

ESE 5310: Digital Signal Processing

Lecture 24: April 18, 2023

Wavelet Transform



Discrete Time-Dependent Fourier Transform

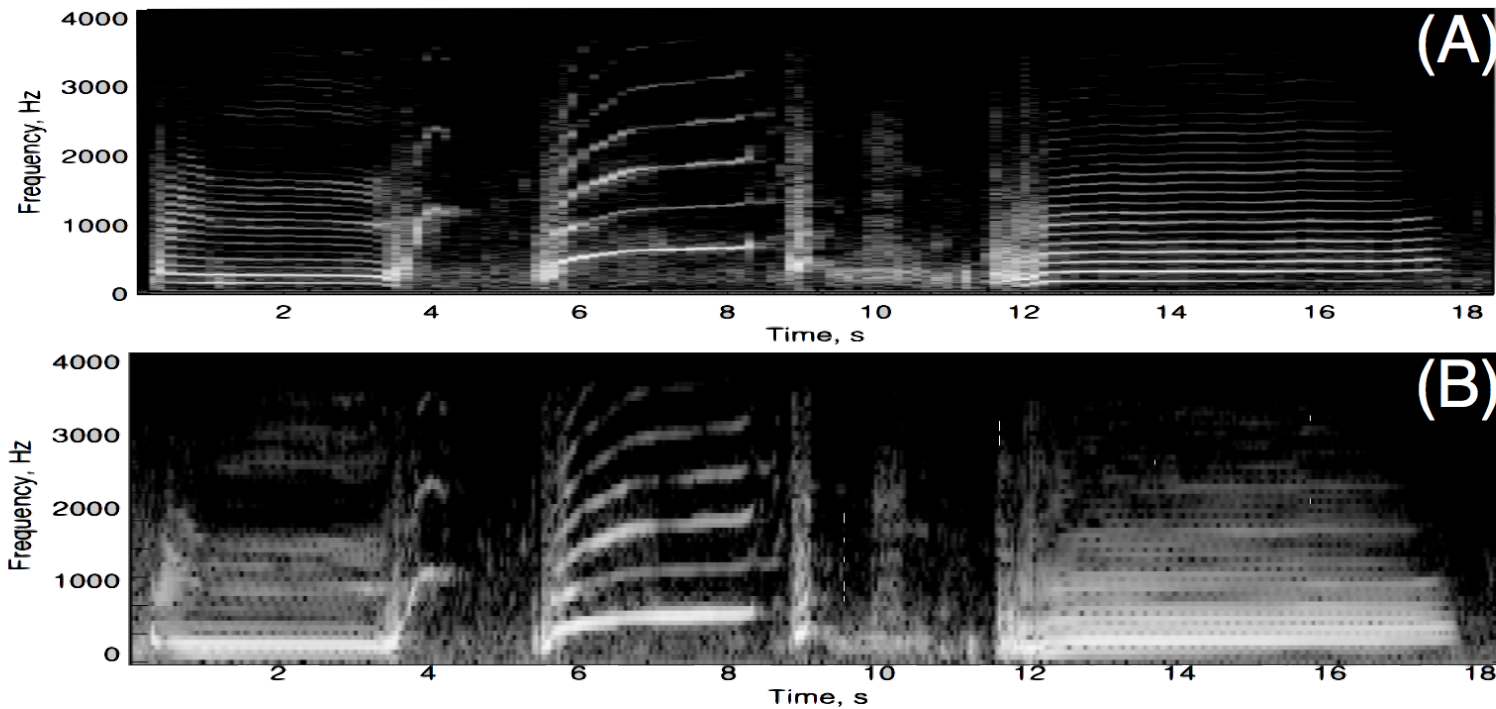
$$X[n, \lambda) = \sum_{m=-\infty}^{\infty} x[n+m]w[m]e^{-j\lambda m}$$

$$X[rR, k] = X[rR, 2\pi k / N) = \sum_{m=0}^{L-1} x[rR+m]w[m]e^{-j(2\pi/N)km}$$

$$X_r[k] = \sum_{m=0}^{L-1} x[rR+m]w[m]e^{-j(2\pi/N)km}$$



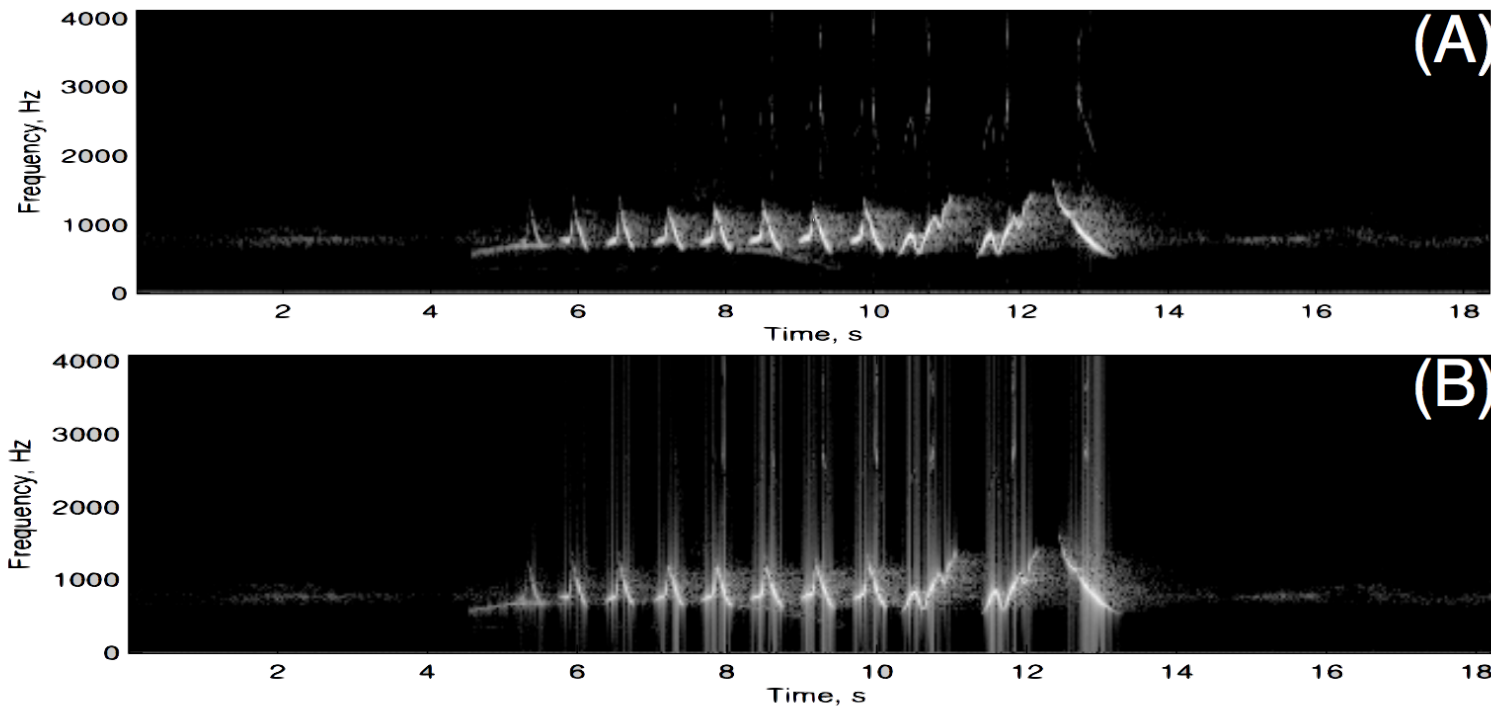
Spectrogram



- What is the difference between the spectrograms?
 - a) Window size $B < A$
 - b) Window size $B > A$
 - c) Window type is different



Spectrogram

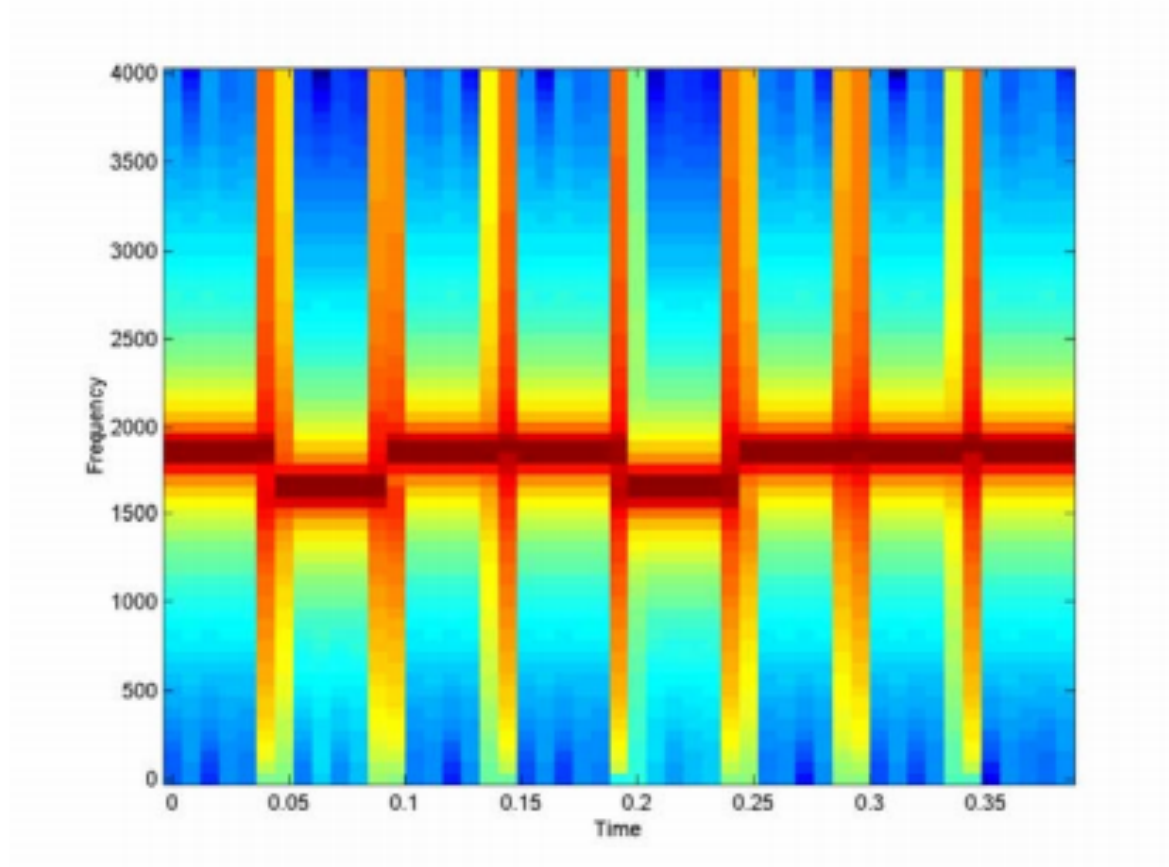


- What is the difference between the spectrograms?
 - a) Window size $B < A$
 - b) Window size $B > A$
 - c) Window type is different



Application – Frequency Shift Keying

- ❑ FSK Communications
 - Spectrogram transmitting 'H' (ASCII H = 01001000)





STFT Reconstruction

- ❑ If $R \leq L \leq N$, then we can recover $x[n]$ block-by-block from $X_r[k]$
- ❑ For non-overlapping windows, $R=L$

$$x_r[m] = \frac{1}{N} \sum_{k=0}^{N-1} X_r[k] e^{j2\pi km/N}$$



STFT Reconstruction

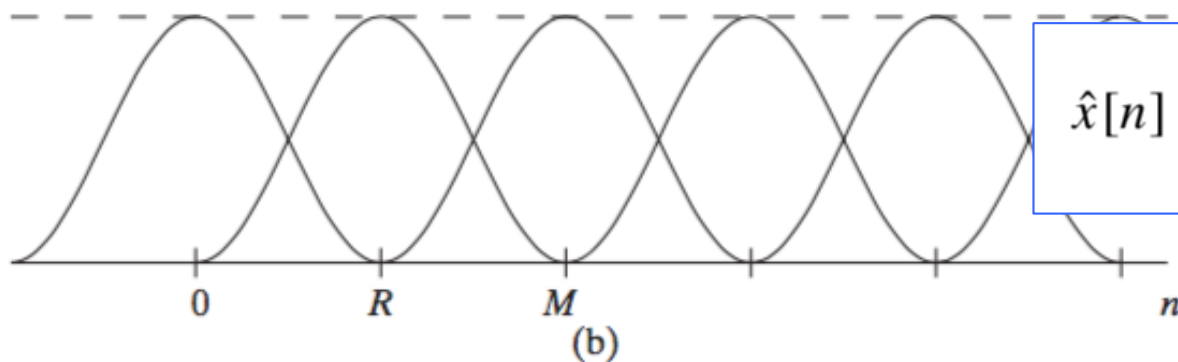
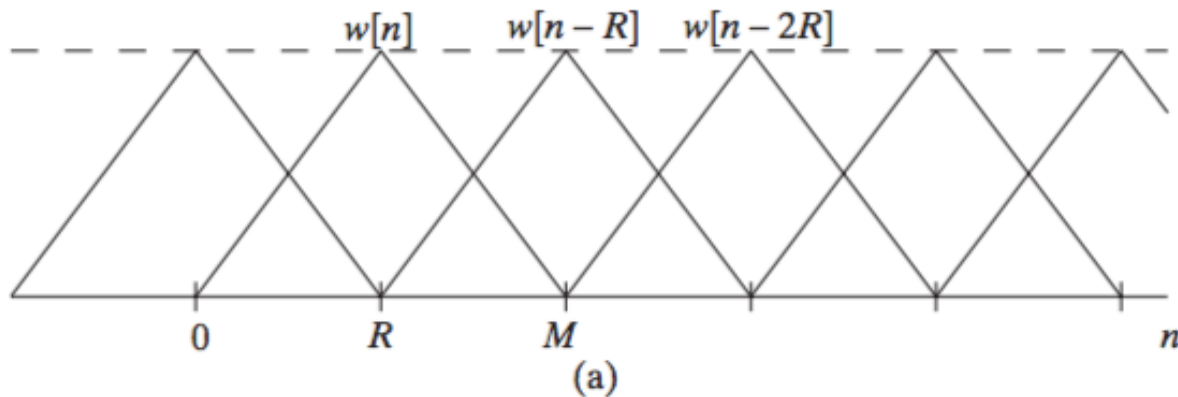
- If $R \leq L \leq N$, then we can recover $x[n]$ block-by-block from $X_r[k]$
- For non-overlapping windows, $R=L$

$$x_r[m] = \frac{1}{N} \sum_{k=0}^{N-1} X_r[k] e^{j2\pi km/N}$$

$$x[n] = \frac{x_r[n - rR]}{w[n - rR]} \quad \forall \quad rR \leq n \leq (r+1)R - 1$$

SFTF Reconstruction with overlap

- ❑ Practically make $R < L < N$
- ❑ If we choose R , L , and N appropriately with window, the overlap-add will negate the window effects



$$\hat{x}[n] = \sum_{r=-\infty}^{\infty} x_r[n - rR].$$

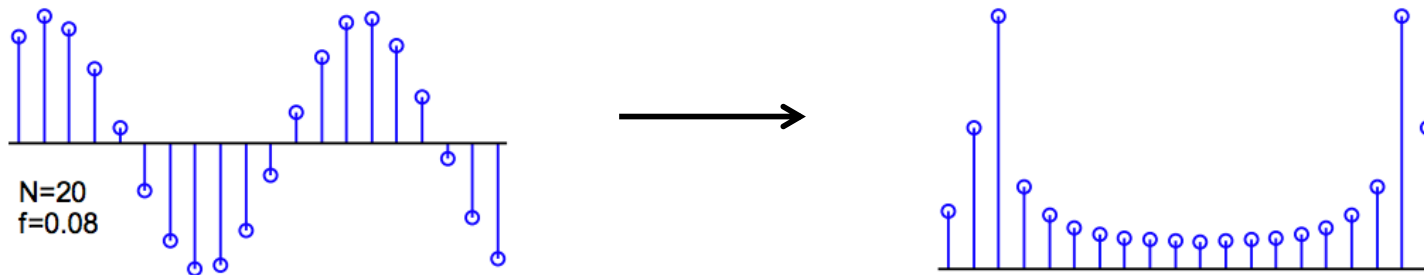


Discrete Cosine Transform

- ❑ Similar to the discrete Fourier transform (DFT), but using only real numbers
- ❑ Widely used in lossy compression applications (eg. Mp3, JPEG)
- ❑ Why use it?

