

ESE 531: Digital Signal Processing

Week 4

Lecture 7: February 7, 2021

Sampling



Lecture Outline

- Sampling
 - Frequency Response of Sampled Signal

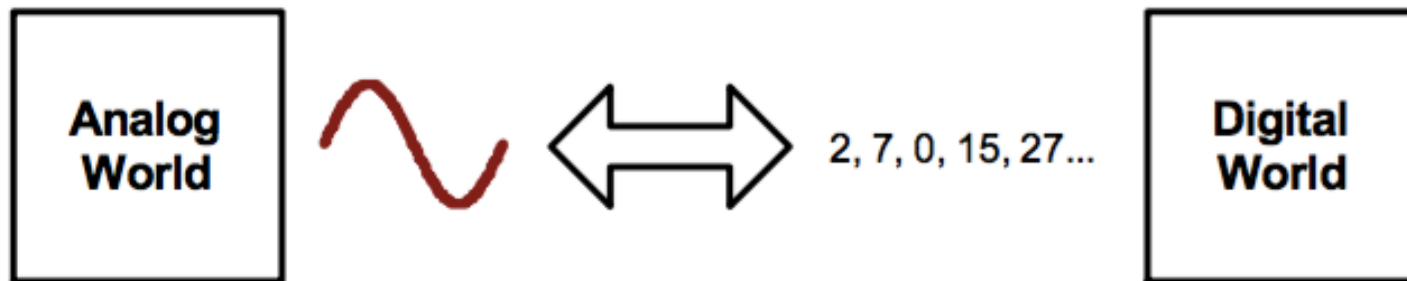


Video Example



❑ <https://www.youtube.com/watch?v=ByTsISFXUoY>

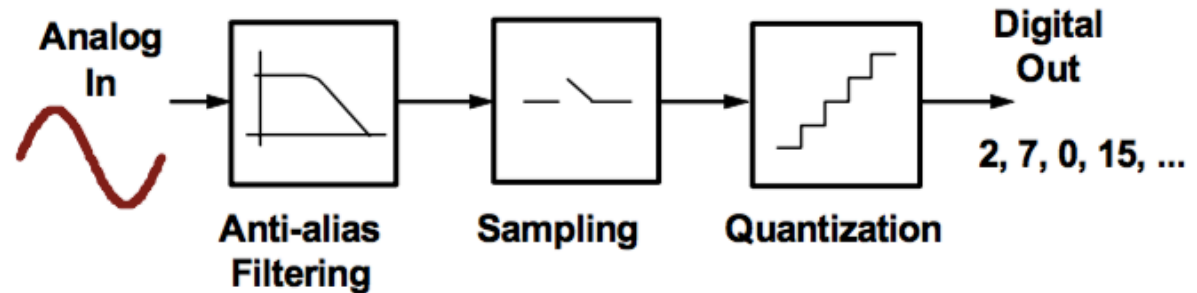
The Data Conversion Problem



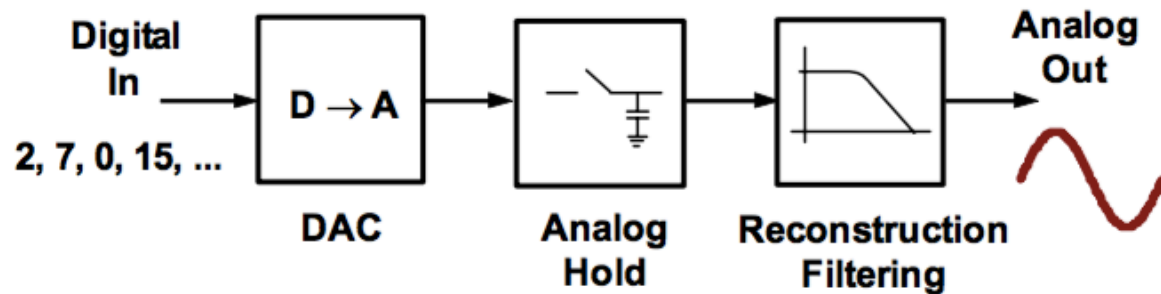
- ❑ Real world signals
 - Continuous time, continuous amplitude
- ❑ Digital abstraction
 - Discrete time, discrete amplitude
- ❑ Two problems
 - How to go discretize in time and amplitude
 - A/D conversion
 - How to "undescretize" in time and amplitude
 - D/A conversion

