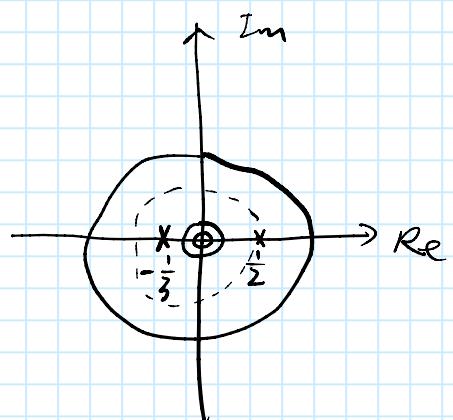


ESE 531 Recitation 13: Final (2020) Review

Q1 (ver. 4) . Q2 (ver. 3)

Q3 , Q4

Q1 (ver 4).



$$H(z) = 6 \text{ at } z=1$$

$$ROC: |z| > \frac{1}{2}$$

(a). $ROC: |z| > \frac{1}{2}$

$$H(z) = \frac{A}{(1 - \frac{1}{2}z^{-1})(1 + \frac{1}{3}z^{-1})} \quad \xrightarrow{\text{constant}}$$

$$H(z)|_{z=1} = H(1) = \frac{A}{(1 - \frac{1}{2})(1 + \frac{1}{3})} = 6$$

$$\Rightarrow A = 4$$

$$H(z) = \frac{4}{(1 - \frac{1}{2}z^{-1})(1 + \frac{1}{3}z^{-1})} \quad \times \frac{4z^2}{(z - \frac{1}{2})(z + \frac{1}{3})}$$

(b). $h[n] = z^{-1} \{ H(z) \}$

$$h[n] = \frac{12}{5} \quad \frac{d}{5}$$

$$H(z) = \frac{\frac{12}{5}}{1 - \frac{1}{2}z^{-1}} + \frac{\frac{8}{5}}{1 + \frac{1}{3}z^{-1}}$$

↓
inverse z-transform, ROC $|z| > \frac{1}{2}$

$$h[n] = \frac{12}{5} \times \left(\frac{1}{2}\right)^n u[n] + \frac{8}{5} \left(-\frac{1}{3}\right)^n u[n]$$

(c). Yes.

(d). First find out $x[n]$.

$$x[n] = x[nT] \quad T = \frac{1}{f}$$

t
 $\omega_s = 2\pi f$

$$\Rightarrow t = \frac{2\pi}{\omega_s} \cdot n = \frac{2\pi}{2\pi \times 40} \times n$$

$$= \frac{n}{40}$$

$$\Rightarrow x[n] = 50 + 30 \cos(40\pi \times \frac{n}{40})$$

$$= 50 + 30 \cos(\pi n)$$

Next:

$$y[n] = \underline{x[n]} * \underline{h[n]}$$

$$Y(e^{j\omega}) = \underline{X(e^{j\omega})} \underline{H(e^{j\omega})}$$

$$X(e^{j\omega}) = \text{DTFT} \{ x[n] \}$$

$$Y(e^{j\omega}) = \sum_{k=-\infty}^{\infty} [z \delta(\omega - z + 2\pi k) + z \delta(\omega + z + 2\pi k)]$$

↳ from cos

$$+ 50 \sum_{k=-\infty}^{\infty} 2z \delta(\omega + 2\pi k)$$

$$\begin{aligned} Y(e^{j\omega}) &= \frac{\downarrow}{\sqrt{4}} \times H(e^{j\omega}) \\ &= \frac{8()}{\sqrt{4}} \times \frac{(1-i e^{-j\omega})(1+\frac{1}{3} e^{-j\omega})}{(1-i e^{-j\omega})(1+\frac{1}{3} e^{-j\omega})} \end{aligned}$$