DFT Problems

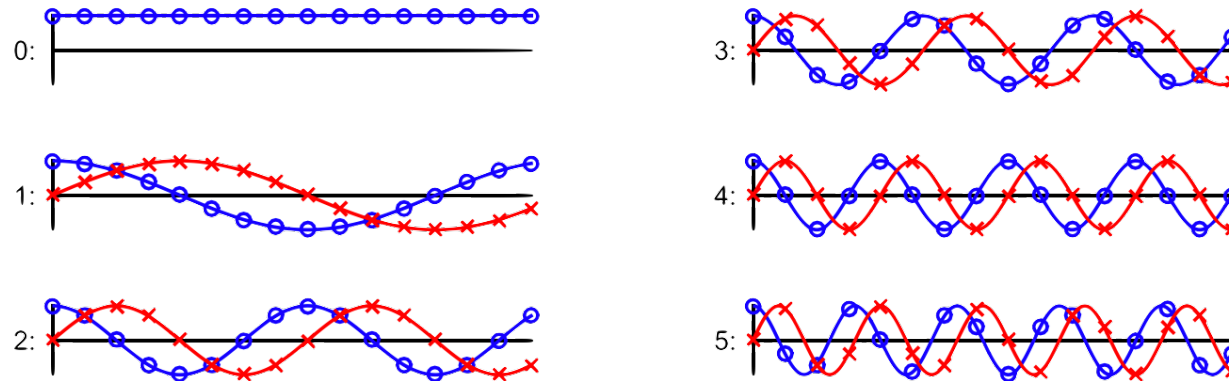
- ❑ For processing 1-D or 2-D signals (especially coding), a common method is to divide the signal into “frames” and then apply an invertible transform to each frame that compresses the information into few coefficients.
- ❑ The DFT has some problems when used for this purpose:
 - N real $x[n] \leftrightarrow N$ complex $X[k]$: 2 real, $N/2 - 1$ conjugate pairs
 - DFT is of the periodic signal formed by replicating $x[n]$
 $\Rightarrow$ Spurious frequency components from boundary discontinuity



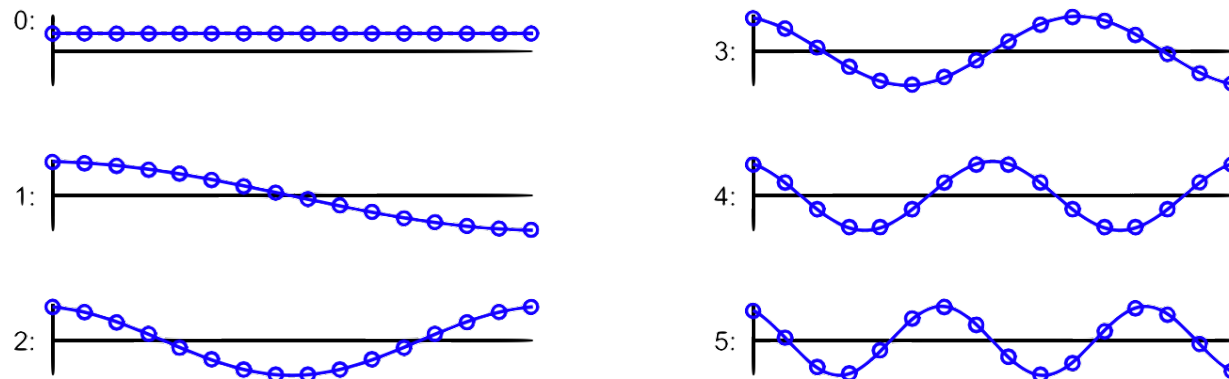
The Discrete Cosine Transform (DCT) overcomes these problems.

Basis Functions

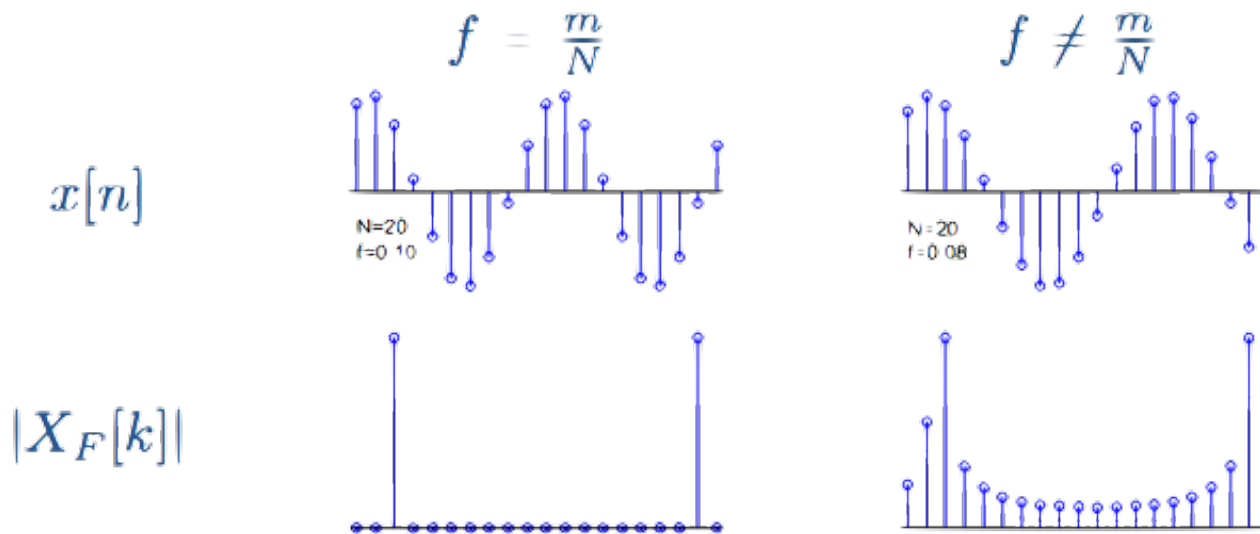
DFT basis functions: $x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j2\pi \frac{kn}{N}}$



DCT basis functions: $x[n] = \frac{1}{N} X[0] + \frac{2}{N} \sum_{k=1}^{N-1} X[k] \cos \frac{2\pi(2n+1)k}{4N}$

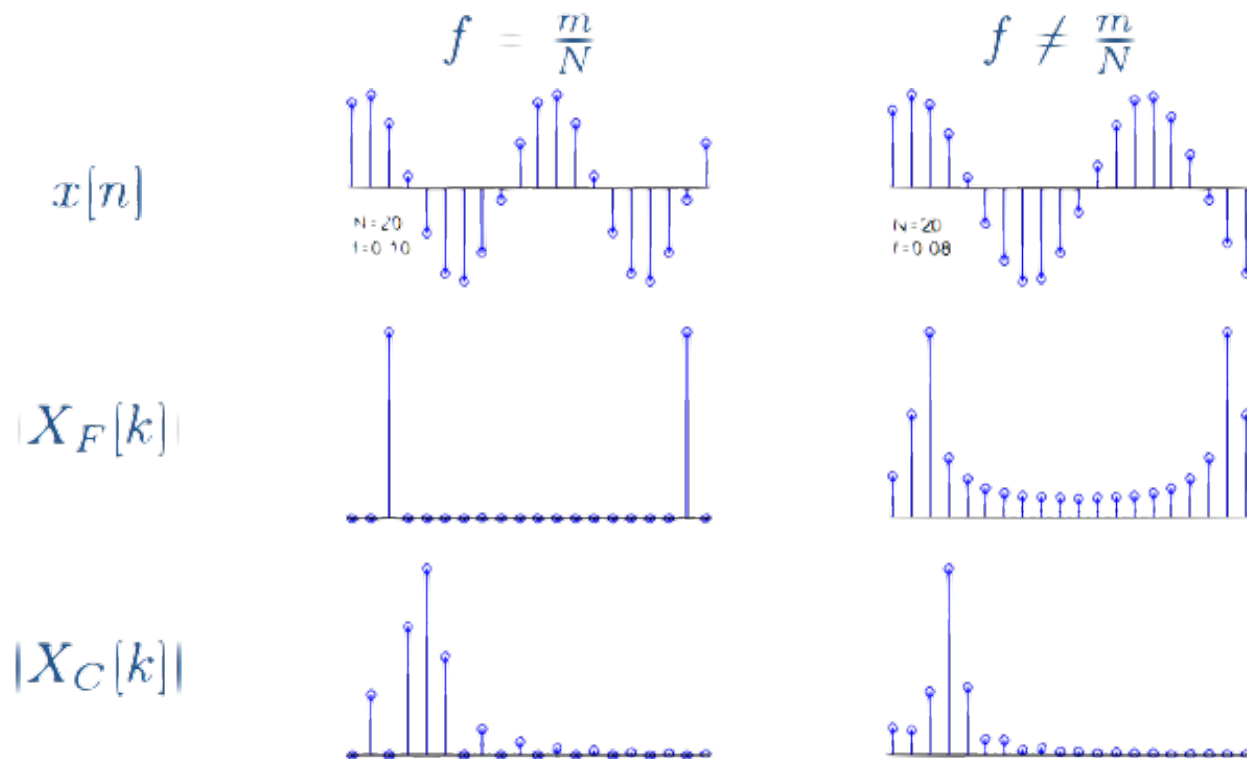


DFT of Sine Wave



DFT: Real $\rightarrow$ Complex; Freq range $[0, 1]$; Poorly localized unless $f = \frac{m}{N}$; $|X_F[k]| \propto k^{-1}$ for $Nf < k \ll \frac{N}{2}$

DCT of Sine Wave



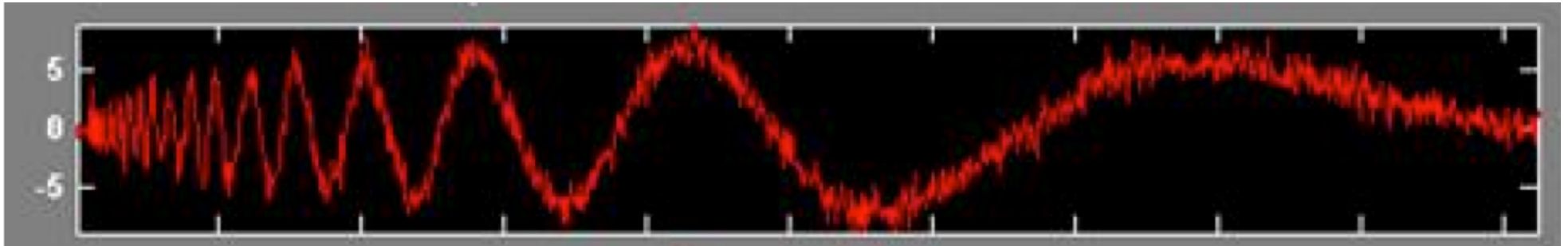
- DFT:** Real $\rightarrow$ Complex; Freq range $[0, 1]$; Poorly localized unless $f = \frac{m}{N}$; $|X_F[k]| \propto k^{-1}$ for $Nf < k \ll \frac{N}{2}$
- DCT:** Real $\rightarrow$ Real; Freq range $[0, 0.5]$; Well localized $\forall f$; $|X_C[k]| \propto k^{-2}$ for $2Nf < k < N$

Wavelet Transform



Motivation

- ❑ Some signals obviously have spectral characteristics that vary with time



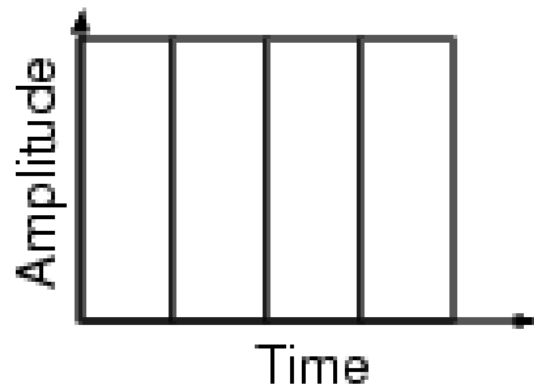


Criticism of Fourier Spectrum

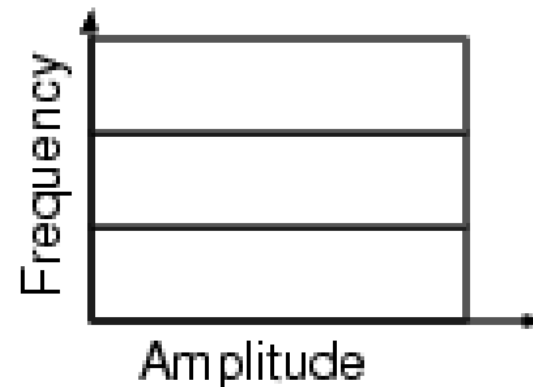
- ❑ It's giving you the spectrum of the 'whole time-series'
- ❑ Which is OK if the time-series is stationary. But what if it's not?
- ❑ We need a technique that can “march along” a time series and that is capable of:
 - Analyzing spectral content in different places
 - Detecting sharp changes in spectral character



Transform Comparison



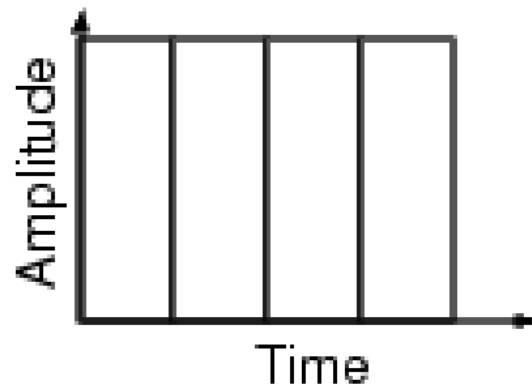
Time Domain (Shannon)



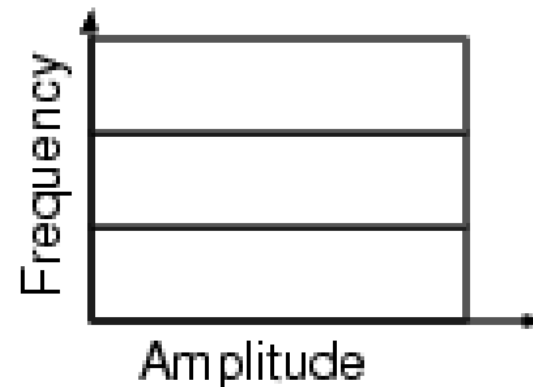
Frequency Domain (Fourier)



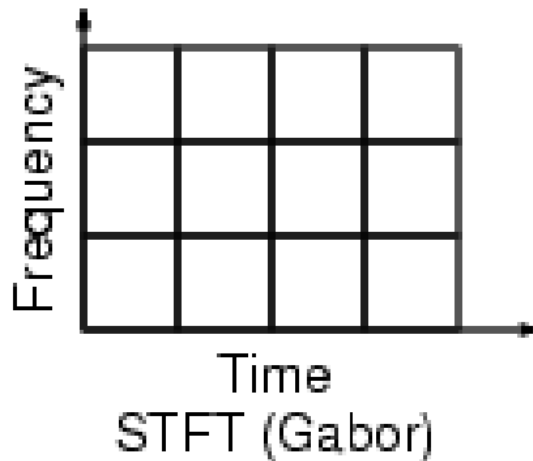
Transform Comparison



Time Domain (Shannon)



Frequency Domain (Fourier)

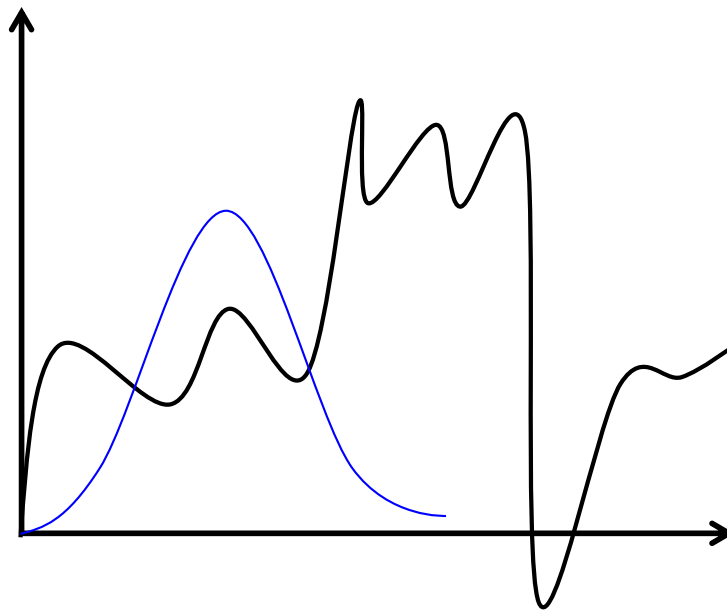


STFT (Gabor)