Overview

A/D Conversion



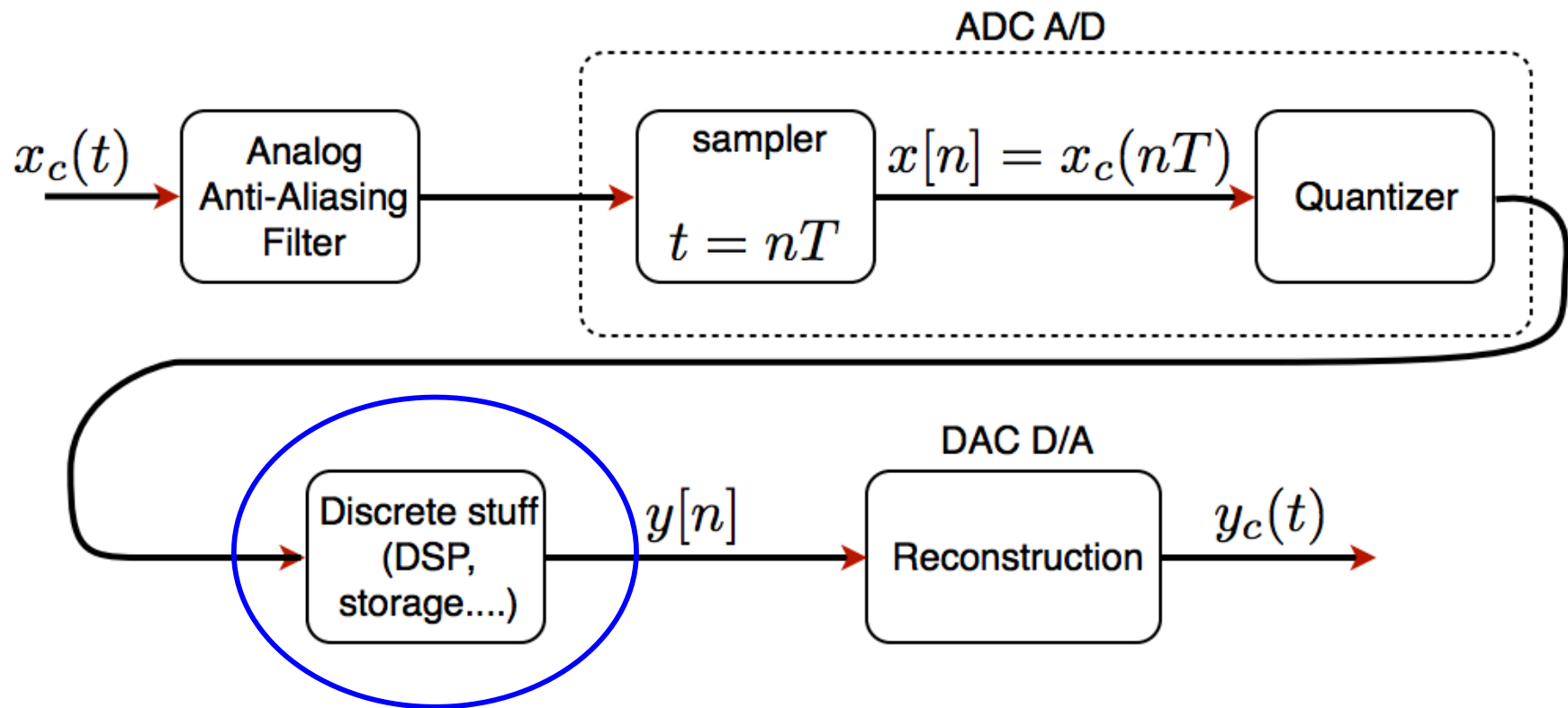
D/A Conversion



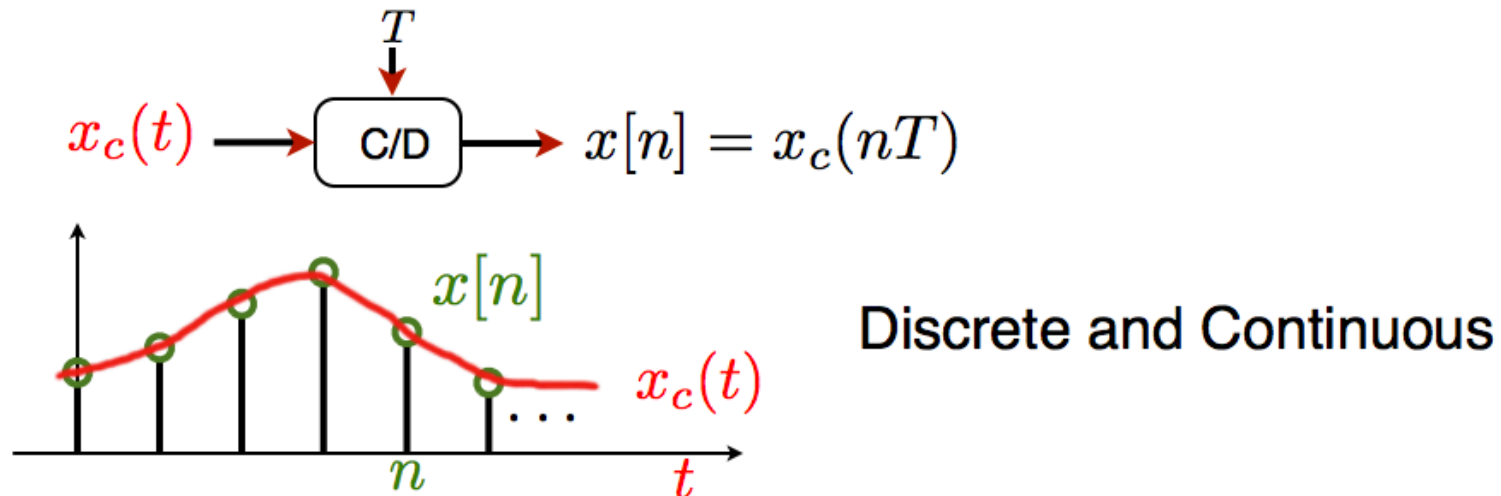
- ❑ We'll first look at these building blocks from a functional, "black box" perspective



DSP System

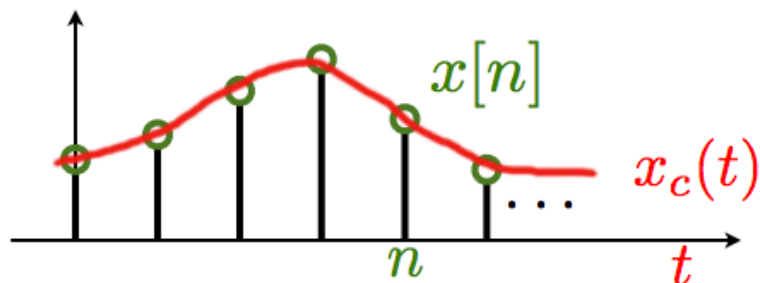
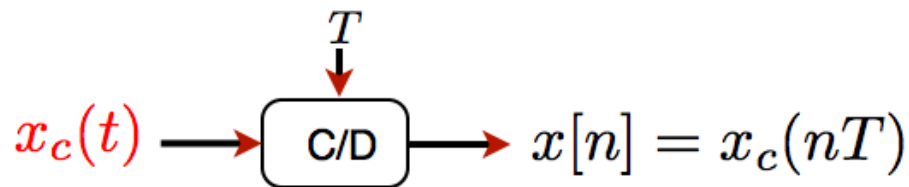


Ideal Sampling Model



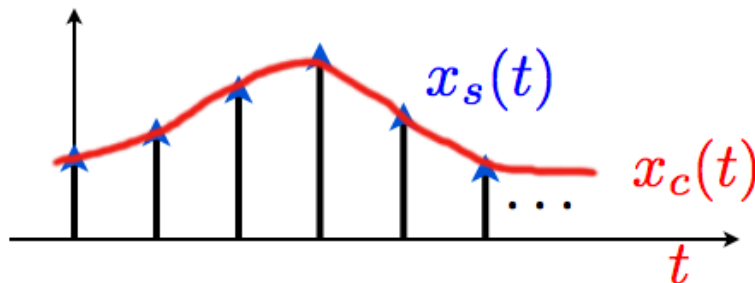
- ❑ Ideal continuous-to-discrete time (C/D) converter
 - T is the sampling period
 - $f_s = 1/T$ is the sampling frequency
 - $\Omega_s = 2\pi/T$

Ideal Sampling Model



Discrete and Continuous

define impulsive sampling:



Continuous

$$x_s(t) = \cdots + x_c(0)\delta(t) + x_c(T)\delta(t - T) + \cdots$$
$$x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$



Ideal Sampling Model

$$x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

- Dirac delta function, $\delta(t)$
 - Infinitely high and thin, area of 1
 - Not physical—for modeling

$$x[n] \leftrightarrow x_s(t) \leftrightarrow x_c(t)$$

- Three signals. How are they related? In time? In frequency?