$Y(e^{j\omega})$ is non-zero only at ω' , where

ω' is in $\delta(\omega')$

$$Y(e^{j\omega}) = \left| \begin{array}{c} \text{from ① } \omega = \pm z \\ \frac{4}{(1+\frac{1}{2})(1-\frac{1}{3})} \end{array} \right|$$

$$\left| \begin{array}{c} \frac{9}{2 \times \frac{2}{3}} \times ① \\ + \frac{4}{\frac{1}{2} \times \frac{4}{3}} \times ② \end{array} \right|$$

$$\left| \begin{array}{c} \text{from ② } \omega = 0 \\ \frac{4}{(1-\frac{1}{2})(1+\frac{1}{3})} \end{array} \right|$$

$$\begin{aligned} &= 4 \times 30 \sum_{k=-\infty}^{\infty} [z \delta(\omega - z + 2\pi k) + z \delta(\omega + z + 2\pi k)] \\ &+ 1 \times 50 \sum_{k=-\infty}^{\infty} 2z \delta(\omega + 2\pi k) \end{aligned}$$

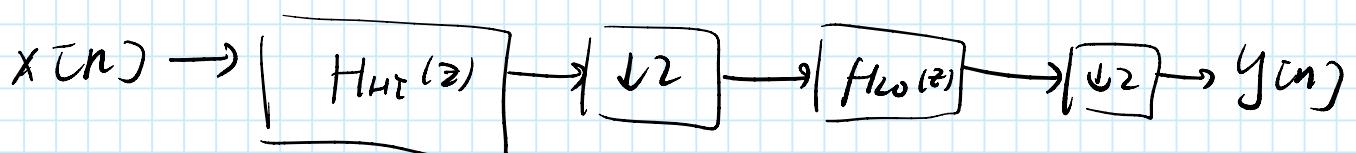
$$+ 6 \times 50 \sum_{k=-\infty}^{\infty} 2z \delta(\omega + 2\pi k)$$