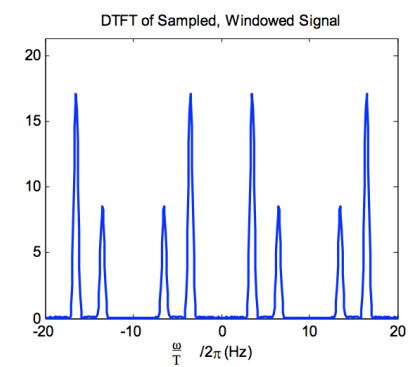
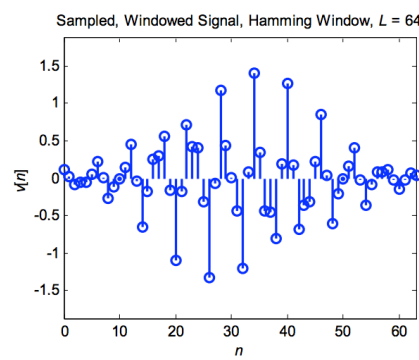
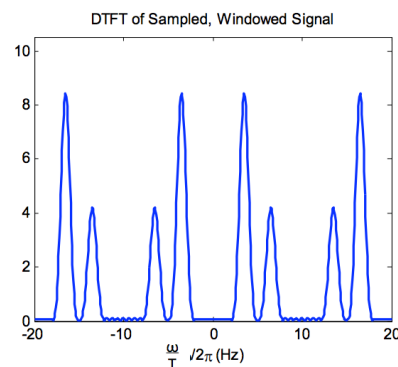
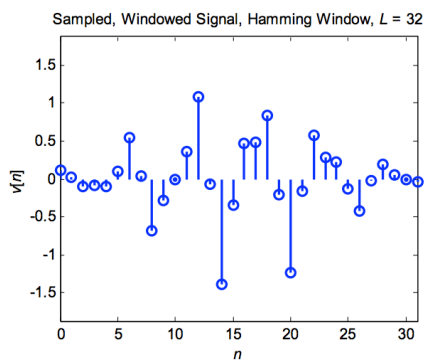
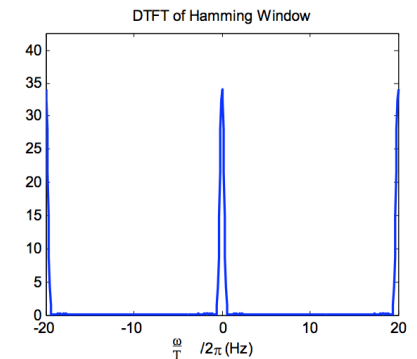
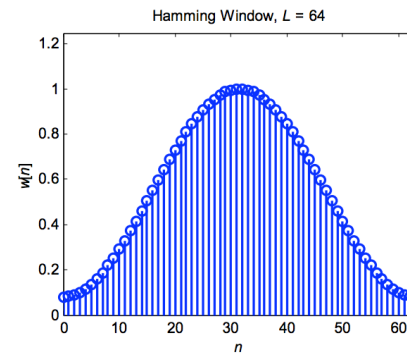
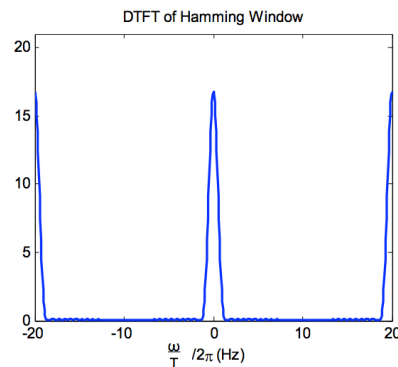
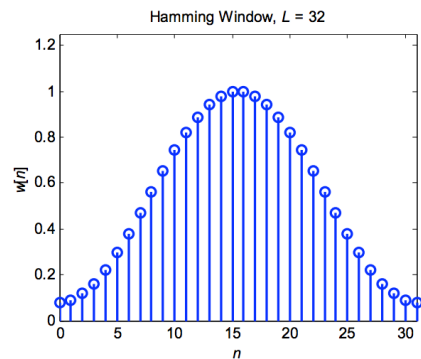
Discrete Time-Dependent FT

- ❑ Fixed window size, shift in time (Gabor)



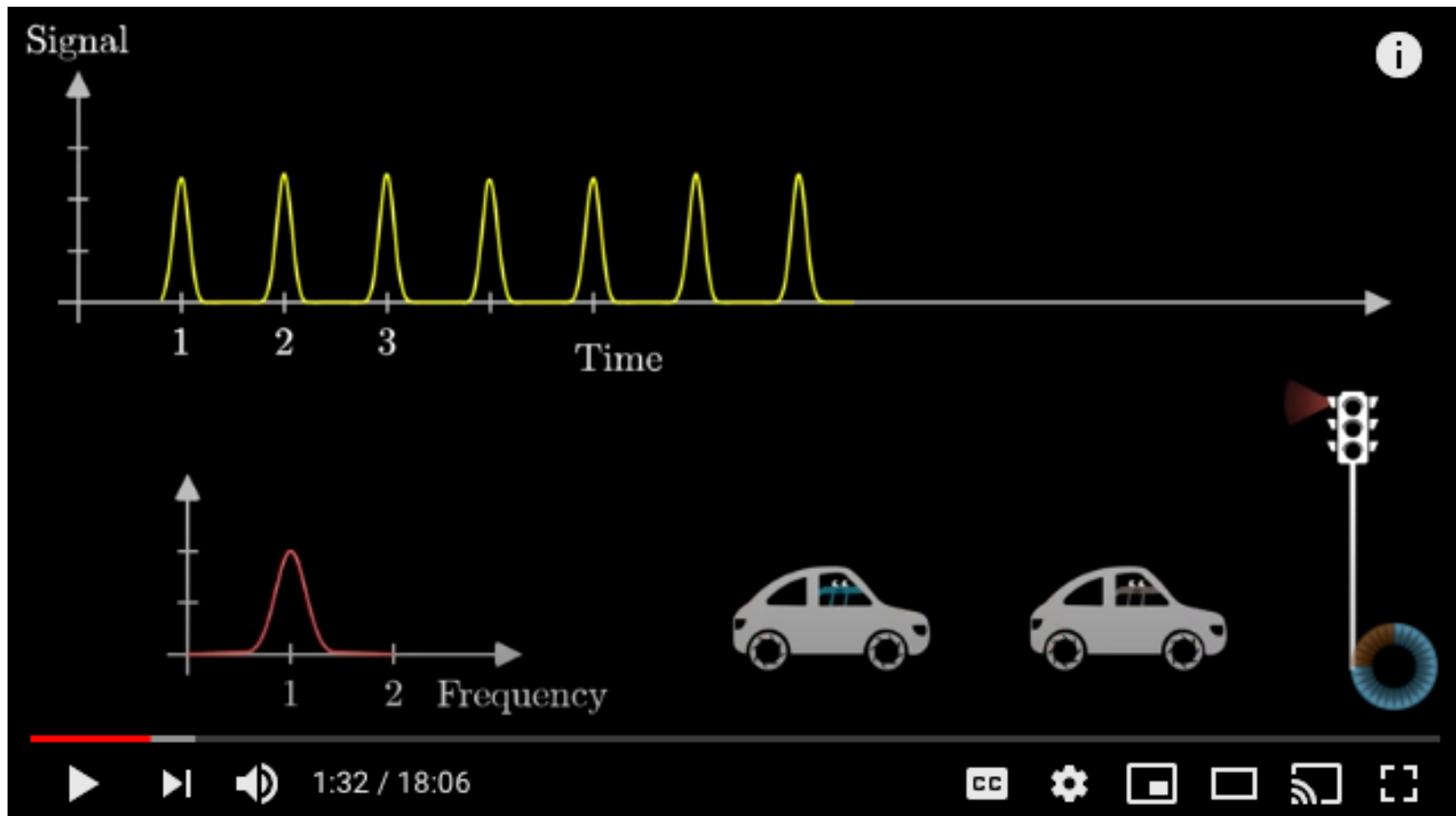
Windowed Sampled CT Signal Example

- ❑ As before, the sampling rate is $\Omega_s/2\pi=1/T=20\text{Hz}$
- ❑ Hamming Window, $L = 32$ vs. $L = 64$





Uncertainty Principle

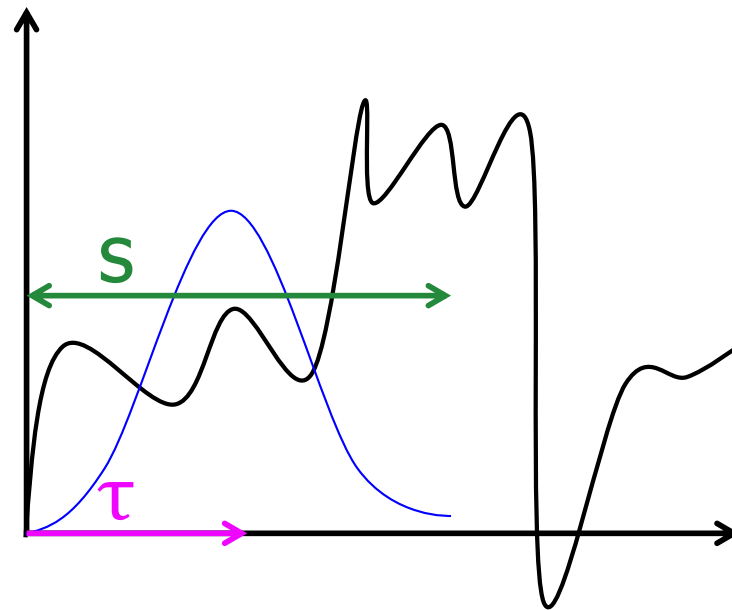


<https://youtu.be/MBnnXbOM5S4?t=49>



Discrete Time-Dependent FT

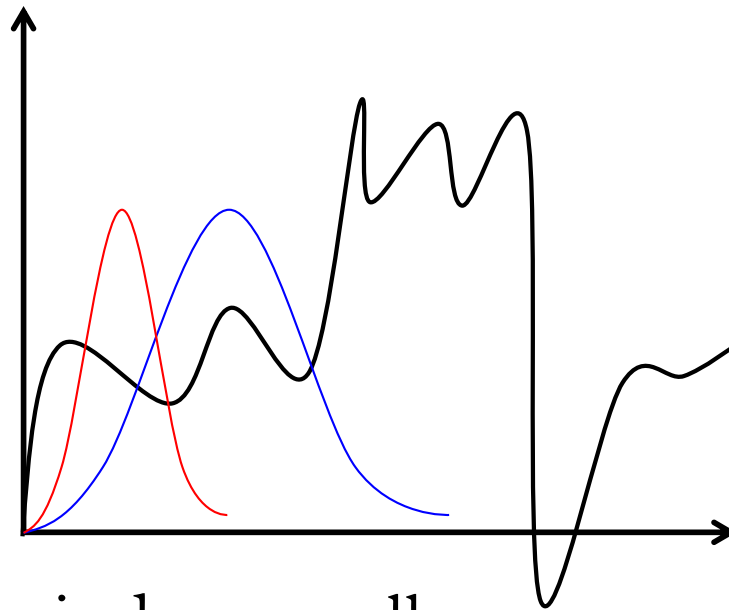
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Discrete Time-Dependent FT

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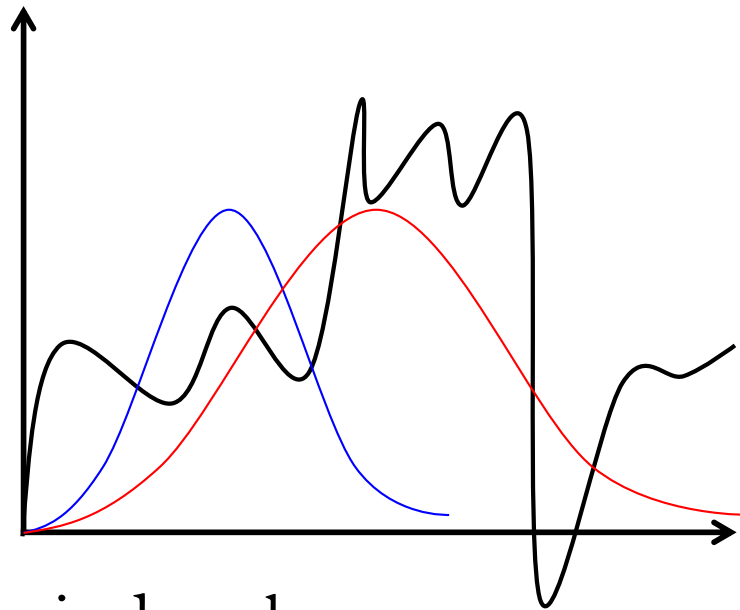


- ❑ Make the window smaller
 - Better localization in time
 - Less spectral resolution



Discrete Time-Dependent FT

- ❑ Fixed window size, shift in time (Gabor)

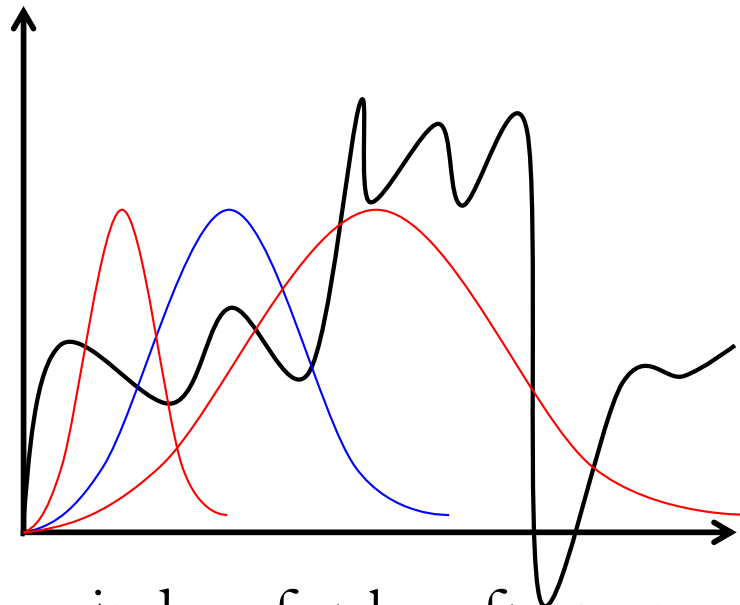


- ❑ Make the window larger
 - Worse localization in time
 - More spectral resolution

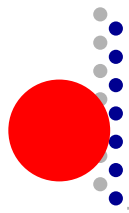


Discrete Time-Dependent FT

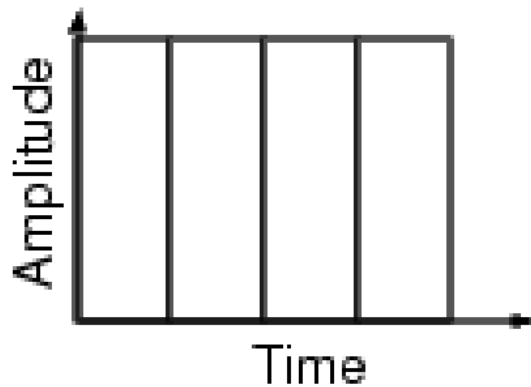
- ❑ Fixed window size, shift in time (Gabor)



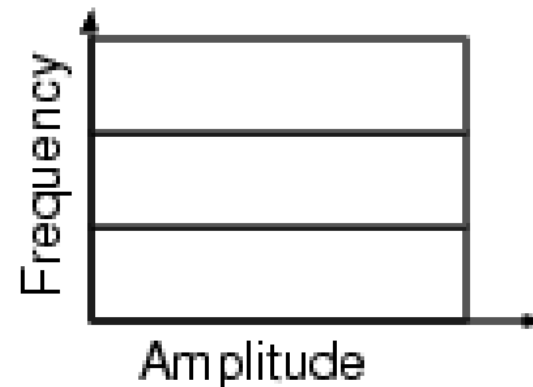
- Use a big window for low frequency content that is not localized in time
- Use a small window for high frequency content that is localized in time



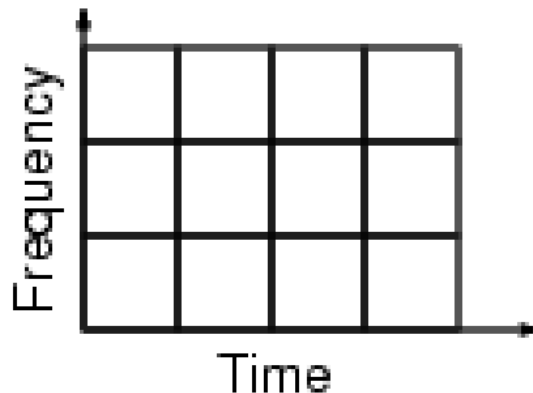
Transform Comparison



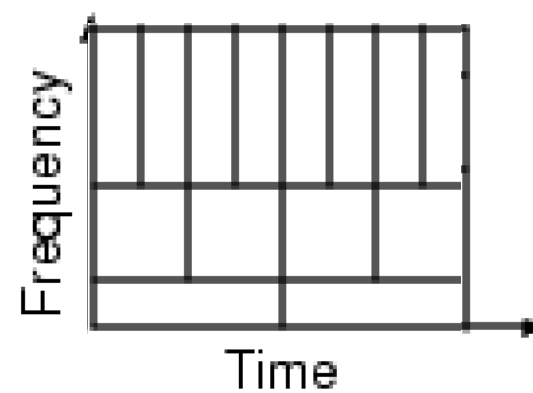
Time Domain (Shannon)



Frequency Domain (Fourier)



STFT (Gabor)