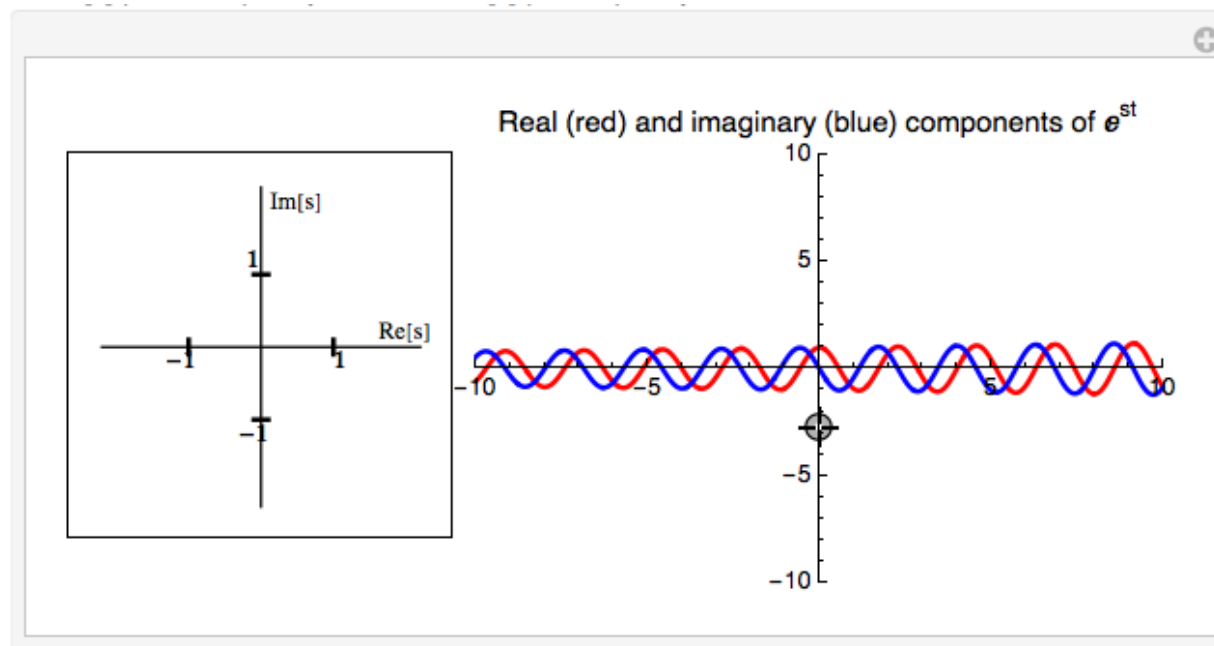
Reminder: Laplace Transform

- ❑ The Laplace transform takes a function of time, t , and transforms it to a function of a complex variable, s .
- ❑ Because the transform is invertible, no information is lost and it is reasonable to think of a function $f(t)$ and its Laplace transform $F(s)$ as two views of the same phenomenon.
- ❑ Each view has its uses and some features of the phenomenon are easier to understand in one view or the other.