$$y(n) = \text{IDFT} \left\{ Y(e^{j\omega}) \right\}$$

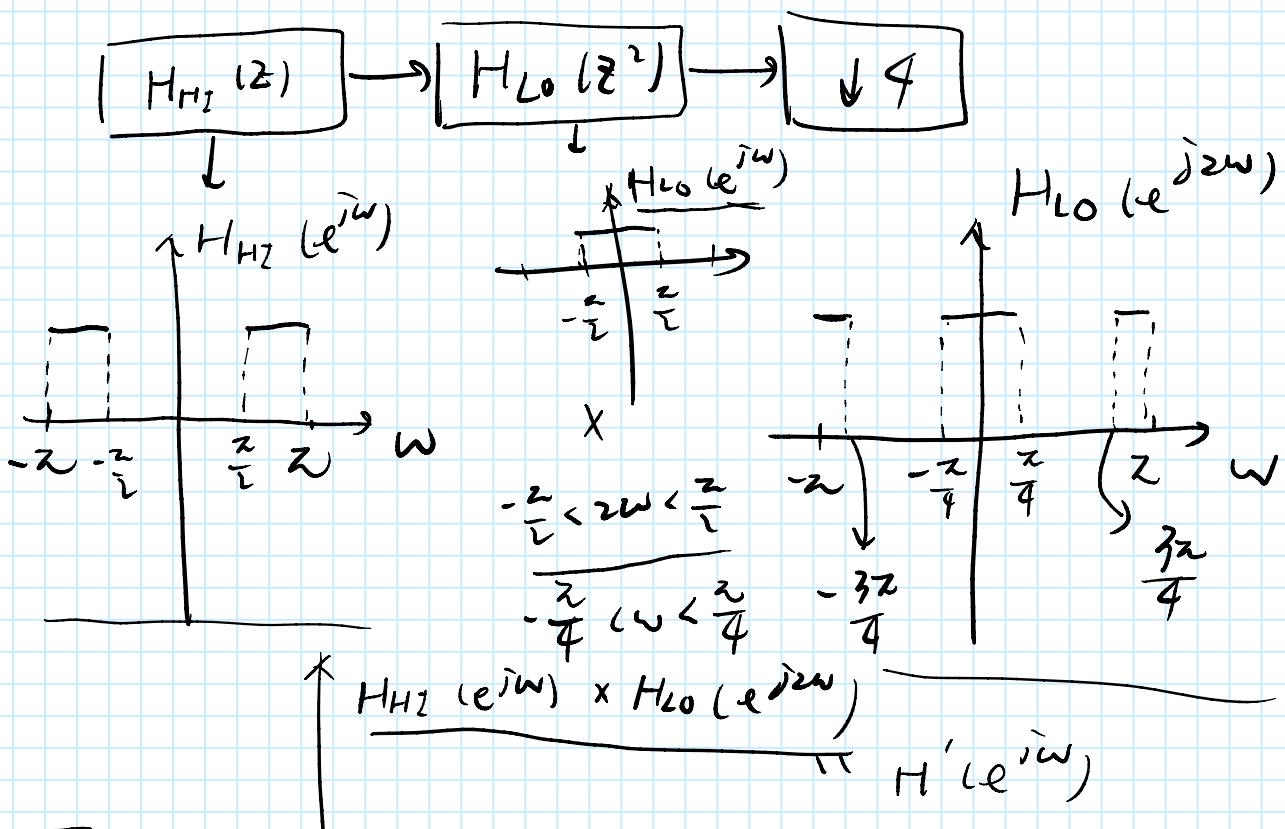
$$= 120 \cos(2n) + 300$$