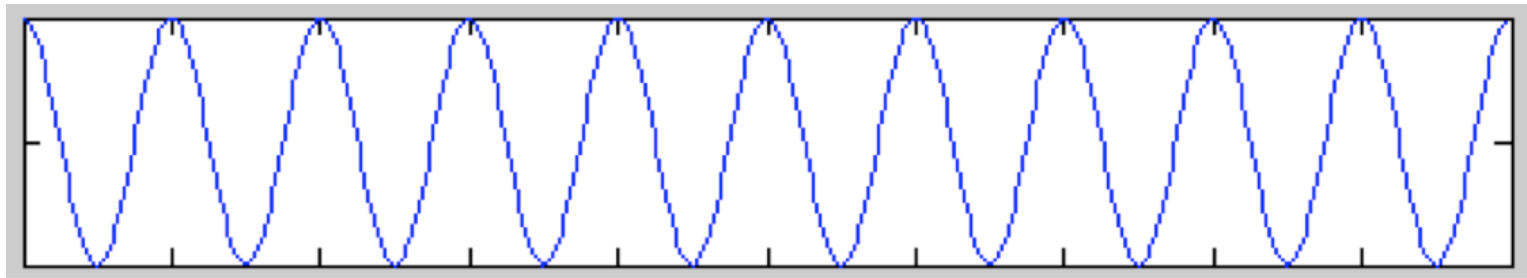


Wavelet Analysis

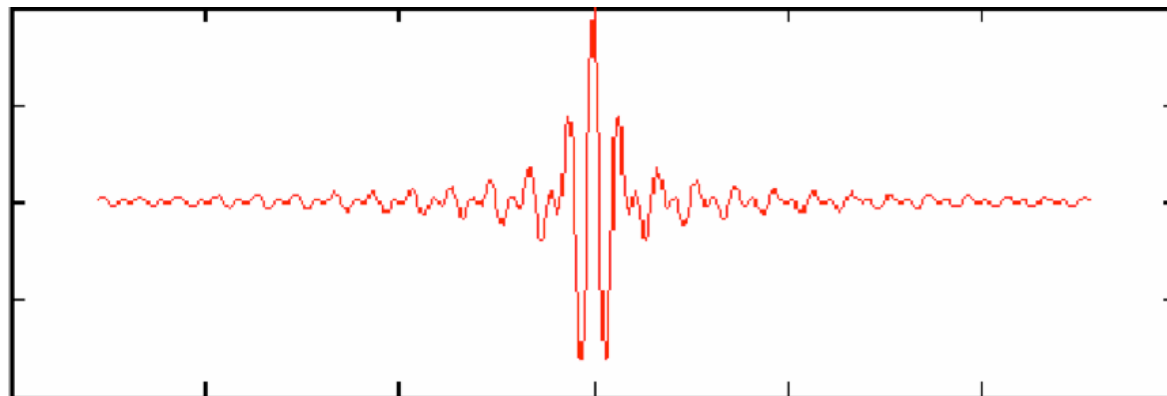


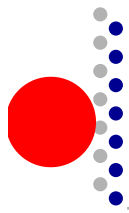
Fourier vs. Wavelet

- ❑ Fourier Analysis is based on an indefinitely long cosine wave of a specific frequency



- ❑ Wavelet Analysis is based on a short duration wavelet of a specific center frequency

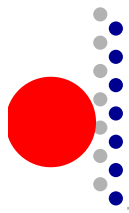




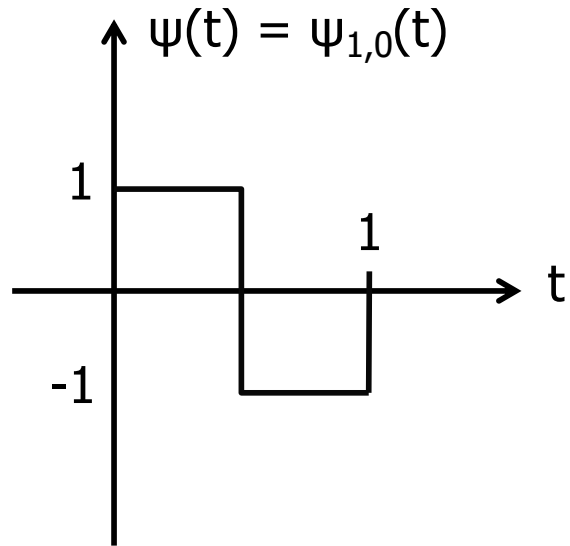
Wavelet Transform

- All wavelets derived from *mother* wavelet

$$\psi_{s,\tau}(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t - \tau}{s}\right)$$



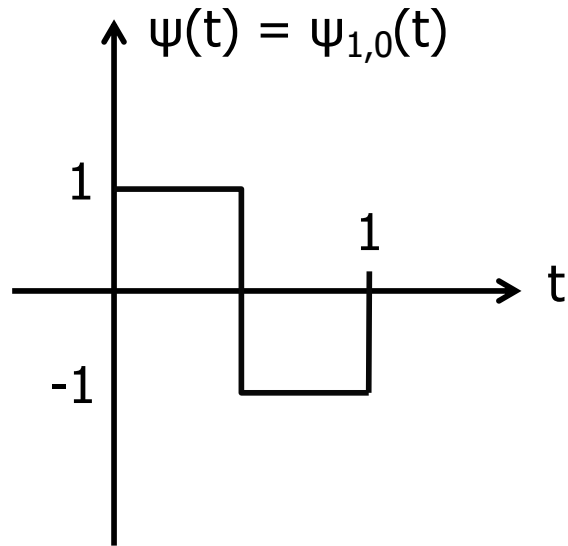
Example: Haar Wavelet



$$\psi_{s,\tau}(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t-\tau}{s}\right)$$

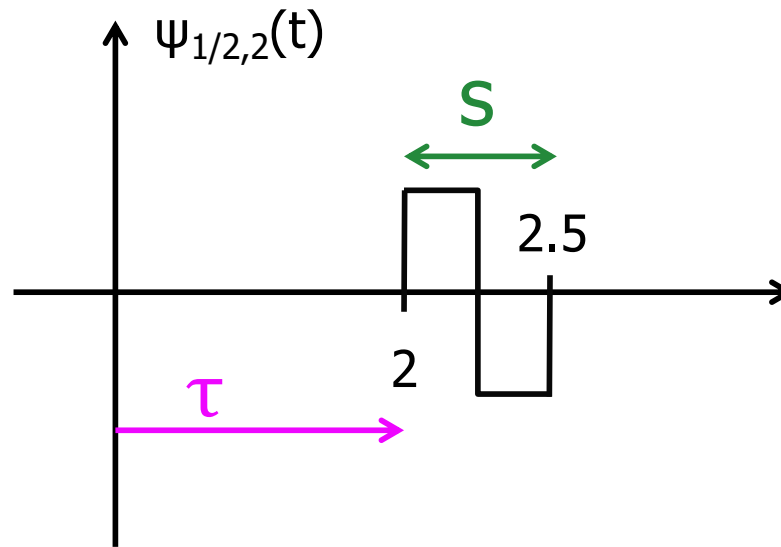


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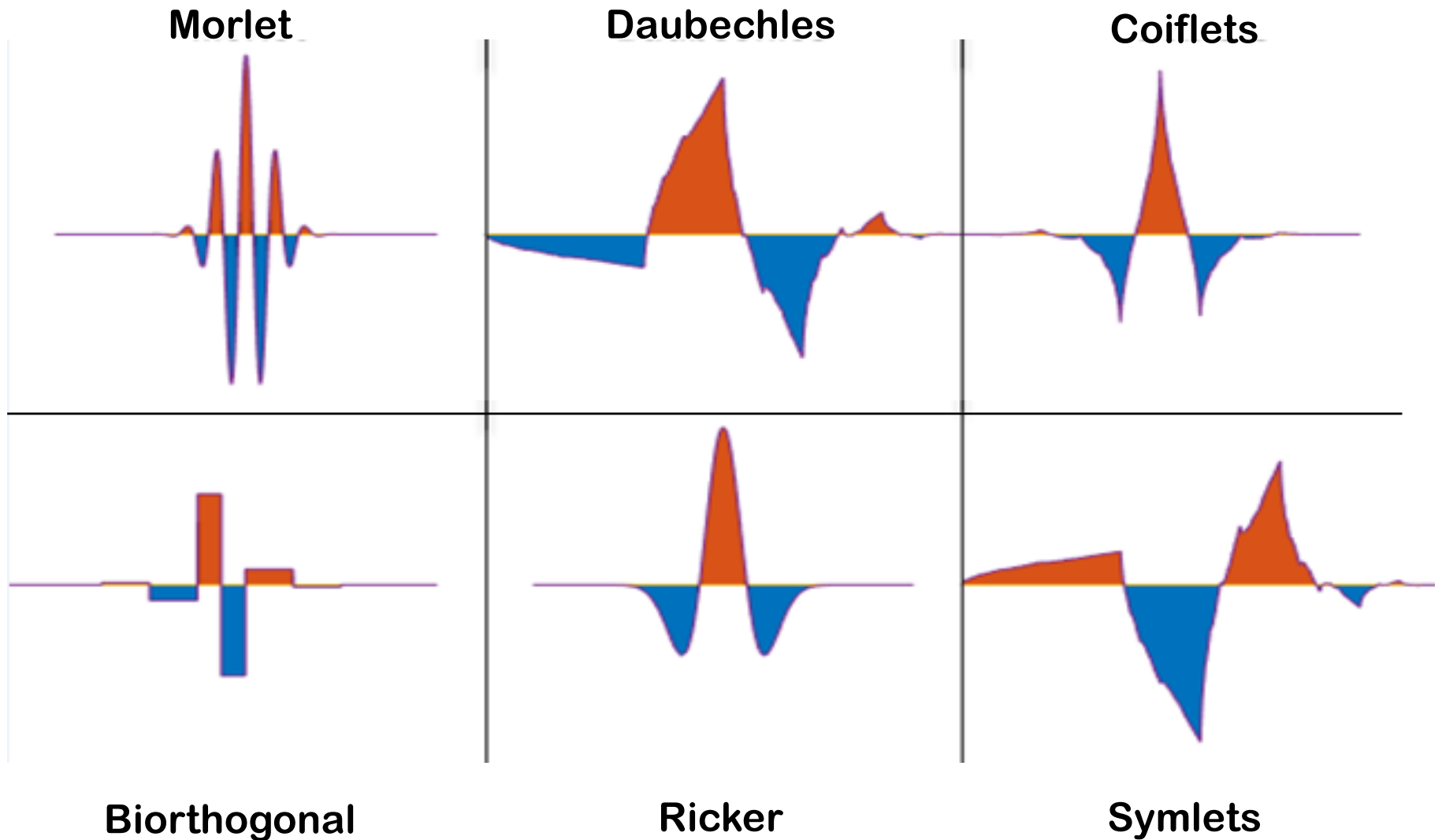
$$\psi_{s,\tau}(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t-\tau}{s}\right)$$

$$s=1/2, \tau=2$$





Examples of Wavelets





Ingrid Daubechies



- ❑ Defined worldwide standard for image compression
- ❑ <https://www.fi.edu/en/laureates/ingrid-daubechies#:~:text=Her%20contributions%20have%20revolutionized%20and,the%20JPEG2000%20image%20processing%20standard.>

Wavelet – Scaled and Shifted

$$\psi_{s,\tau}(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t - \tau}{s}\right)$$

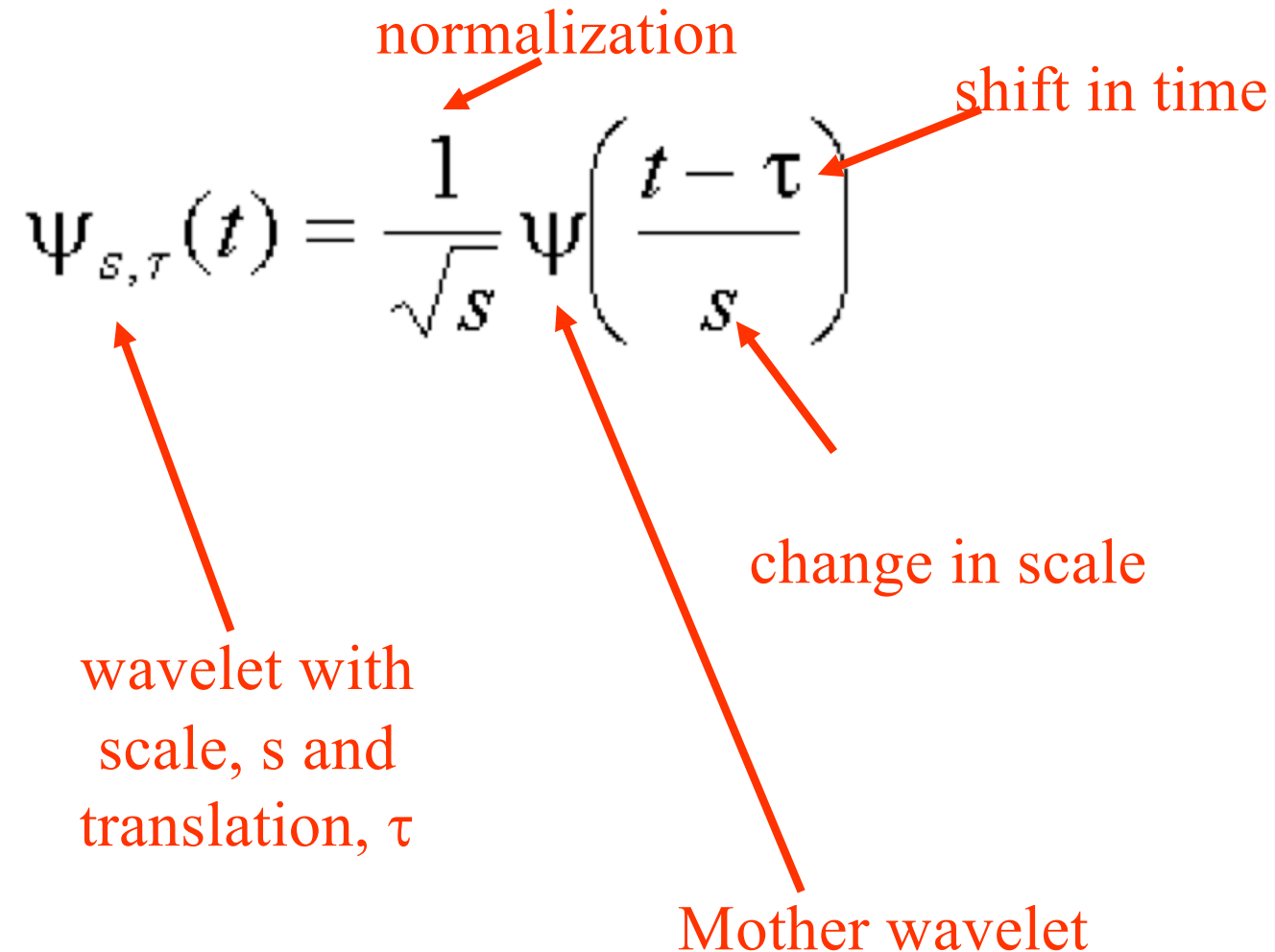
normalization

shift in time

change in scale

Mother wavelet

wavelet with scale, s and translation, τ



Continuous Wavelet Transform

time-series

$$\gamma(s, \tau) = \int f(t) \Psi_{s, \tau}(t) dt$$

coefficient of wavelet
with
scale, s and time, τ

wavelet with
scale, s , and shift, τ

Inverse Wavelet Transform

- Build up a time-series as sum of wavelets of different scales, s , and positions, t

