

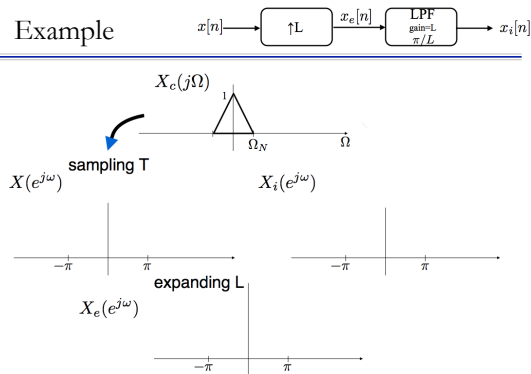
$$\begin{aligned}
 & x[n] \rightarrow \uparrow L \rightarrow x_e[n] \rightarrow \text{LPF}_{\text{gain}=L, \pi/L} \rightarrow x_i[n] \\
 & X_e(e^{j\omega}) = \sum_{n=-\infty}^{\infty} \underbrace{x_e[n]}_{\substack{\neq 0 \text{ only for } n=mL \\ \text{(integer } m)}} e^{-j\omega n} \\
 & = \sum_{m=-\infty}^{\infty} \underbrace{x_e[mL]}_{=x[m]} e^{-j\omega mL} = X(e^{j\omega L})
 \end{aligned}$$

Shrink DTFT by a factor of L!

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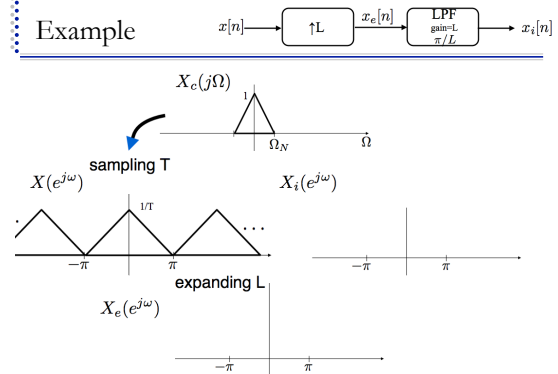
Example



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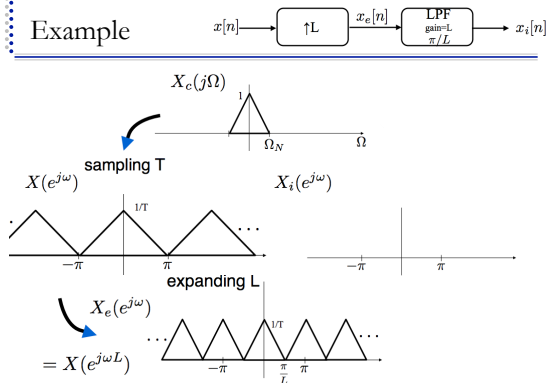
Example



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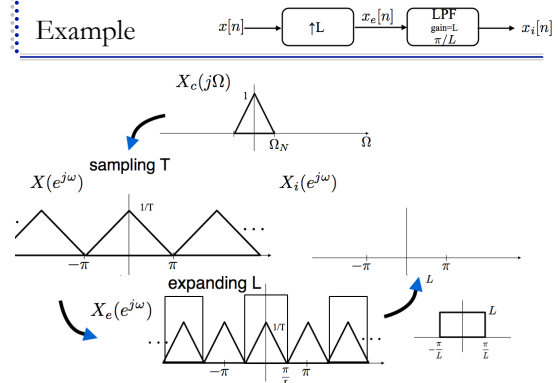
Example