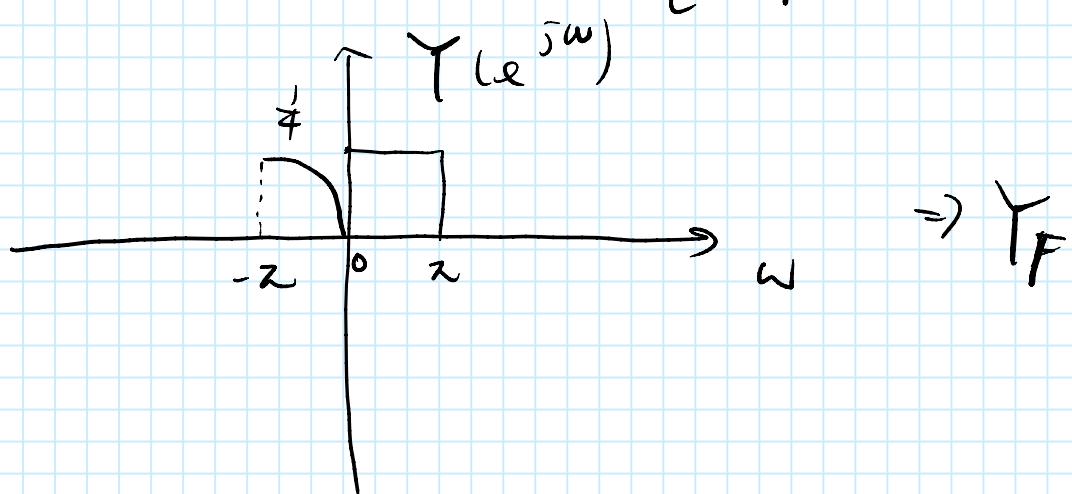
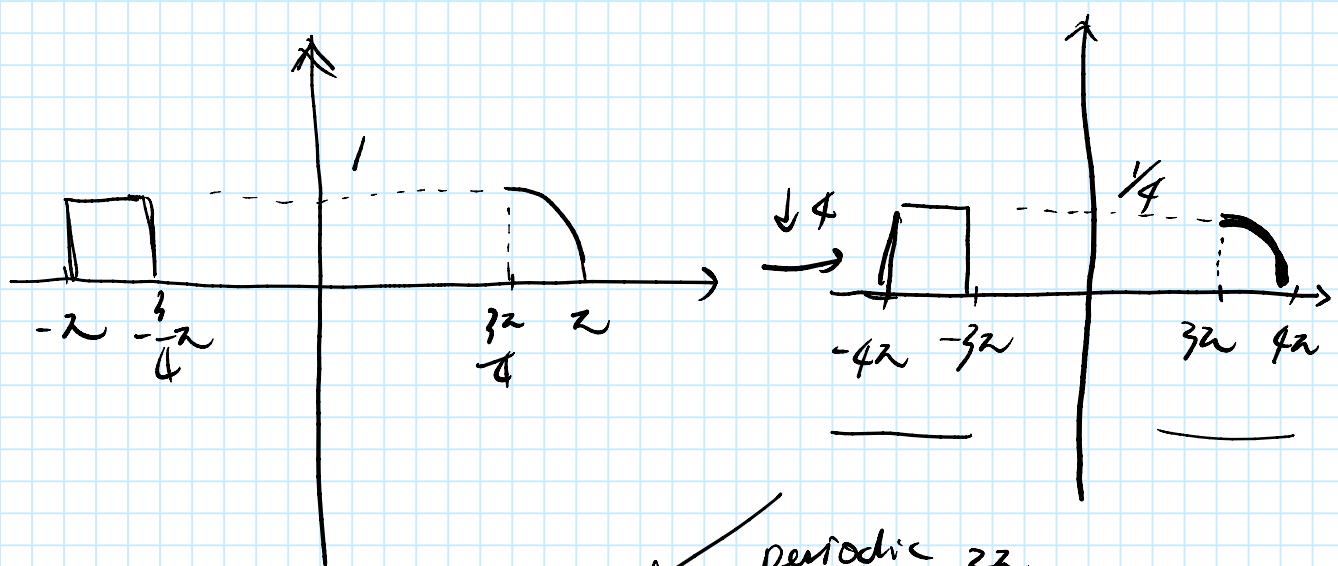
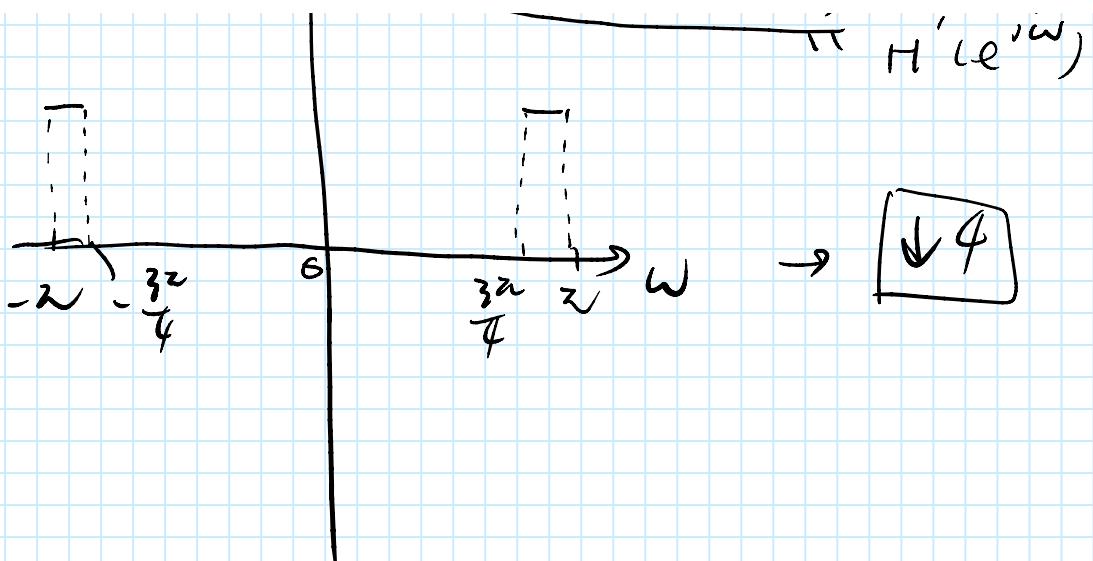
Q2 (ver. 3)