$$f(t) = \int \int \gamma(s, \tau) \psi_{s, \tau}(t) d\tau ds$$

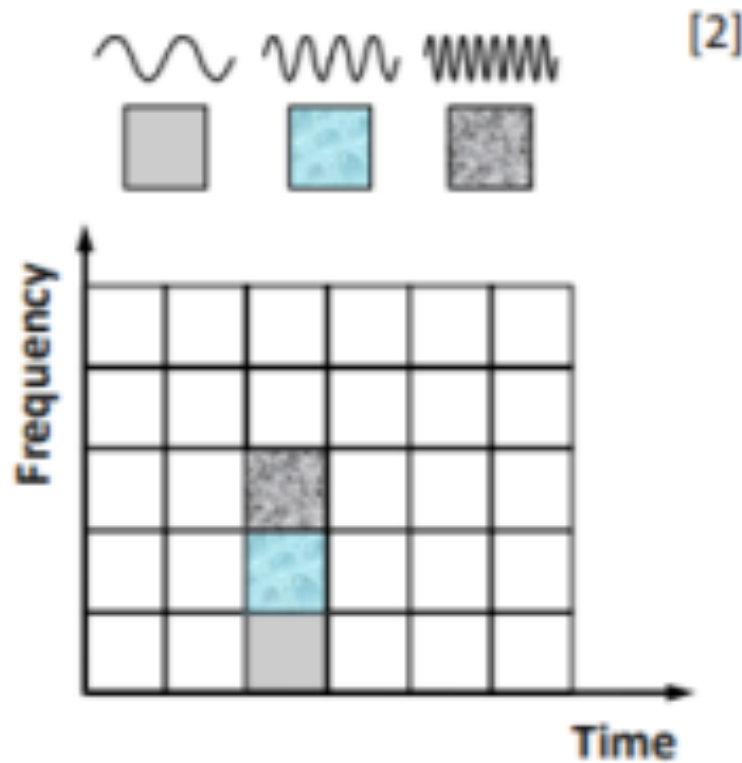
time-series

coefficients
of wavelets

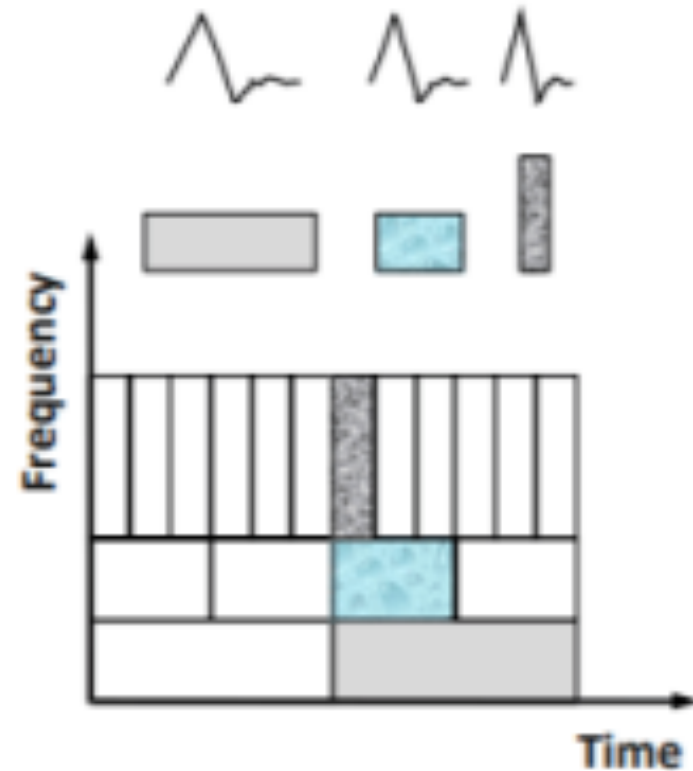
wavelet with
scale, s and time, τ

Wavelet Basis Functions

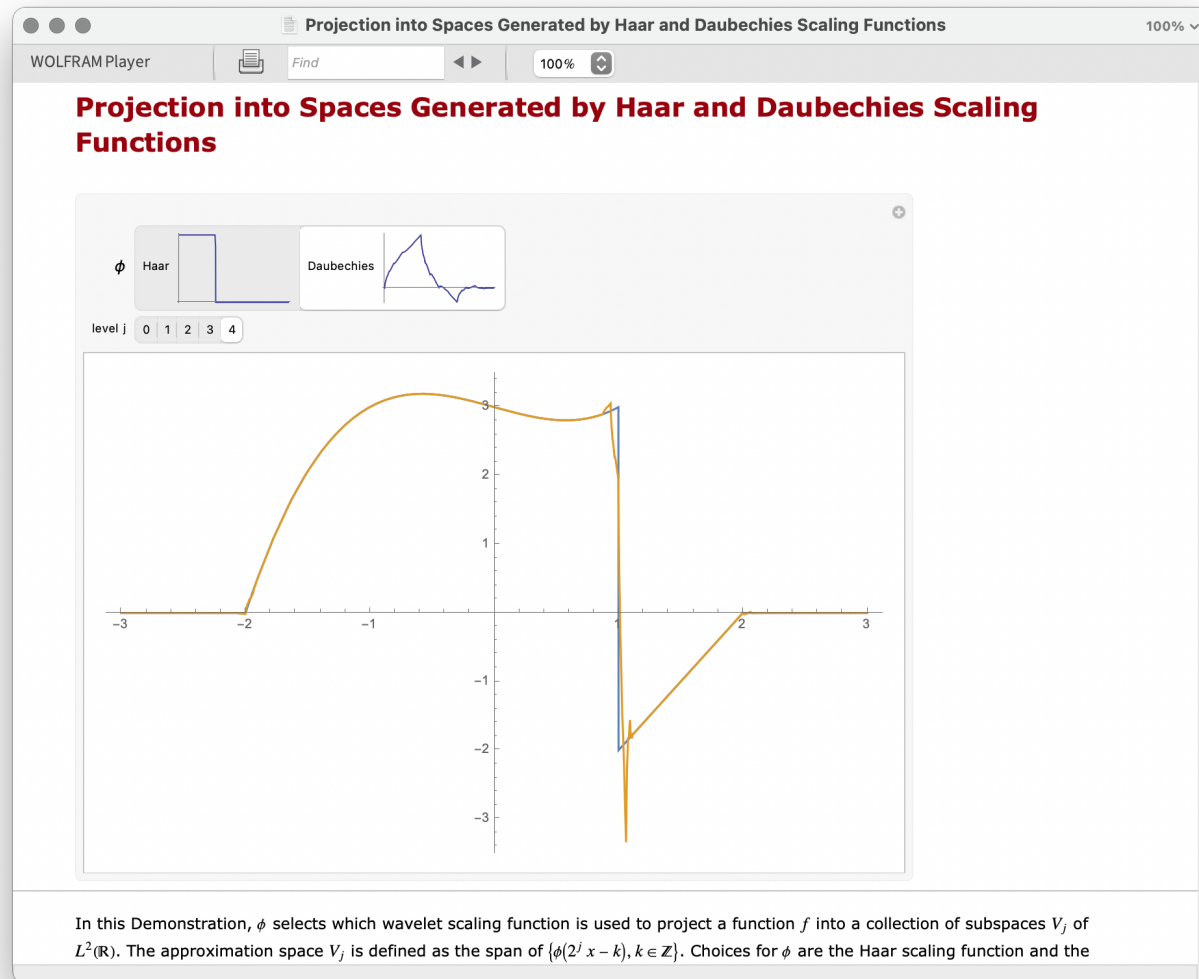
Fourier basis functions



Wavelet basis functions



Wavelet reconstruction demo



<https://demonstrations.wolfram.com/ProjectionIntoSpacesGeneratedByHaarAndDaubechiesScalingFunct/>



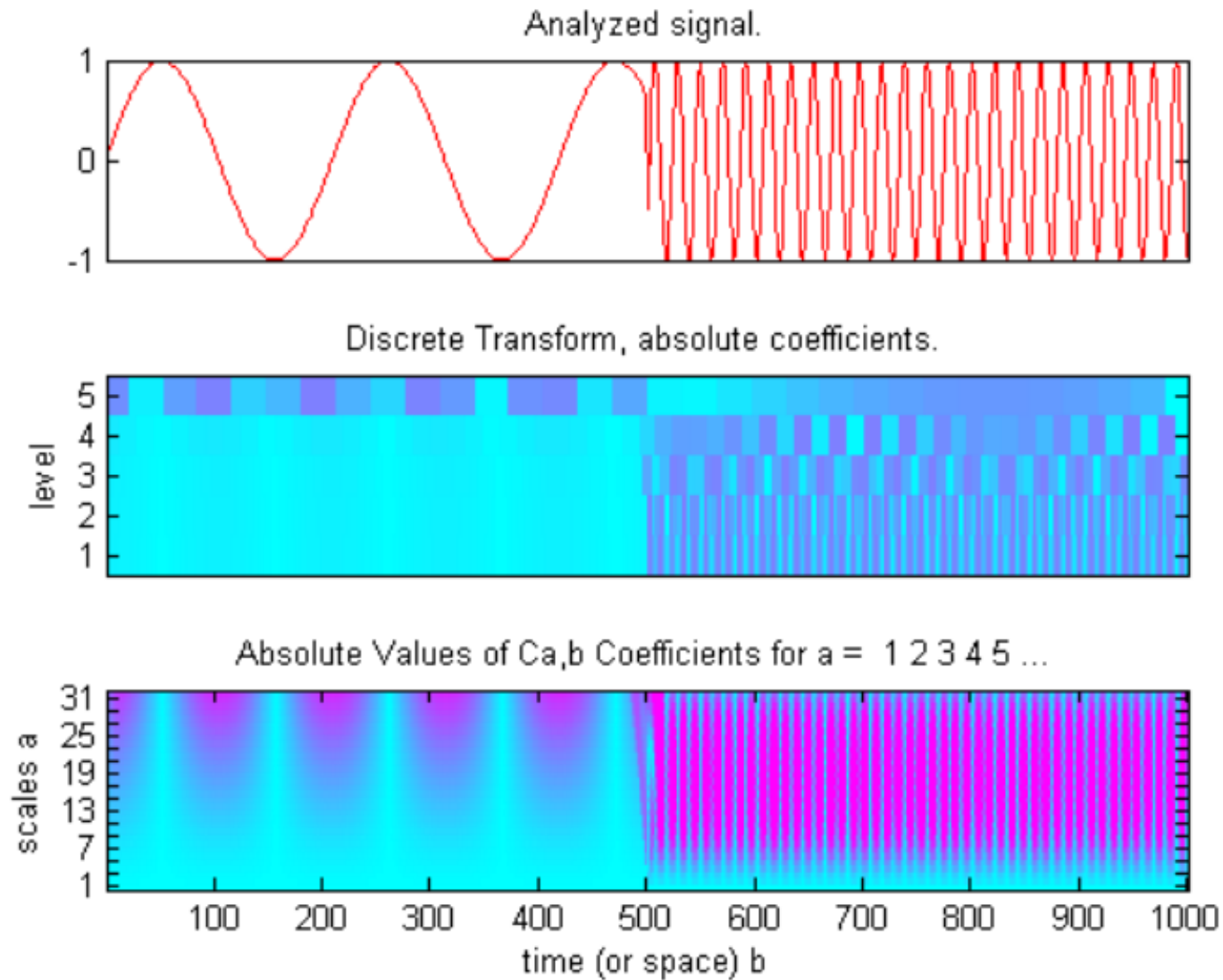
Discrete wavelets:

- Scale wavelets only by integer powers of 2
 - $s_j = 2^j$
- And shifting by integer multiples of s_j for each successive scale
 - $\tau_{j,k} = k2^j$
- Then $\gamma(s_j, \tau_{j,k}) = \gamma_{jk}$
 - where $j = 1, 2, \dots, \infty$, and $k = -\infty \dots -2, -1, 0, 1, 2, \dots, \infty$

$$\gamma_{j,k} = \frac{1}{\sqrt{2^j}} \int f(t) \Psi\left(\frac{t - k2^j}{2^j}\right) dt$$



DWT vs CWT





Wavelet Transform

- Determining the wavelet coefficients for a fixed scale, s , can be thought of as a filtering operation

$$\gamma(s, \tau) = \int f(t) \Psi_{s, \tau}(t) dt$$

$$\gamma_s(\tau) = \int f(t) \Psi_s(t - \tau) dt = f(\tau) * \Psi_s(-\tau)$$

- where

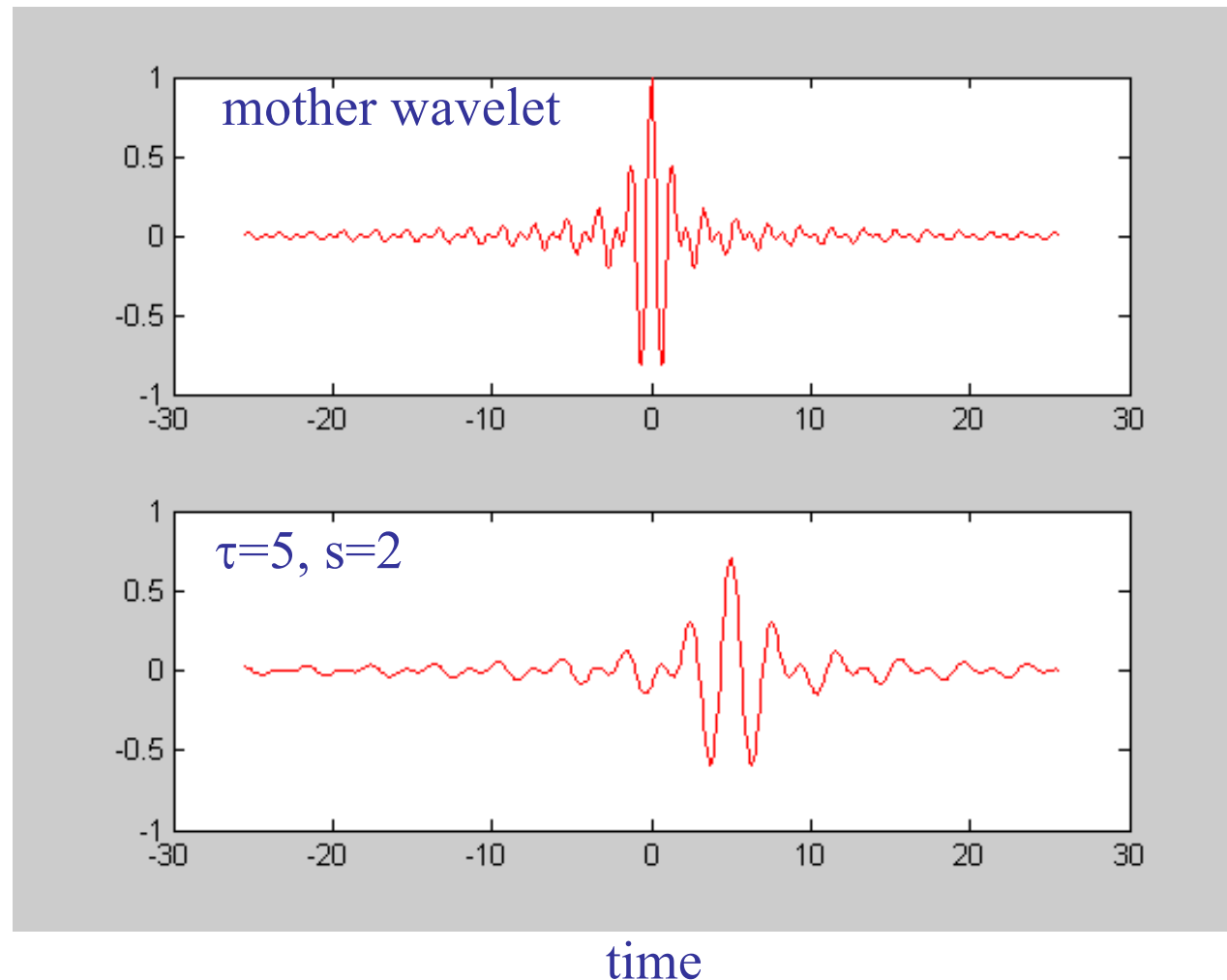
If wavelet is even,
 $\Psi(-\tau) = \Psi(\tau)$

$$\Psi_s(t) = \frac{1}{\sqrt{s}} \Psi\left(\frac{t}{s}\right)$$

Shannon Wavelet

$$\psi_{s,\tau}(t) = \frac{1}{\sqrt{s}} \psi\left(\frac{t-\tau}{s}\right) =$$

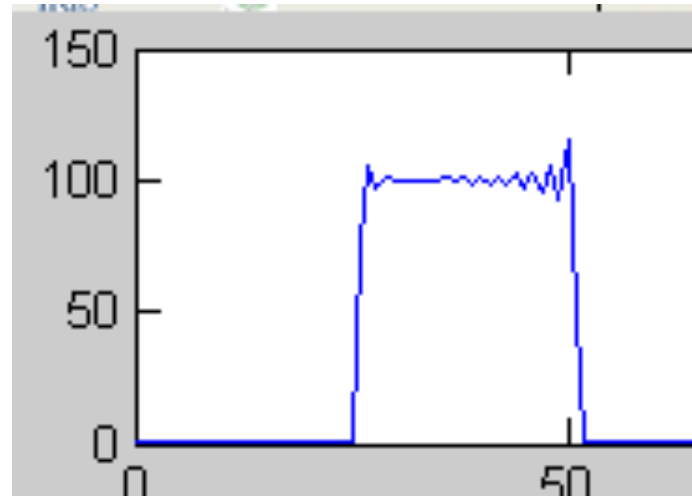
□ $\Psi(t) = 2 \operatorname{sinc}(2t) - \operatorname{sinc}(t)$





Fourier spectrum of Shannon Wavelet

$$\Psi_s(j\Omega)$$



frequency, Ω

- ❑ Wavelet coefficients are a result of bandpass filtering

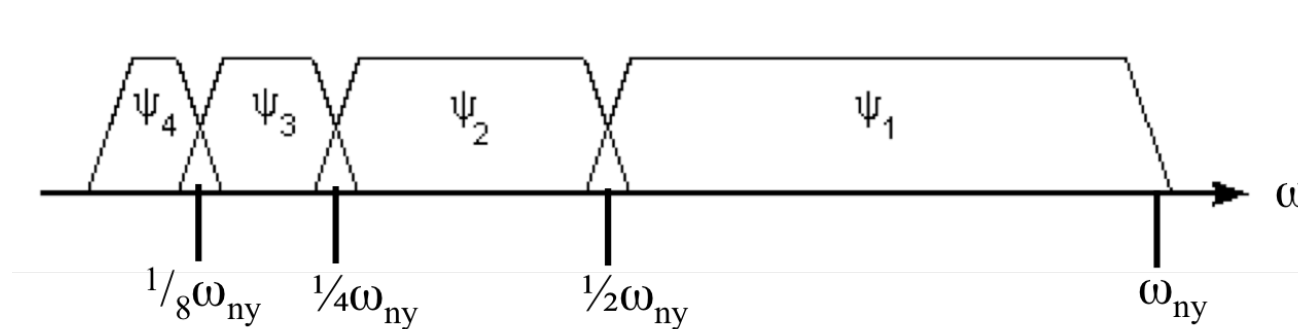


Discrete Wavelet Transform

- The coefficients of Ψ is just the band-pass filtered time-series, where Ψ is the wavelet, now viewed as the impulse response of a bandpass filter.
 - Discrete wavelet $\rightarrow s = 2^j$