S-Plane

□ $s = \sigma + j\Omega$

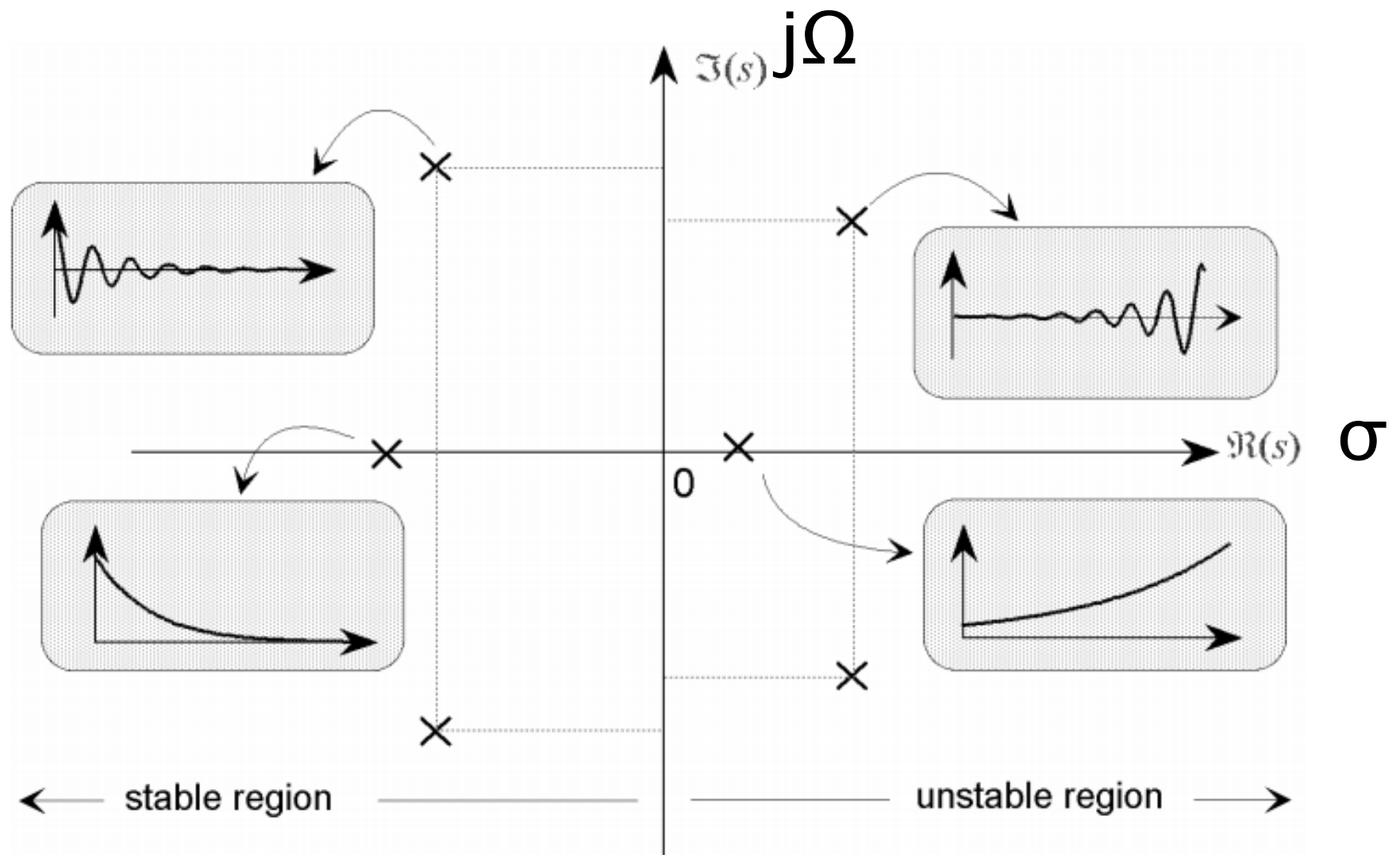
□ Wolfram Demo



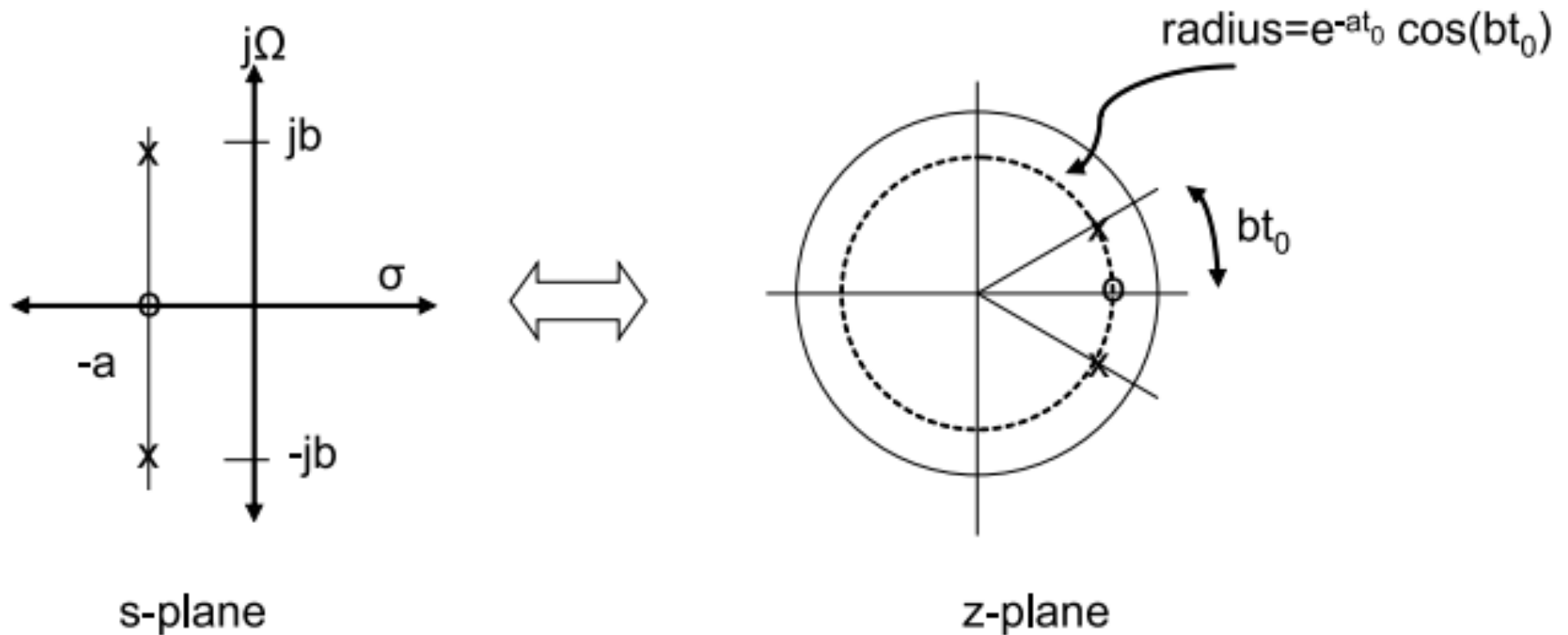
- See piazza for more information on using this demo:
- <https://piazza.com/class/kjloarx1214170?cid=25>



S-plane and stability



S-Plane Mapping to Z-Plane





Frequency Domain Analysis

- How is $x[n]$ related to $x_s(t)$ in frequency domain?

$$x[n] = x_c(nT) \qquad x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

Frequency Domain Analysis

- How is $x[n]$ related to $x_s(t)$ in frequency domain?