(a). Need to determine $Y(e^{j\omega})$

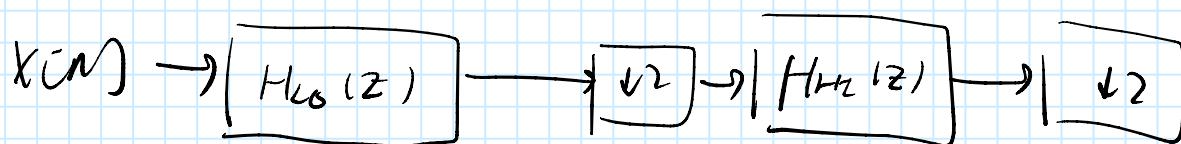


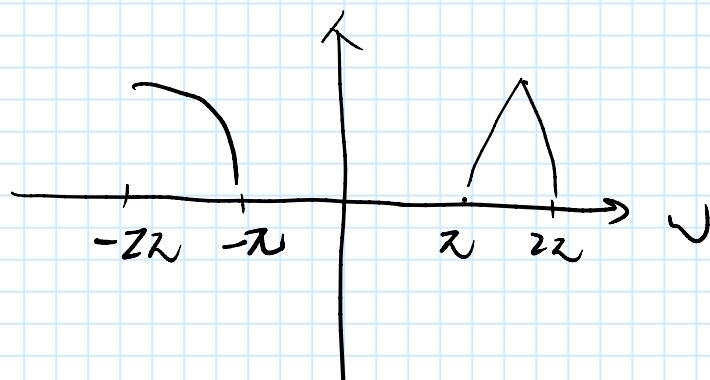
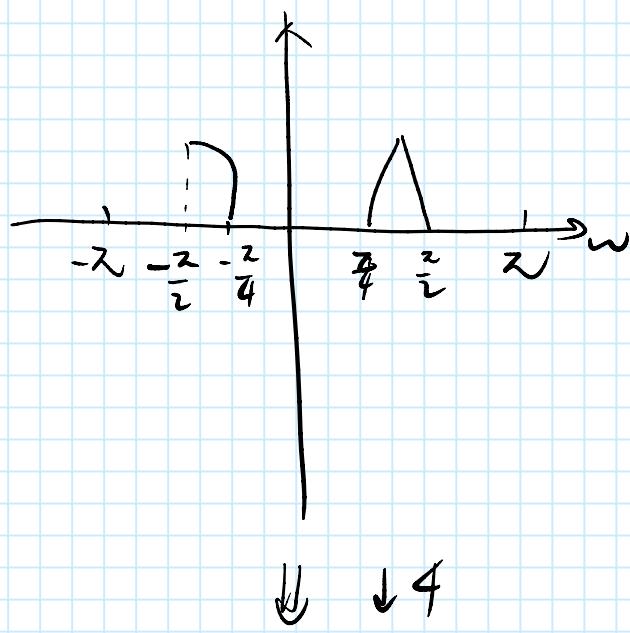
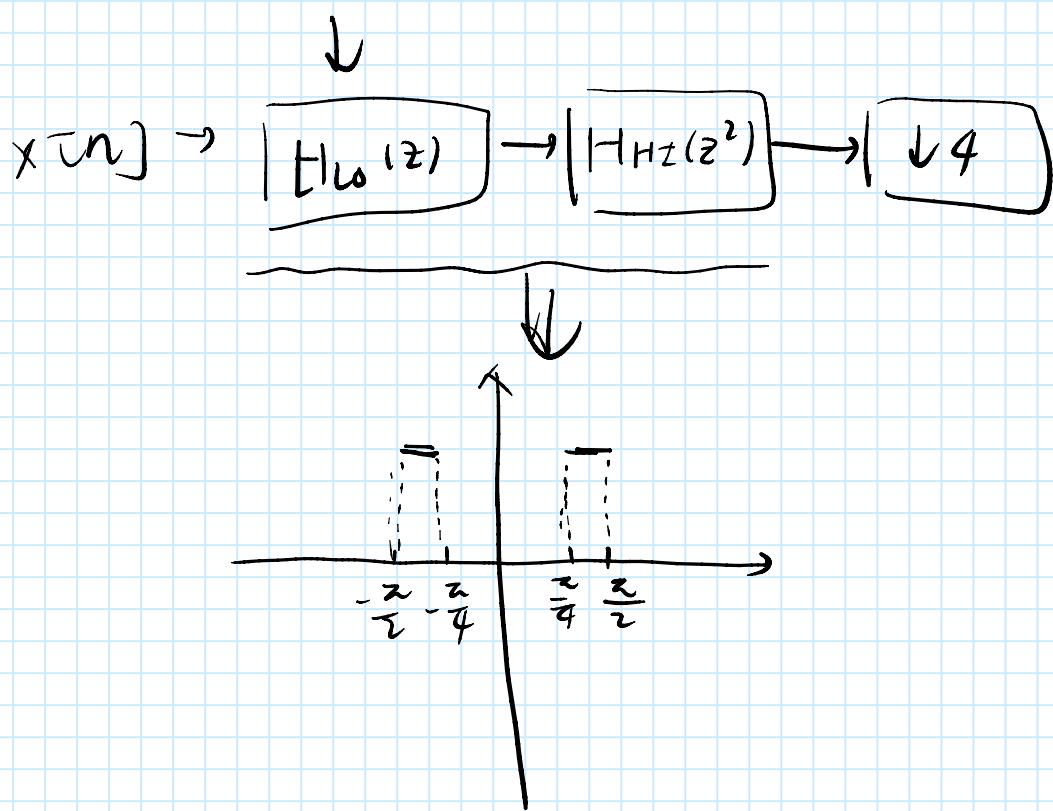
↓ from interchangeable relation