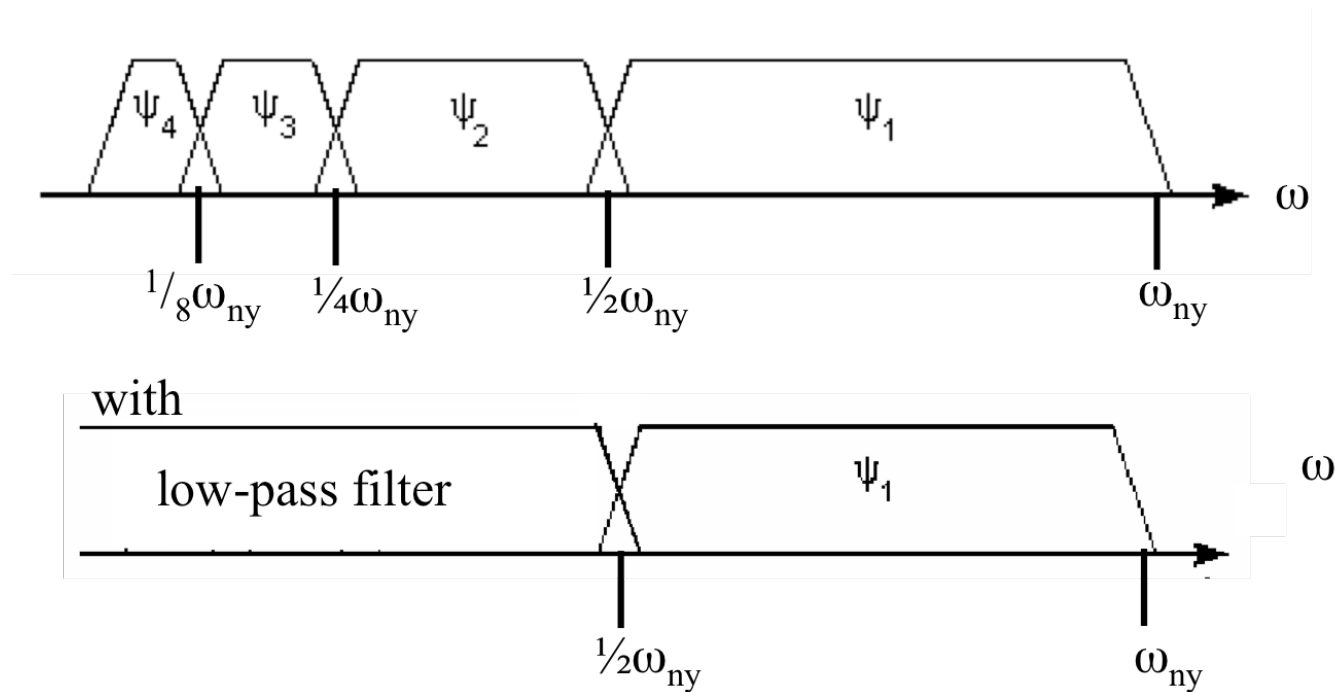
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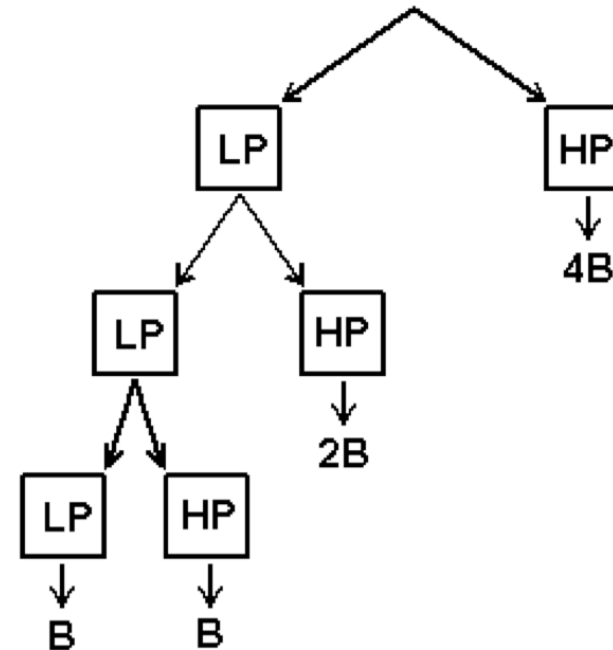
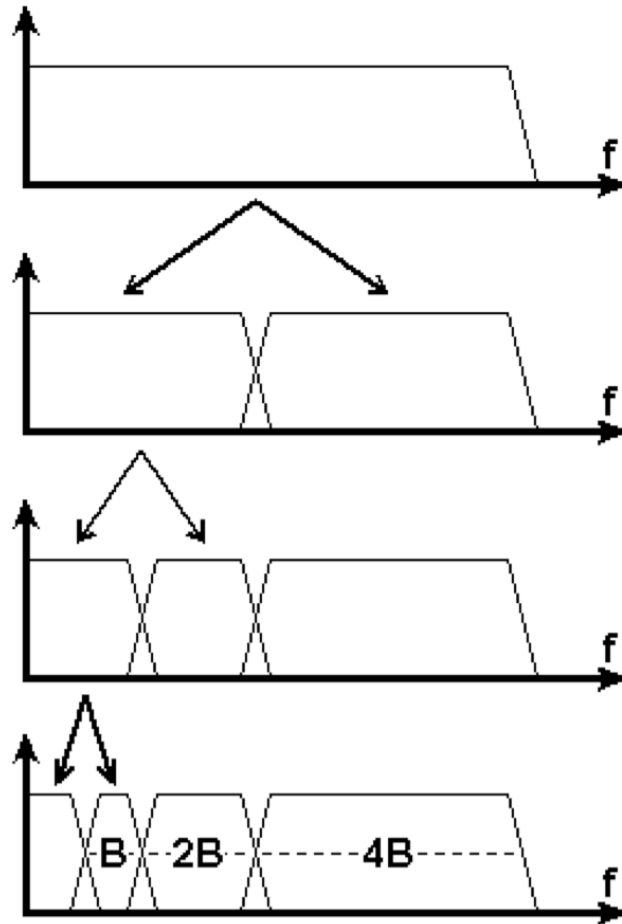
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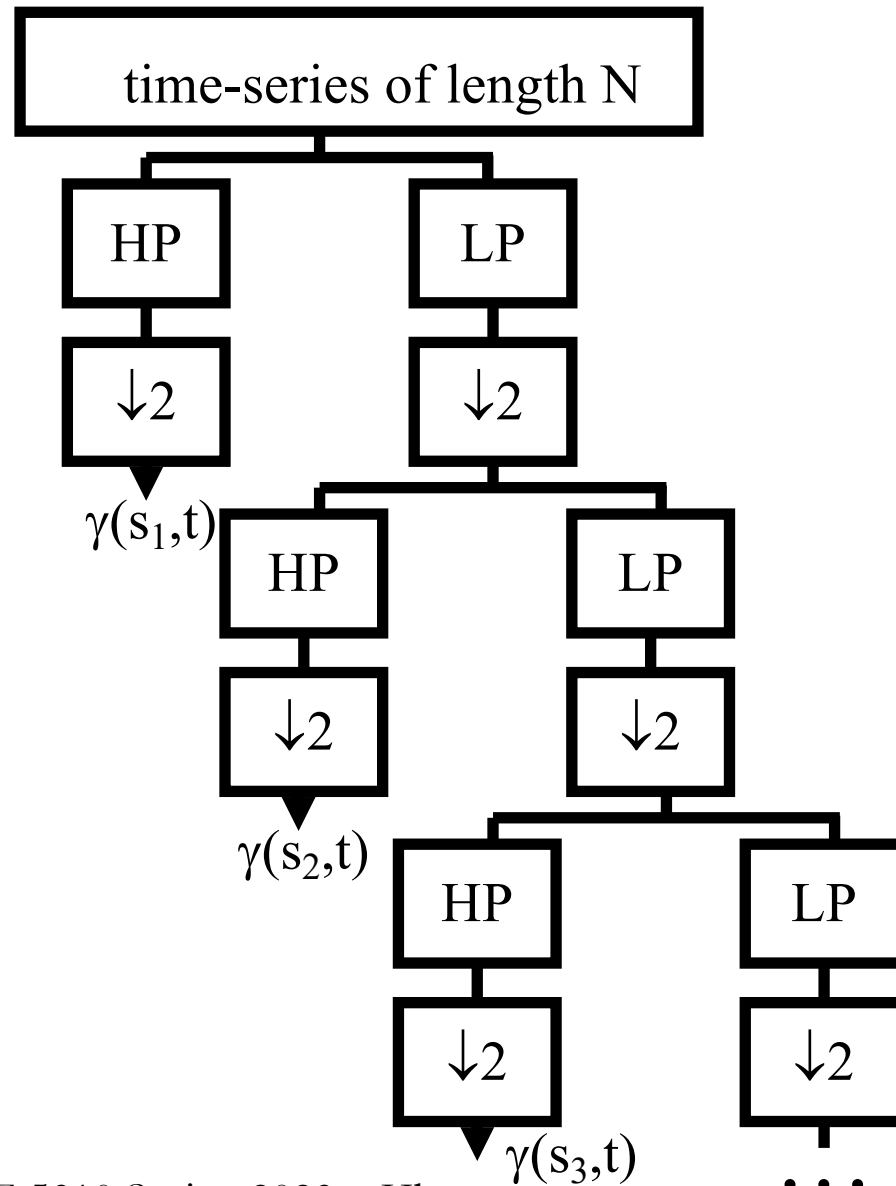


Digital Wavelet as Multirate Filter Bank

□ Repeat recursively!