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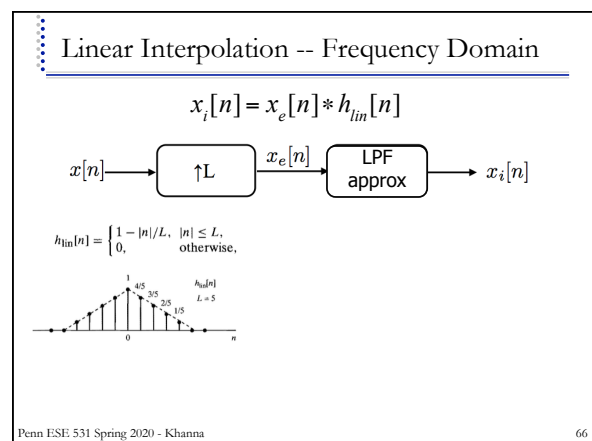
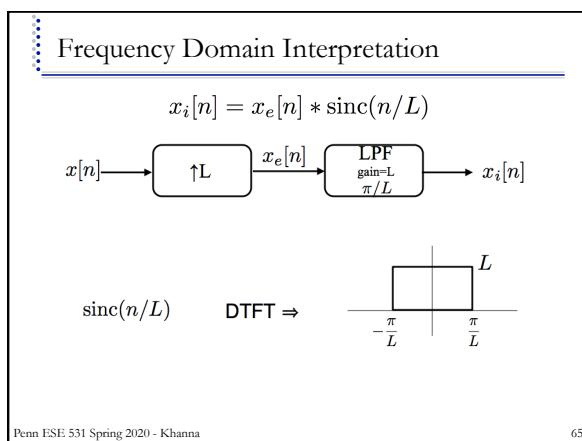
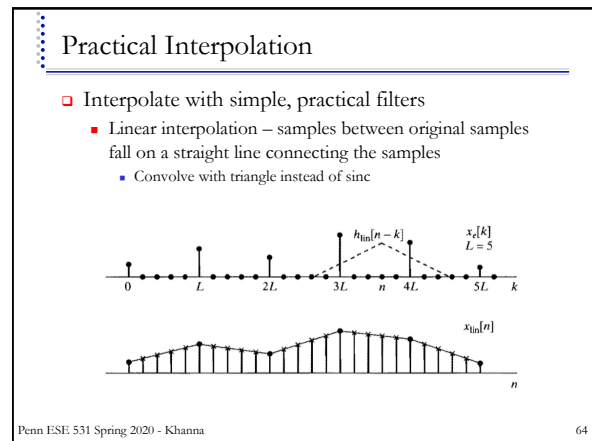
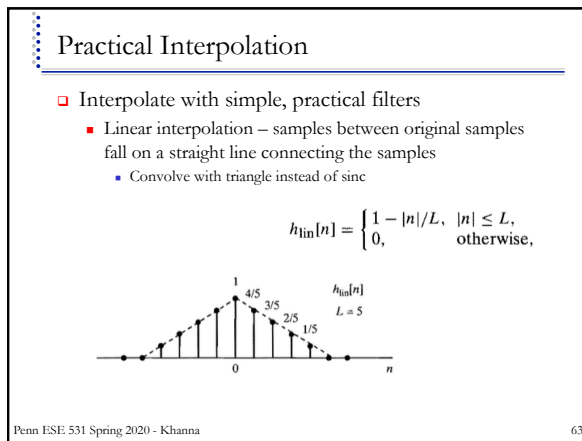
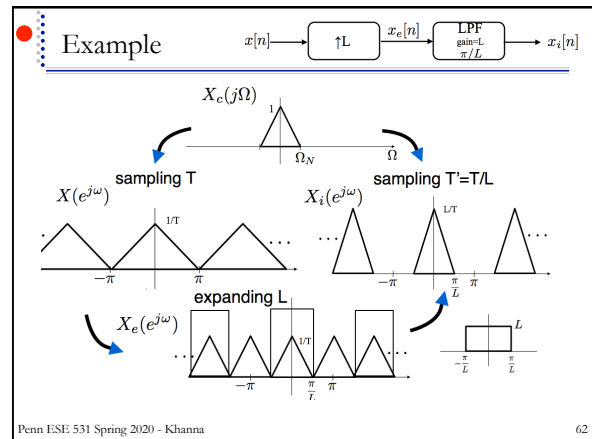
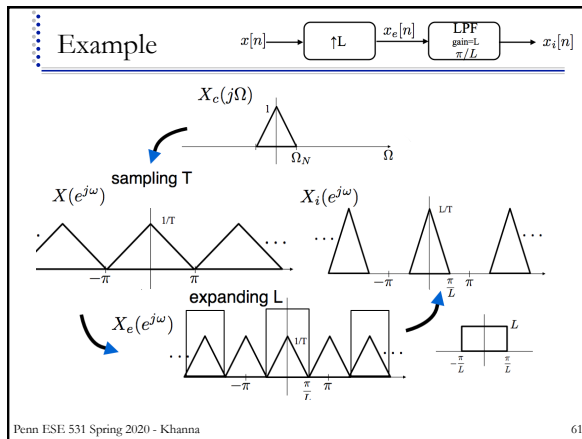
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Example