$$x[n] = x_c(nT) \qquad x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

$$x_s(t) \quad : \text{C.T.}$$

$$X_s(j\Omega) = \sum_n x_c(nT) e^{-j\Omega nT}$$

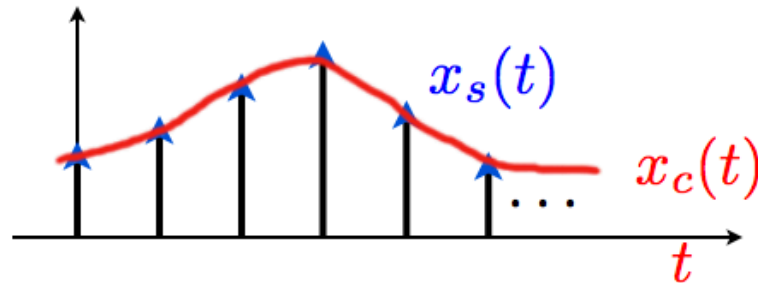
$$x[n] \quad : \text{D.T.}$$

$$X(e^{j\omega}) = \sum_n x[n] e^{-j\omega n} \qquad \omega = \Omega T$$

$$X(e^{j\omega}) = X_s(j\Omega)|_{\Omega=\omega/T}$$

$$X_s(j\Omega) = X(e^{j\omega})|_{\omega=\Omega T}$$

Frequency Domain Analysis



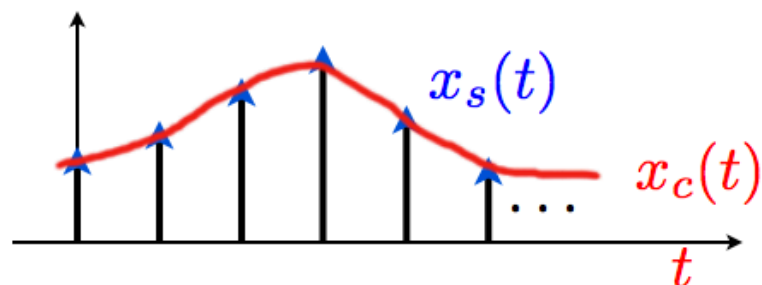
Continuous

$$x_s(t) = \cdots + x_c(0)\delta(t) + x_c(T)\delta(t - T) + \cdots$$

$$x_s(t) = x_c \underbrace{\sum_{n=-\infty}^{\infty} \delta(t - nT)}_{\triangleq s(t)}$$

$$s(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

Frequency Domain Analysis



Continuous

$$x_s(t) = \cdots + x_c(0)\delta(t) + x_c(T)\delta(t - T) + \cdots$$

$$x_s(t) = x_c \underbrace{\sum_{n=-\infty}^{\infty} \delta(t - nT)}_{\triangleq s(t)}$$

$$s(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

$$s(t) \leftrightarrow S(j\Omega)$$

$$S(j\Omega) = \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(\Omega - \frac{2\pi}{T}k)$$

Ω_s



Frequency Domain Analysis

$$x_s(t) = s(t) \cdot x_c(t)$$



Frequency Domain Analysis

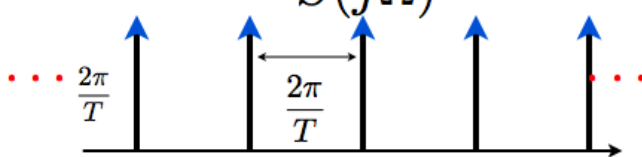
$$x_s(t) = s(t) \cdot x_c(t)$$

$$X_s(j\Omega) = \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega)$$

Frequency Domain Analysis

$$x_s(t) = s(t) \cdot x_c(t)$$

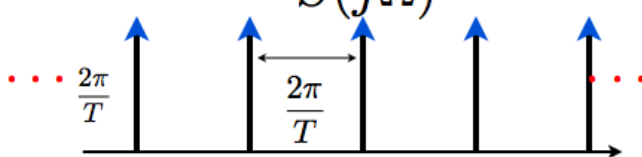
$$X_s(j\Omega) = \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega)$$

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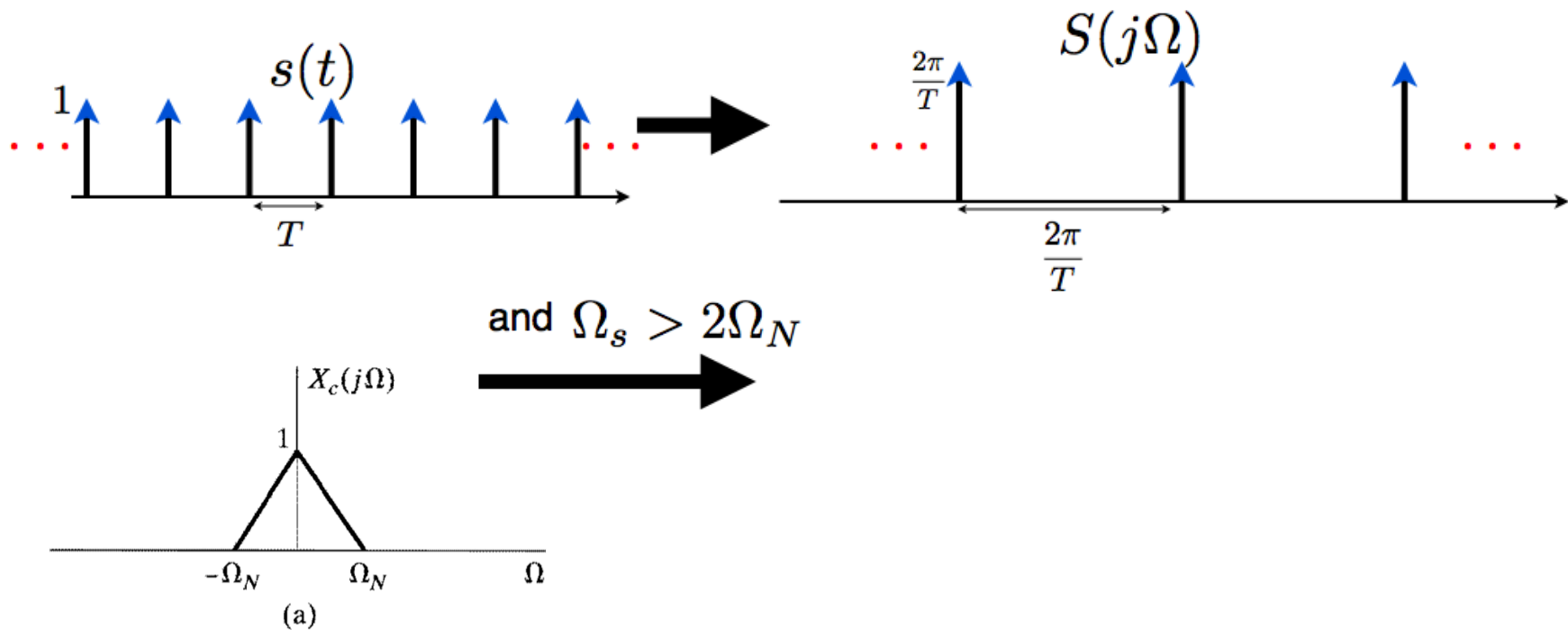
Frequency Domain Analysis