Digital Wavelet as Multirate Filter Bank



$\gamma(s_1,t)$: $N/2$ coefficients

$\gamma(s_2,t)$: $N/4$ coefficients

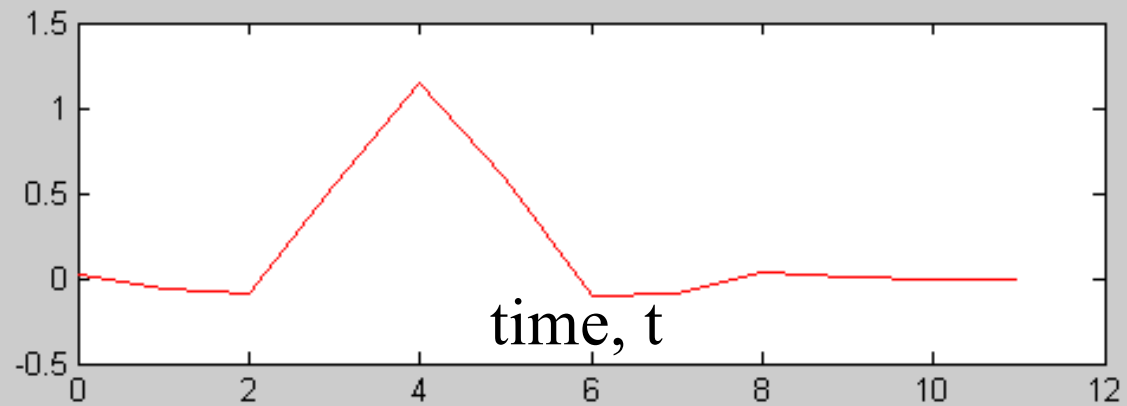
$\gamma(s_2,t)$: $N/8$ coefficients

Total: N coefficients

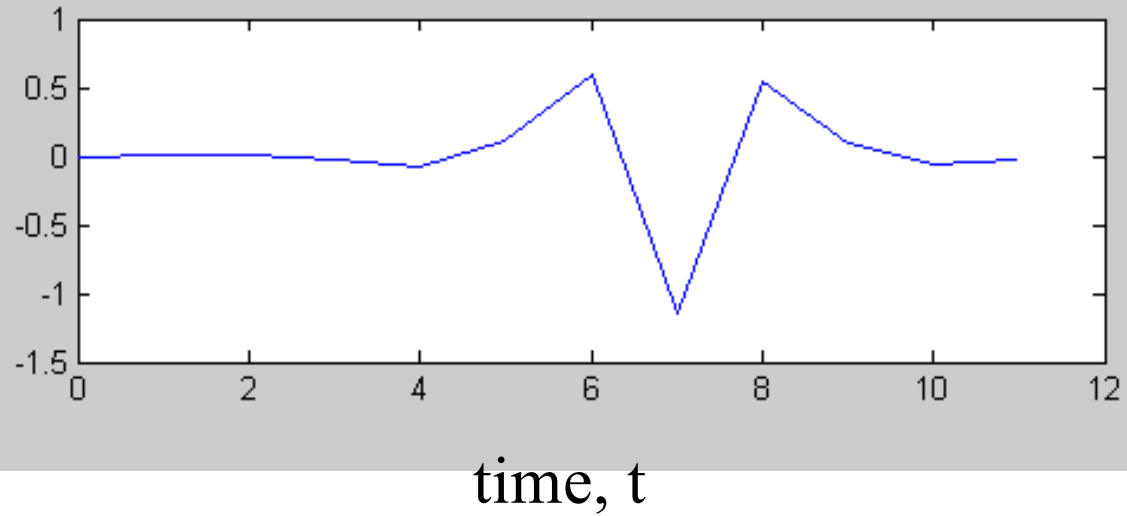


Impulse Responses

Coiflet low pass filter



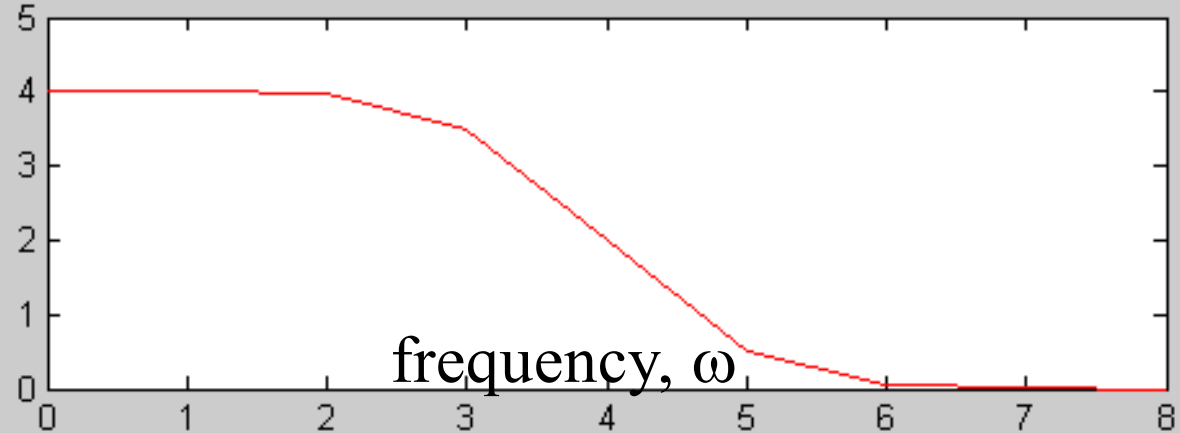
Coiflet high-pass filter



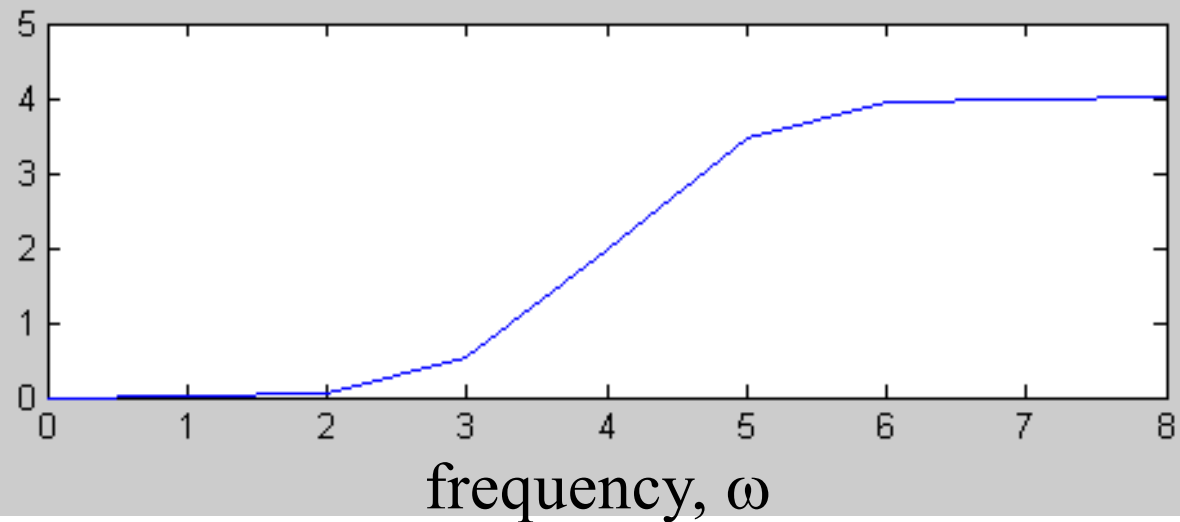


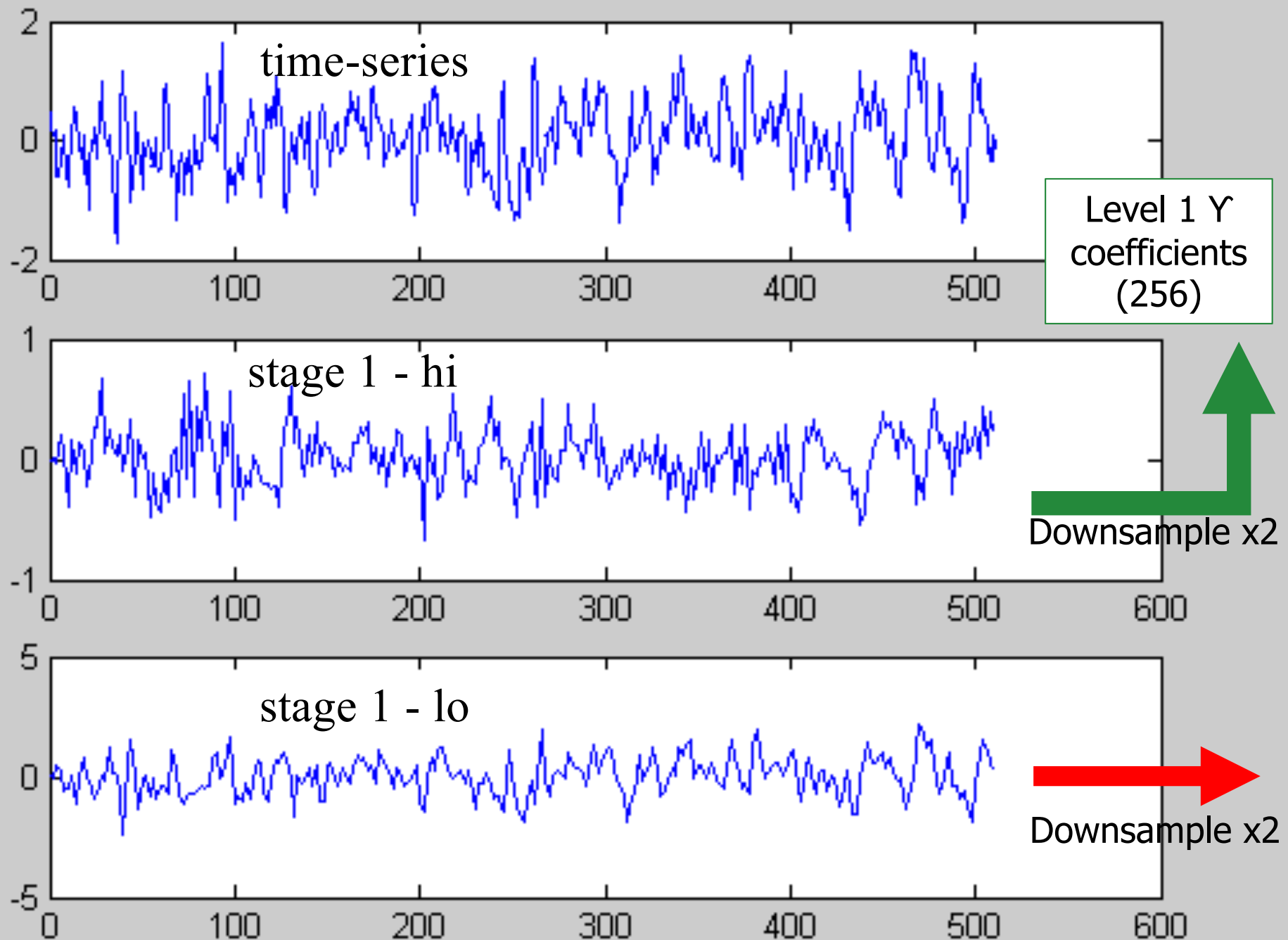
Filter Responses

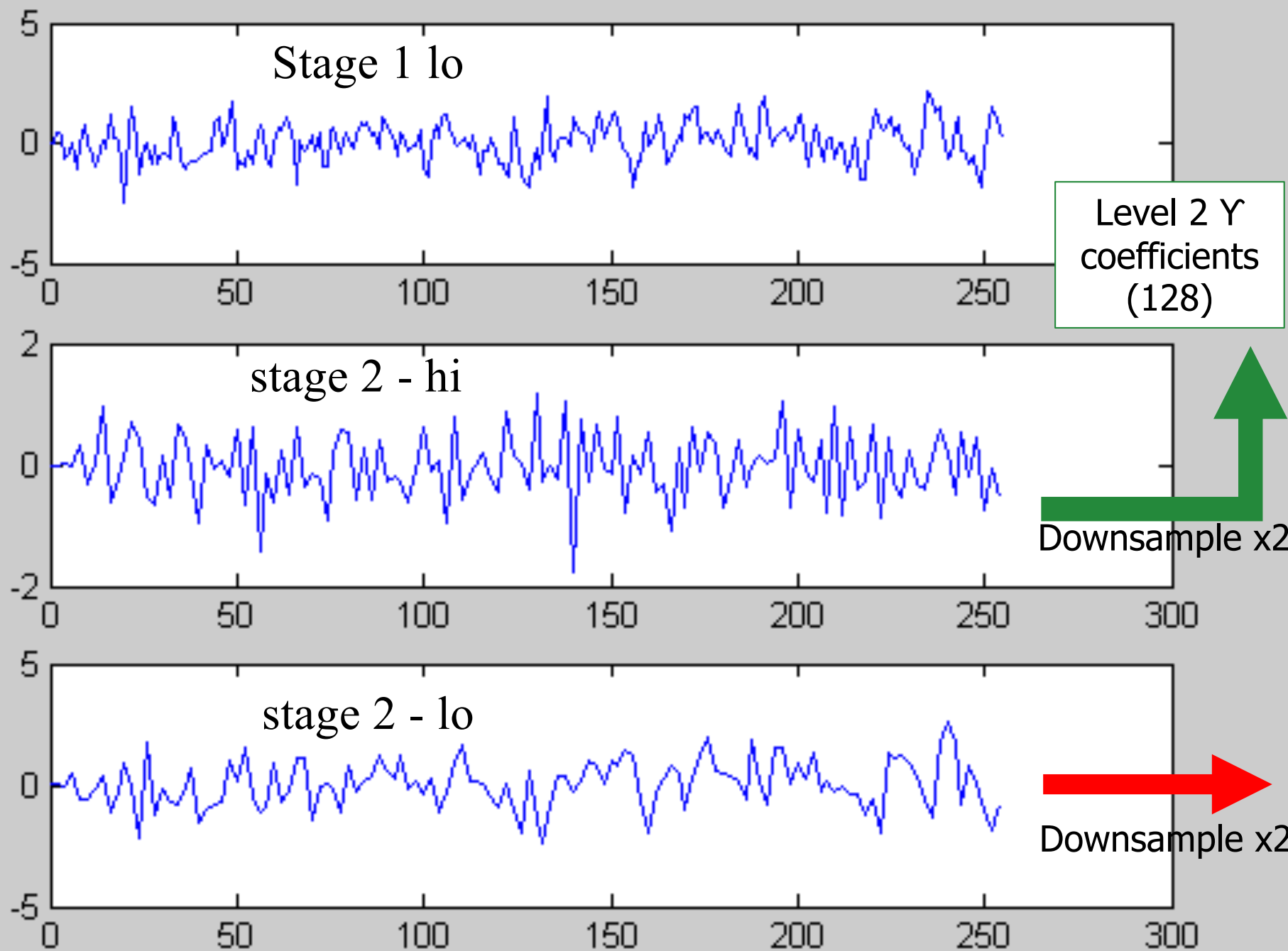
Spectrum of low pass filter

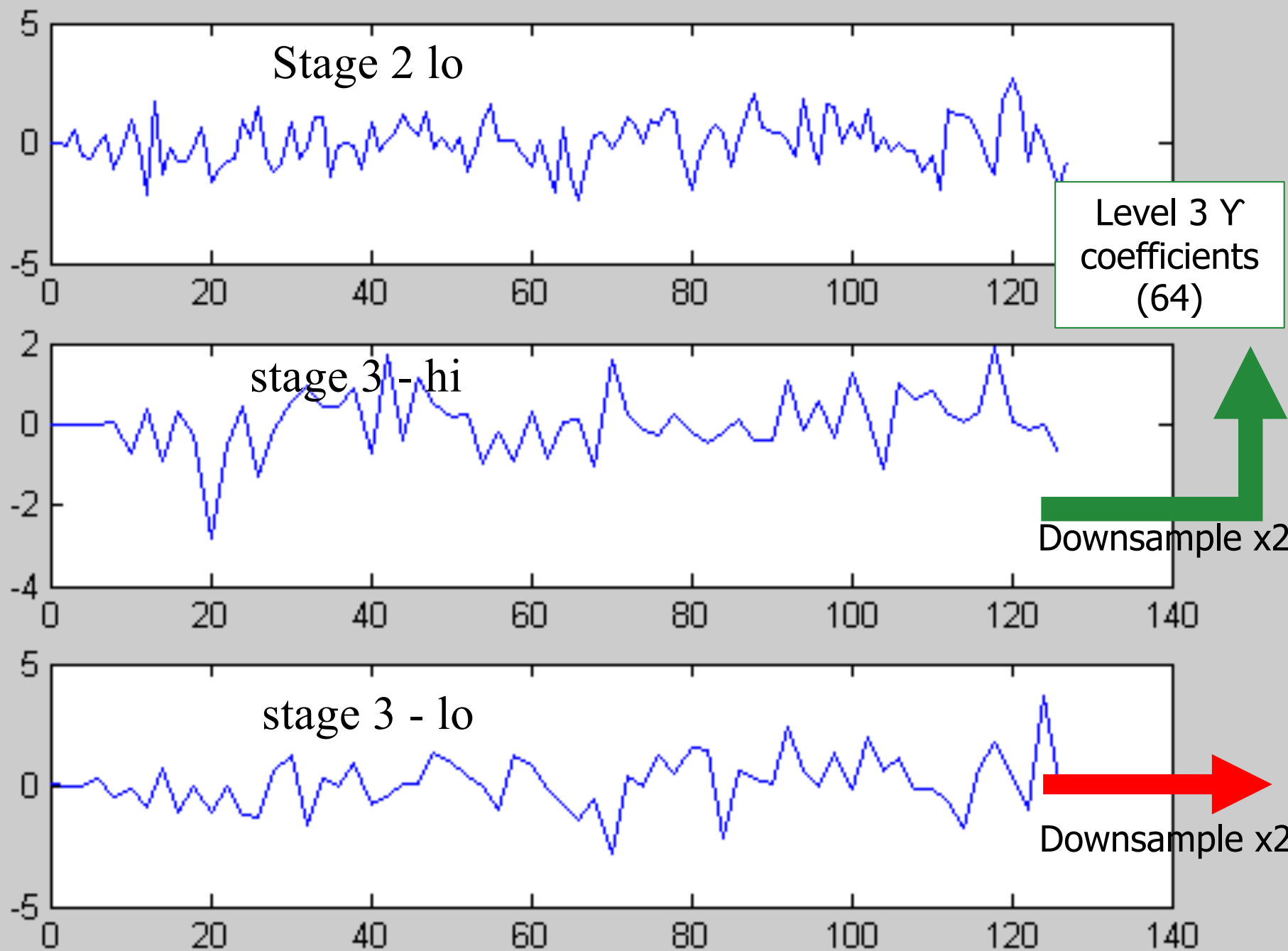


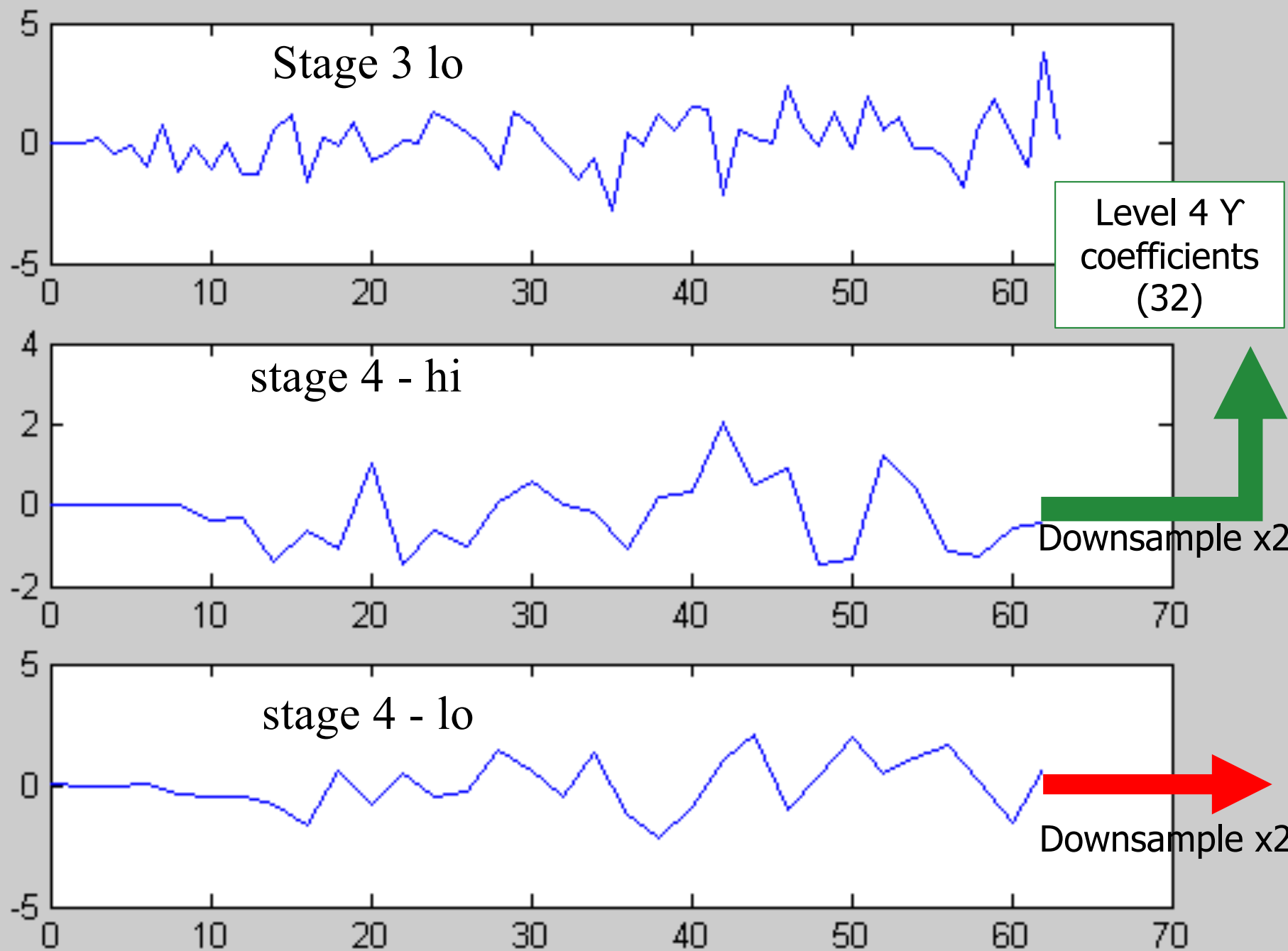
Spectrum of high pass filter

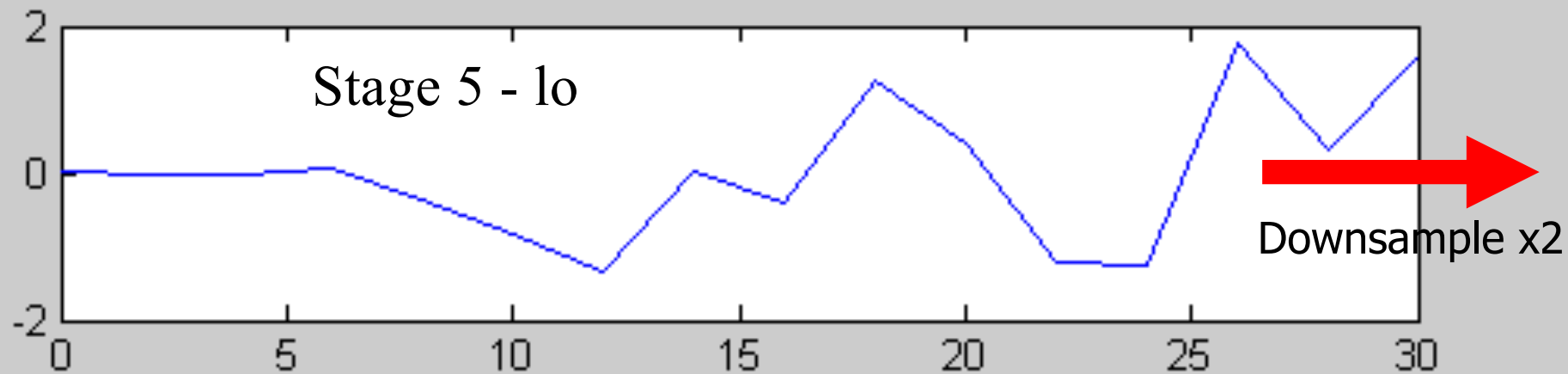
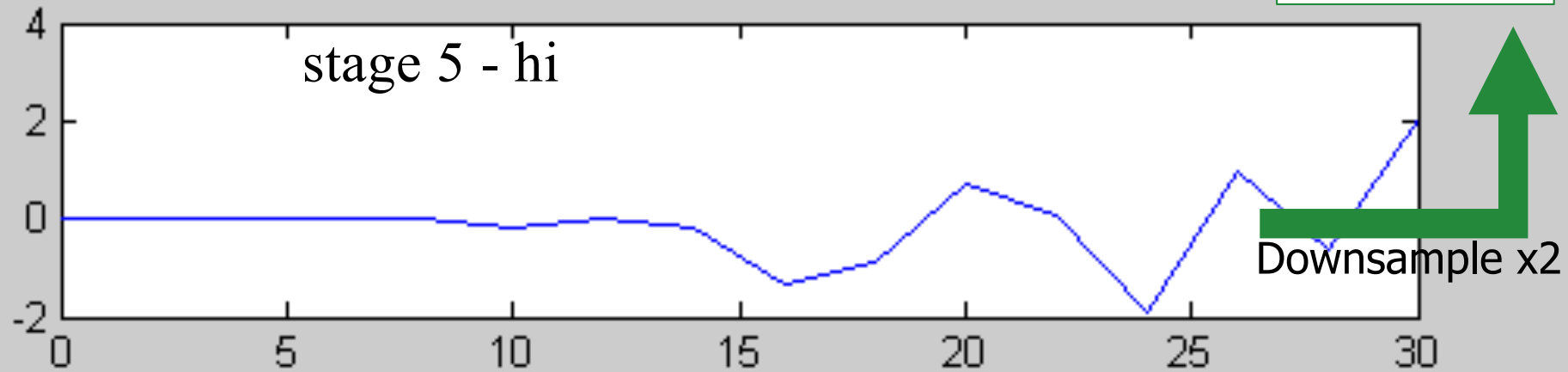
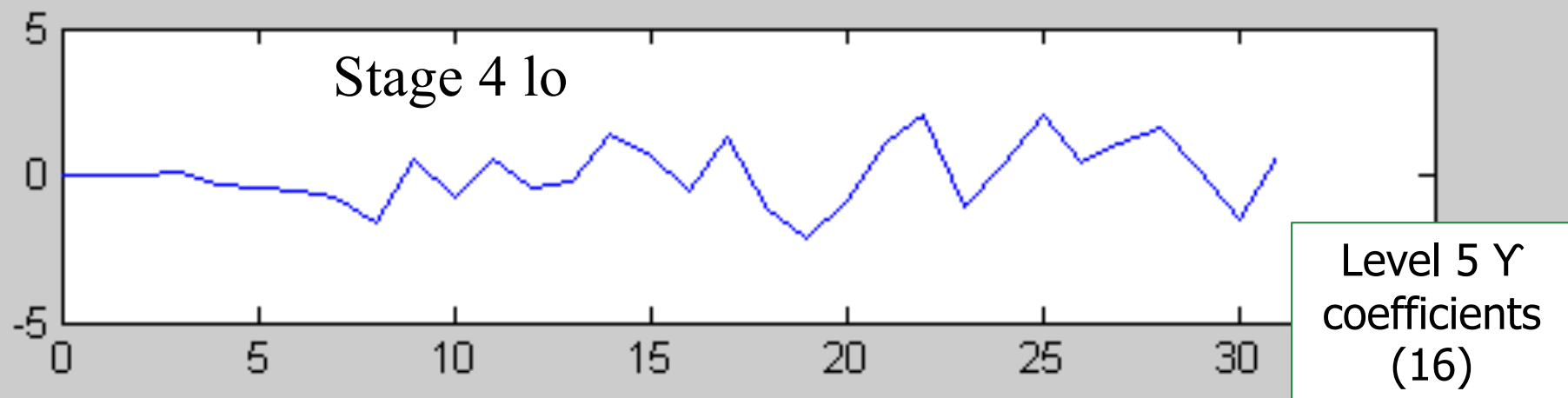