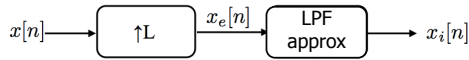
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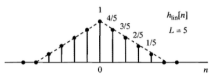


Linear Interpolation -- Frequency Domain

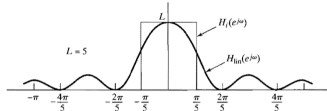
$$x_i[n] = x_e[n] * h_{lin}[n]$$



$$h_{lin}[n] = \begin{cases} 1 - |n|/L, & |n| \leq L, \\ 0, & \text{otherwise,} \end{cases}$$



DTFT $\Rightarrow$

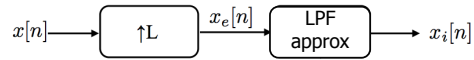


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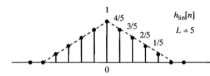
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Linear Interpolation -- Frequency Domain

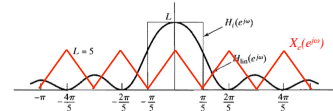
$$x_i[n] = x_e[n] * h_{lin}[n]$$



$$h_{lin}[n] = \begin{cases} 1 - |n|/L, & |n| \leq L, \\ 0, & \text{otherwise,} \end{cases}$$



DTFT $\Rightarrow$

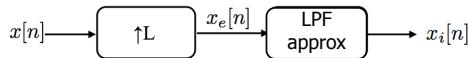


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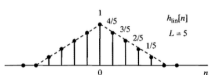
68

Linear Interpolation -- Frequency Domain

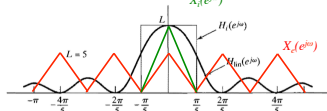
$$x_i[n] = x_e[n] * h_{lin}[n]$$



$$h_{lin}[n] = \begin{cases} 1 - |n|/L, & |n| \leq L, \\ 0, & \text{otherwise,} \end{cases}$$



DTFT $\Rightarrow$

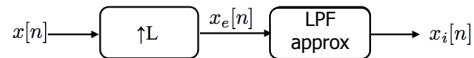


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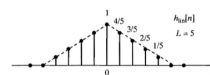
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Linear Interpolation -- Frequency Domain

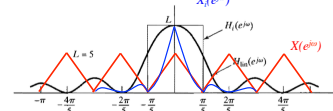
$$x_i[n] = x_e[n] * h_{lin}[n]$$



$$h_{lin}[n] = \begin{cases} 1 - |n|/L, & |n| \leq L, \\ 0, & \text{otherwise,} \end{cases}$$



DTFT $\Rightarrow$

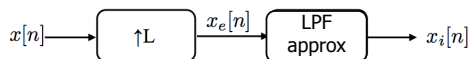


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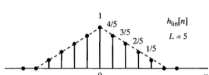
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Linear Interpolation -- Frequency Domain

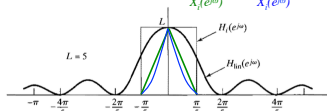
$$x_i[n] = x_e[n] * h_{lin}[n]$$



$$h_{lin}[n] = \begin{cases} 1 - |n|/L, & |n| \leq L, \\ 0, & \text{otherwise,} \end{cases}$$



DTFT $\Rightarrow$



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Big Ideas

- CT processing of DT signals
 - Allows for interpretation of DT systems
- Downsampling
 - Like a C/D converter
- Upsampling
 - Like a D/C converter
- Practical Interpolation
 - Linear interpolation
 - Approximate sinc function with triangle

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Admin

- HW 4 due Sunday