$$x_s(t) = s(t) \cdot x_c(t)$$

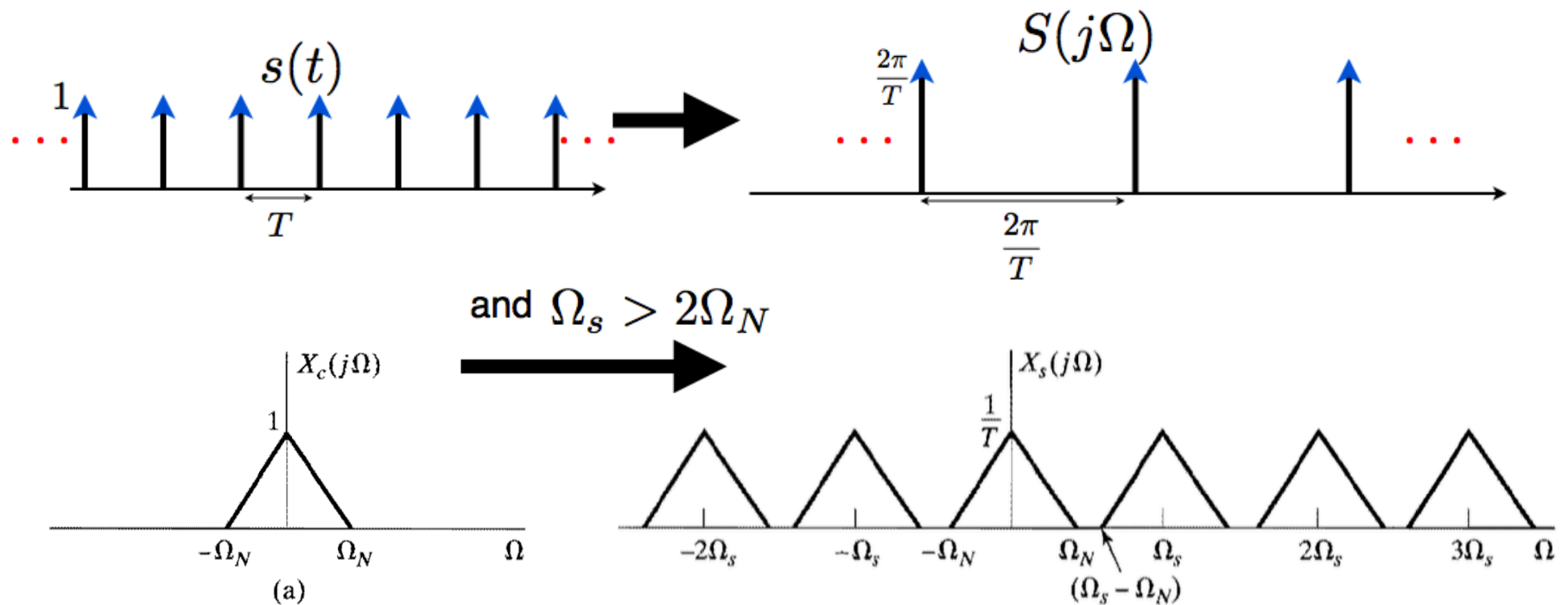
$$\begin{aligned} X_s(j\Omega) &= \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega) \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_s)) \quad | \quad \Omega_s = \frac{2\pi}{T} \end{aligned}$$

$$S(j\Omega) = \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(\Omega - \frac{2\pi}{T}k)$$