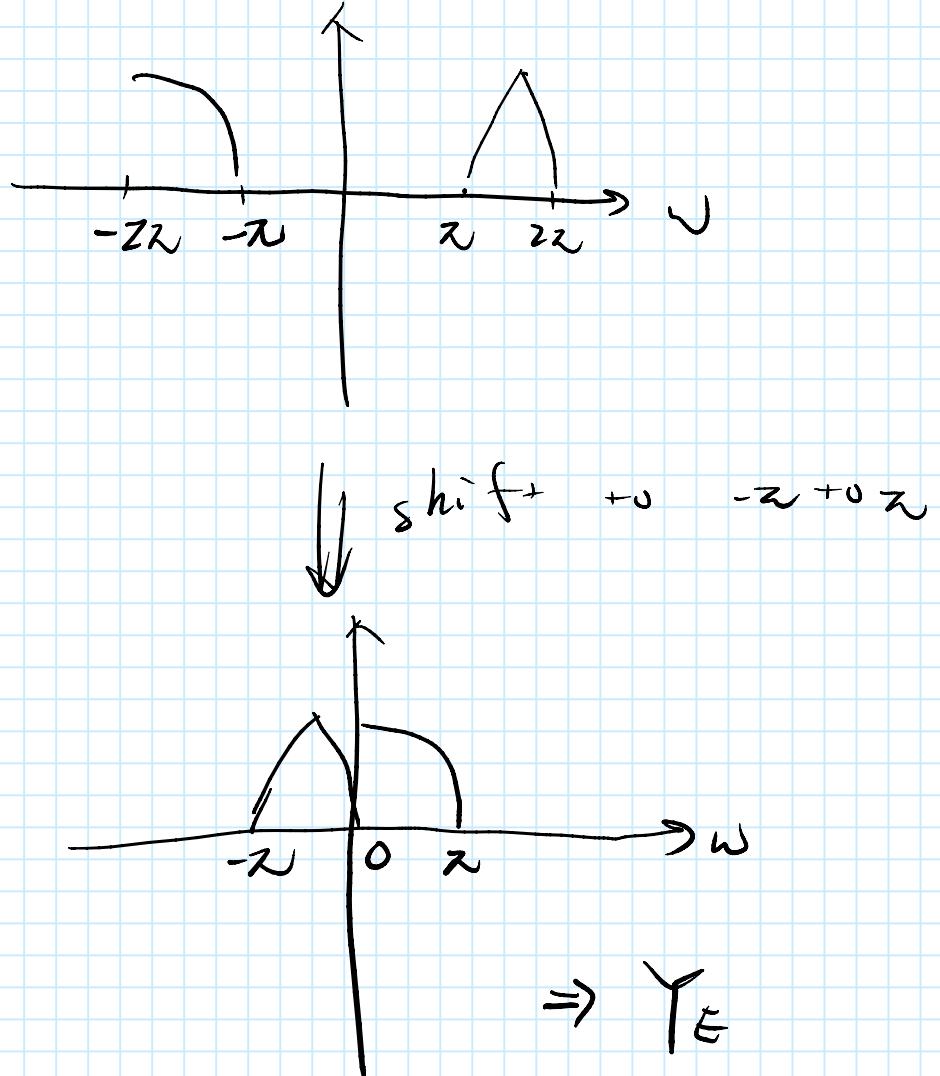




(b).
Gauss







Q3.

8-DFT of $y(n)$

what we know : $\{Y[0], Y[2], Y[4], Y[6]\}$
 $\{y[0], y[1], y[2], y[3]\}$

(a). Want to find out $Y[1], Y[3], Y[5], Y[7]$

Do butterfly operation first

then do two 1-point DFT

+ then do two 4-point DFT

Why? Why this kind structure.

Naturally, from decimation-in-time FFT

\Rightarrow divide input into even & odd groups.

output $\Rightarrow Y[0], Y[1], \dots, Y[7]$

But for this problem, we have only
even samples from Y

\Rightarrow Need a different structure to
compute 8-point DFT instead of
the structure shown in Fig 9.4 P726

in textbook

\Rightarrow Need bit-reversed structure in Fig 9.15
P736

Now, from bit-reversed structure.

can compute $g[0] \dots g[3]$ from

4-point IDFT of $Y[0], Y[2], Y[4], Y[6]$,

which is $g[n] = \text{IDFT}\{Y[\text{even}]\}$

ii. And from $g[n]$ and $y[0], y[1], y[2], y[3]$

\hookrightarrow can compute $(y[0] + jy[1]) / \sqrt{2}, (y[2] + jy[3]) / \sqrt{2}$

→ can compute $y[4], y[5], y[6], y[7]$

iii. And From $y[0] \dots y[7]$, can compute $T[\text{odd}]$

$$\begin{aligned} i. \quad g[n] &= \text{IFT} \{ T[\text{even}] \} \\ &= \frac{1}{4} \sum_{k=0}^3 G[k] W_4^{-kn} \end{aligned}$$

$$\text{where } \begin{cases} G[0] = T[0] \\ G[1] = T[2] \\ G[2] = T[4] \\ G[3] = T[6] \end{cases}$$

$$\begin{aligned} \Rightarrow g[0] &= \frac{1}{4} \left\{ G[0] + G[1]W_4^0 + G[2]W_4^0 + G[3]W_4^0 \right\} \\ &= \frac{1}{4} T[0] \{ 1 + 1 + 1 + 1 \} \\ &= T[0] = 2 \end{aligned}$$

$$g[1] = \frac{1}{4} \left[G[0] + G[1]W_4^{-1} + G[2]W_4^{-2} + G[3]W_4^{-3} \right]$$

$$= \frac{1}{4} T[0] \left\{ 1 + w_4^{-1} + \cancel{w_4^{-2}} + \cancel{w_4^{-3}} \right\}$$

since $w_N = e^{-j \frac{2\pi}{N}}$, is periodic

$$= \frac{1}{4} T[0] \left\{ 1 + w_4^{-1} + (-1) + -w_4^{-1} \right\}$$

$$= 0$$

$$g[2] = \frac{1}{4} T[0] \left\{ 1 + \cancel{w_4^{-2}} + \cancel{w_4^{-4}} + \cancel{w_4^{-6}} \right\}$$

$$= 0$$

$$g[3] = \frac{1}{4} T[0] \left\{ 1 + \cancel{w_4^{-3}} + \cancel{w_4^{-6}} + \cancel{w_4^{-9}} \right\}$$

$$= 0$$

ii. From $g[n], y[0], \dots, y[3]$

$$\begin{aligned} g[0] &= y[0] + y[4] \\ g[1] &= y[1] + y[5] \\ g[2] &= y[2] + y[6] \\ g[3] &= y[3] + y[7] \end{aligned} \Rightarrow \begin{cases} y[4] = 1 \\ y[5] = 0 \\ y[6] = 0 \\ y[7] = 0 \end{cases}$$

$$(g[3]) = \boxed{y[3]} + y[7]$$

iii. From $y[0] \dots y[7]$

can compute $T[\text{odd}]$

⋮

$$T[\text{odd}] = 0.$$

(b). (b) is part (a).

Express step iii. iii. iii. in (a) in a diagram

+ subtraction

7 multiply + on

Q4.

Look for the positions of poles
and zeroes.

usually: $\boxed{\text{pole}} \Rightarrow$ passband
 $\boxed{\text{zero}} \Rightarrow$ stopband