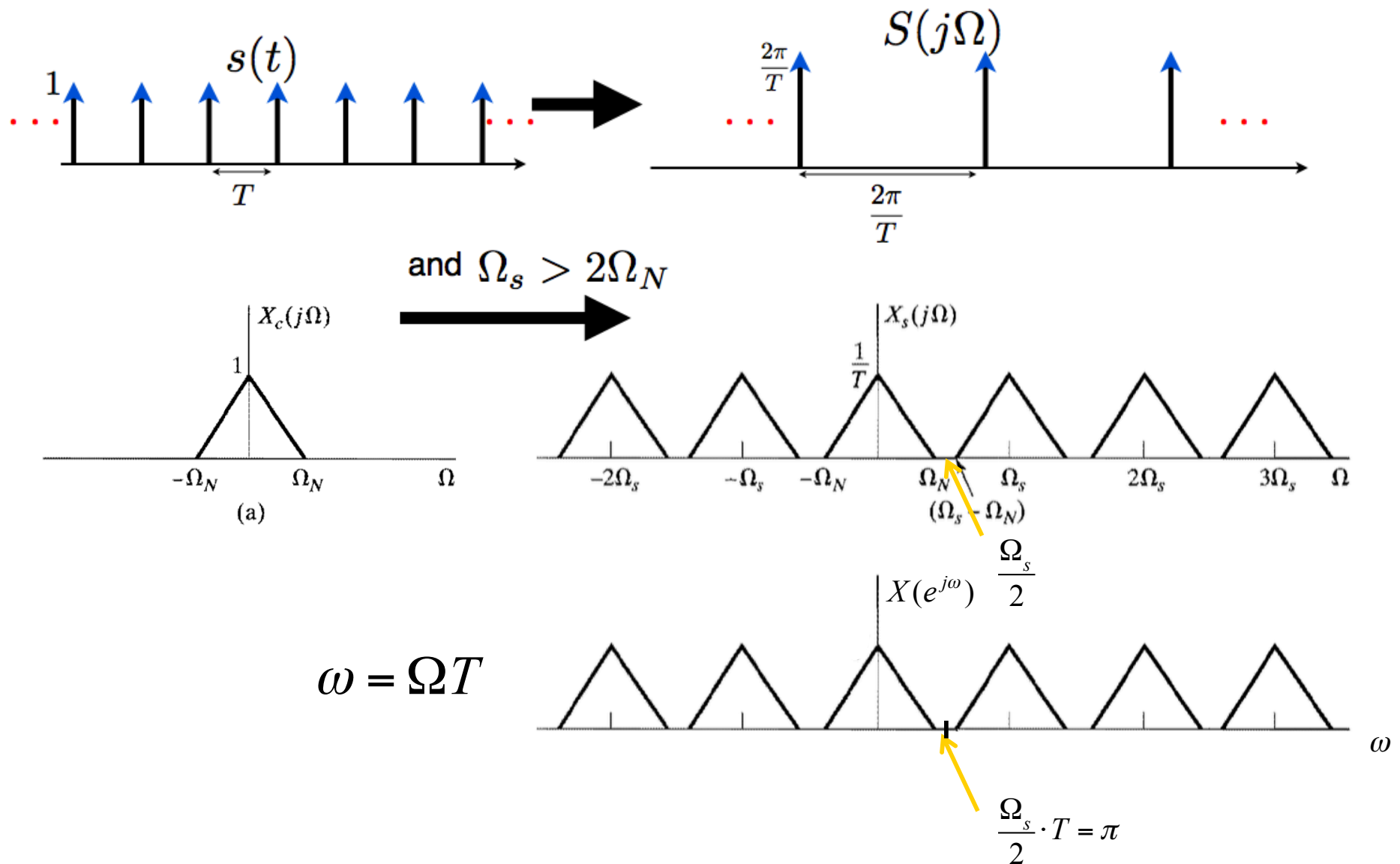
Frequency Domain Analysis



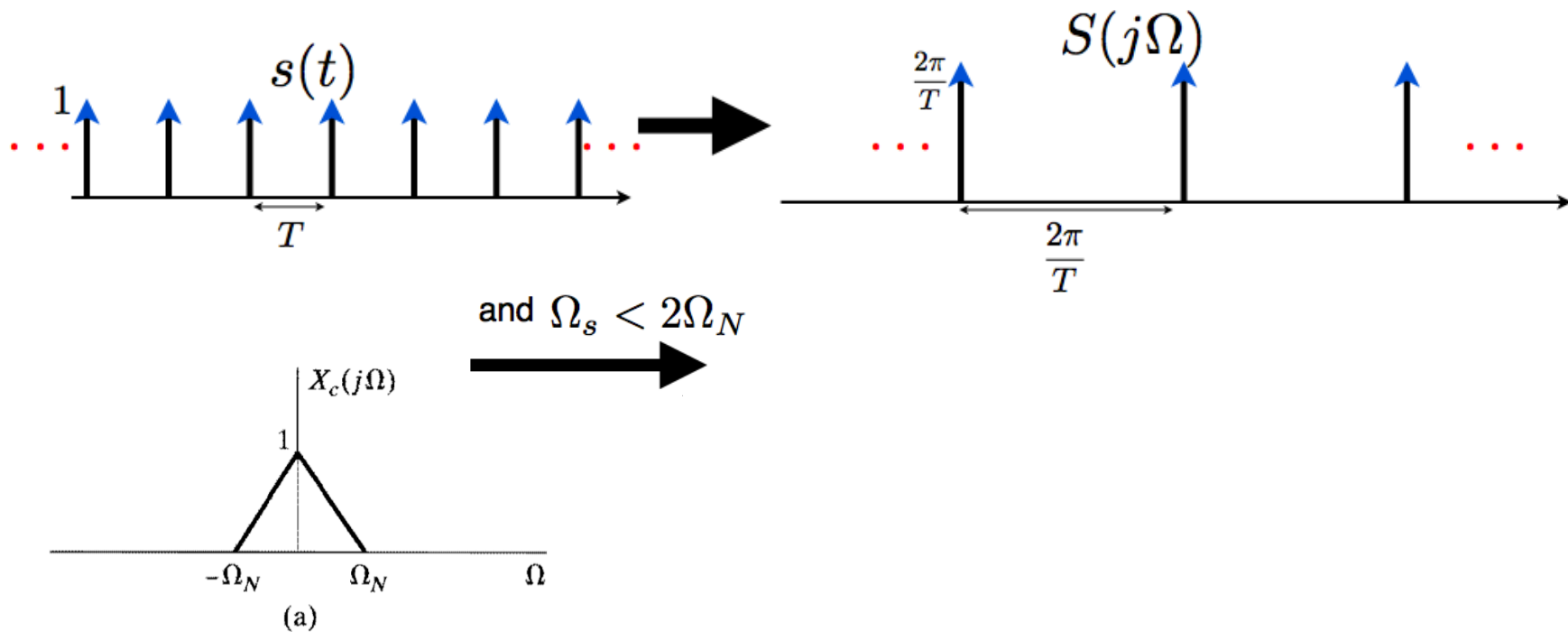
Frequency Domain Analysis



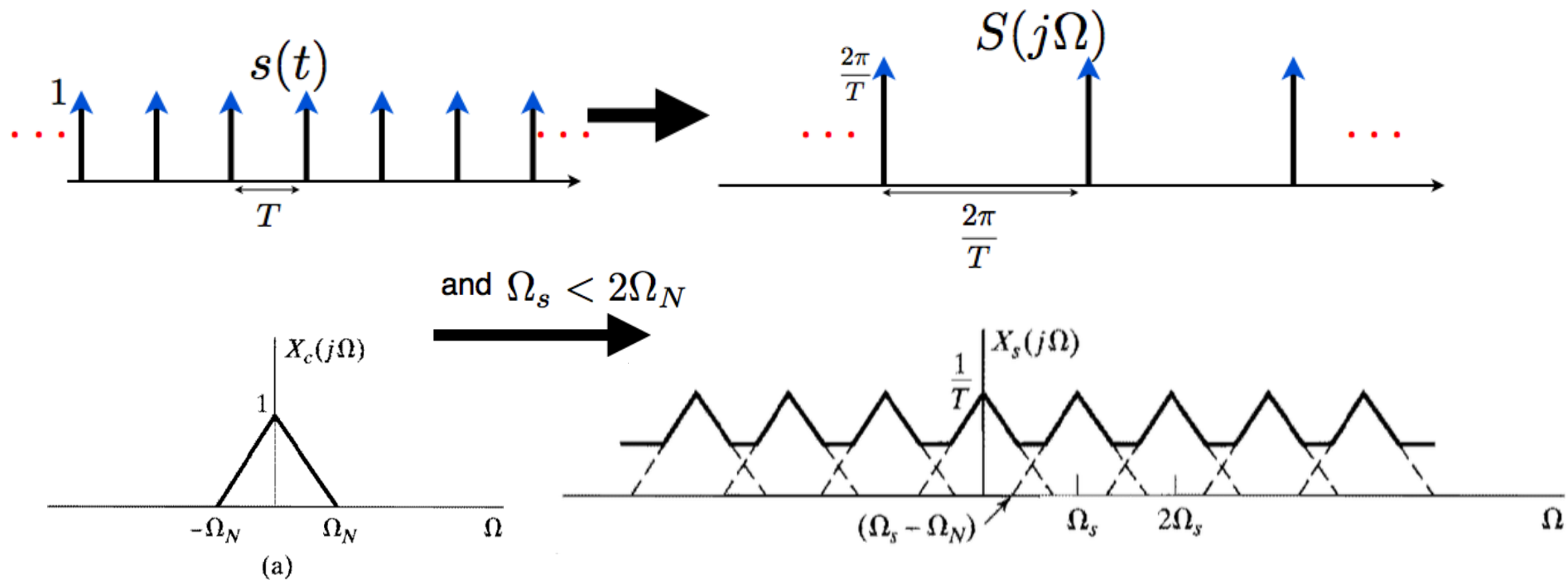
Frequency Domain Analysis



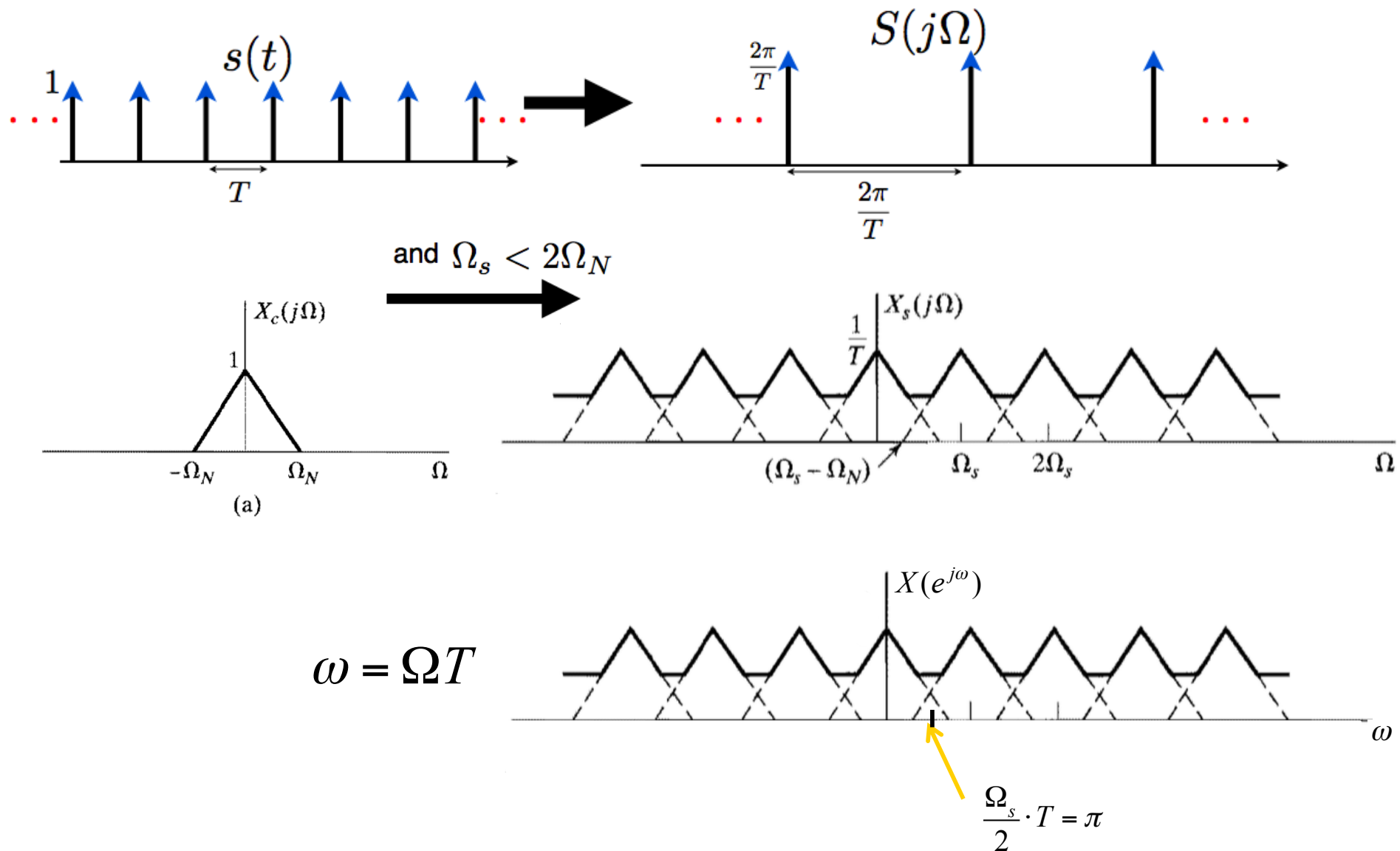
Frequency Domain Analysis w/ Aliasing



Frequency Domain Analysis w/ Aliasing



Frequency Domain Analysis w/ Aliasing





Example: Cosine Input

- Sample the continuous-time signal $x_c(t) = \cos(4000\pi t)$ with sampling period $T = 1/6000$ s ($f_s = 6$ kHz)

$$\begin{aligned} X_s(j\Omega) &= \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega) \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_s)) \quad | \quad \Omega_s = \frac{2\pi}{T} \end{aligned}$$



Example: Cosine Input

- Sample the continuous-time signal $x_c(t) = \cos(16000\pi t)$ with sampling period $T = 1/6000$ ($f_s = 6\text{kHz}$)



Big Ideas

- ❑ Sampling
 - Ideal sampling modeled as impulsive sampling
 - Sample at Nyquist rate for recovery of unique bandlimited signal (i.e. avoid aliasing)
- ❑ Frequency Response of Sampled Signal
 - Sampled signal is period replicated input CT signal



Admin

- ❑ HW 2 due Monday 2/15 at midnight
 - Double check that your submission by viewing it!

- ❑ Recitation videos in Canvas and notes on course webpage

- ❑ Make sure you have access to Matlab
 - Let us know ASAP if you need help!