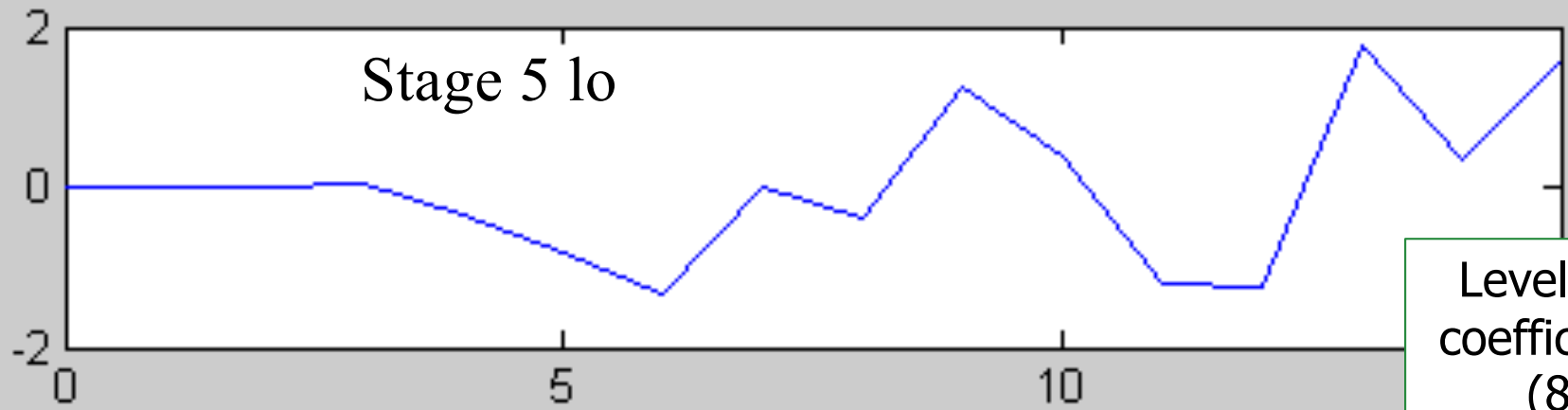
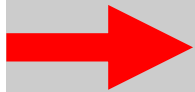




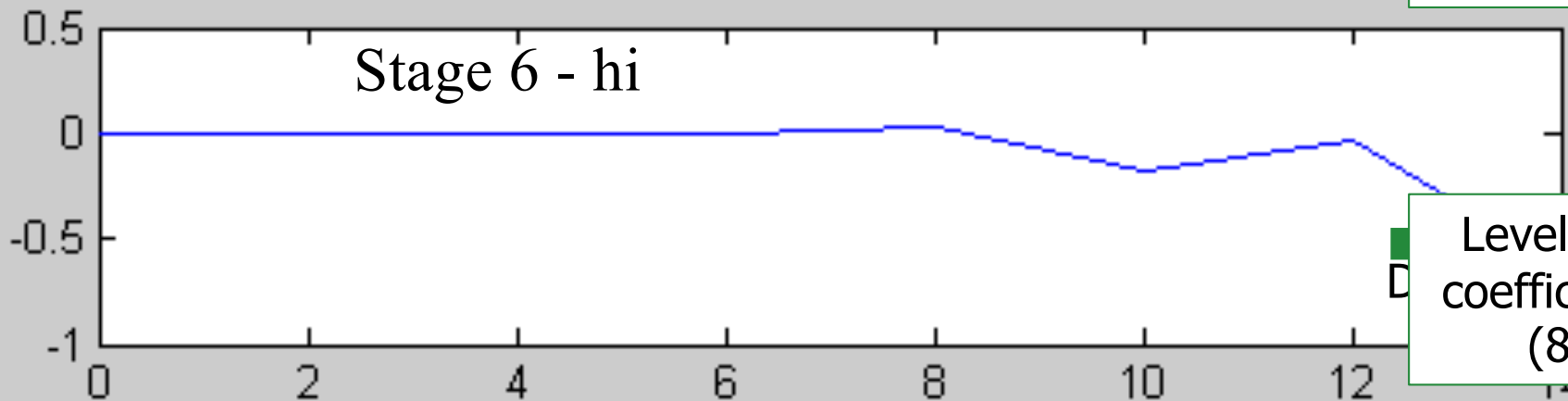




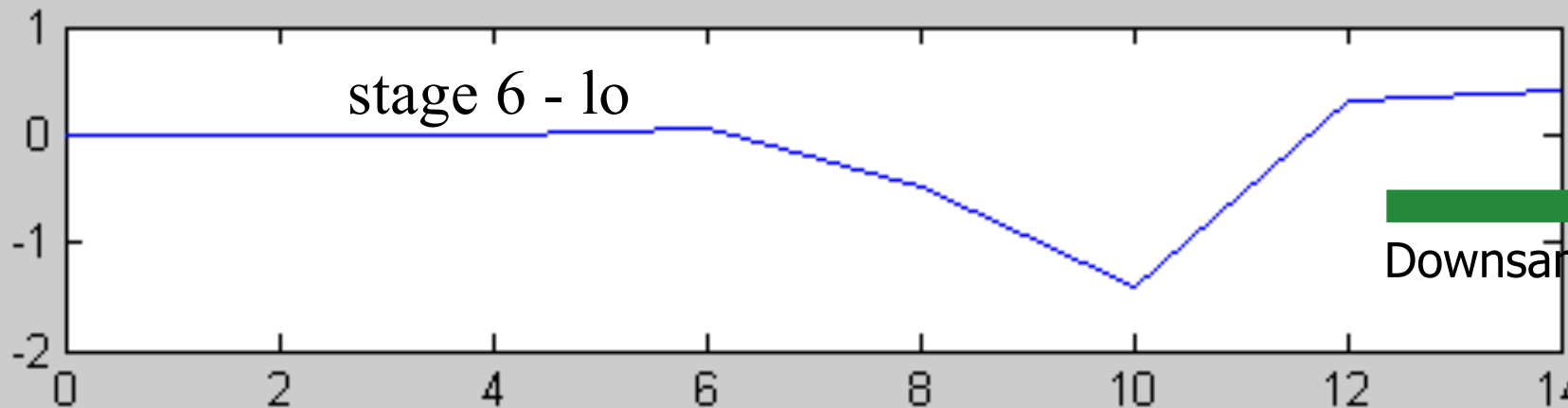




Level 6 Y
coefficients
(8)



Level 7 Y
coefficients
(8)



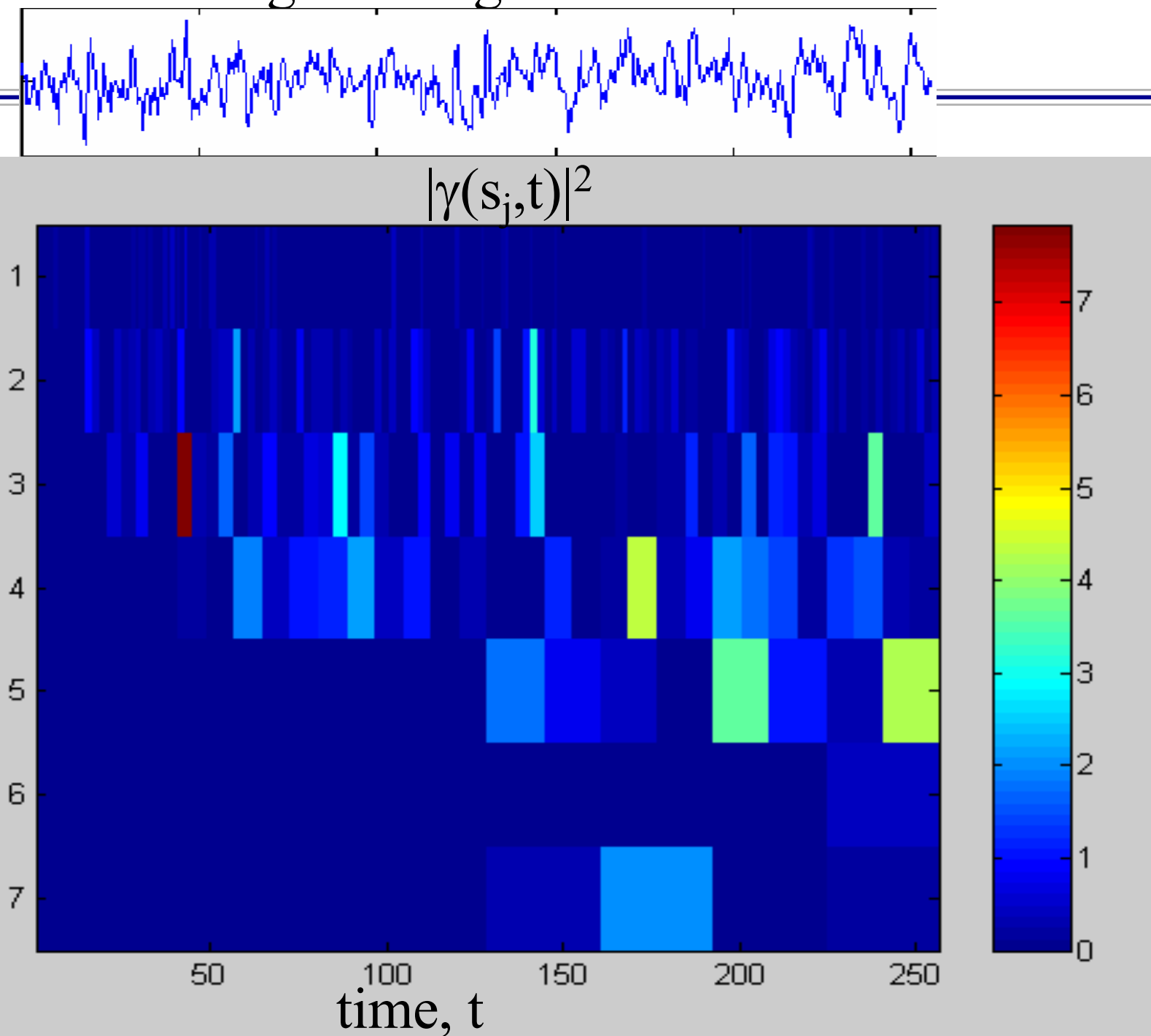
Downsample x2

Putting it all together ...

short
wavelengths

scale

long
wavelengths

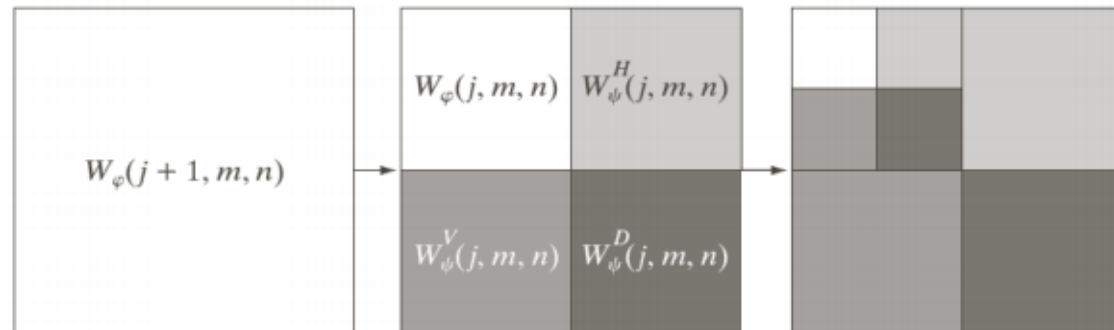
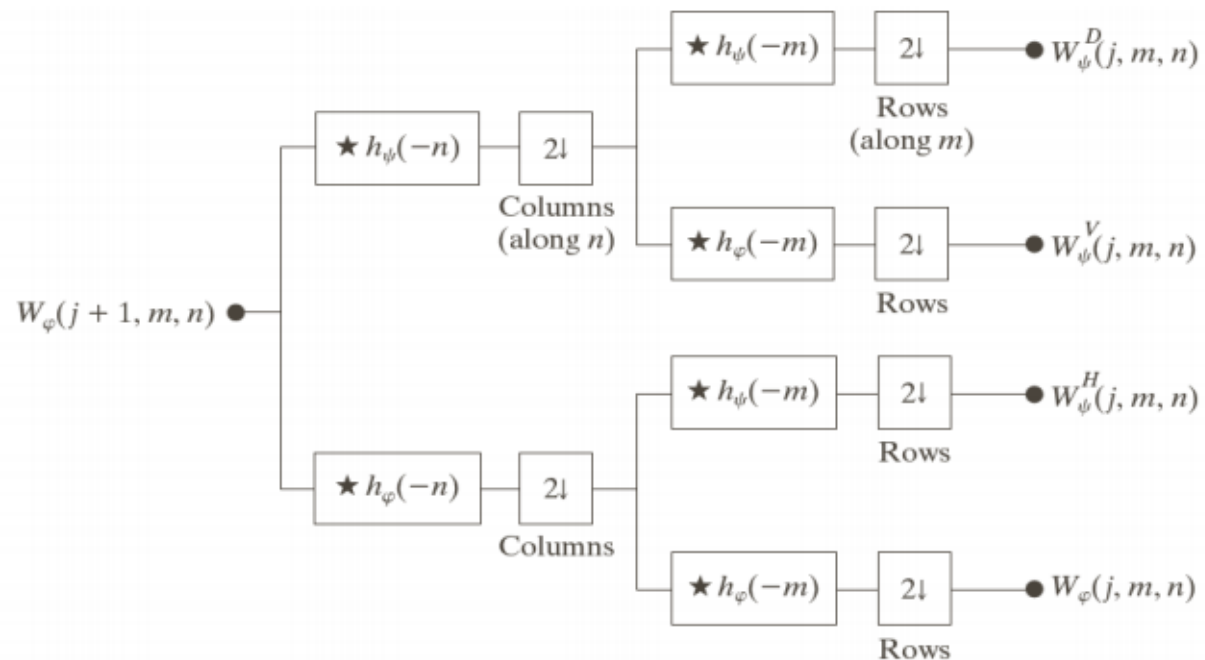




Expanding to Two Dimensions

- In two dimensions, a 2D scaling function $\phi(x, y)$ and 3 2D wavelet functions $\psi^H(x, y)$, $\psi^V(x, y)$, $\psi^D(x, y)$ are required
- We can create these from the 1D scaling and wavelet functions:
 - $\phi(x, y) = \phi(x)\phi(y)$
 - $\psi^H(x, y) = \psi(x)\phi(y)$
 - $\psi^V(x, y) = \phi(x)\psi(y)$
 - $\psi^D(x, y) = \psi(x)\psi(y)$

Expanding to Two Dimensions



Expanding to Two Dimensions

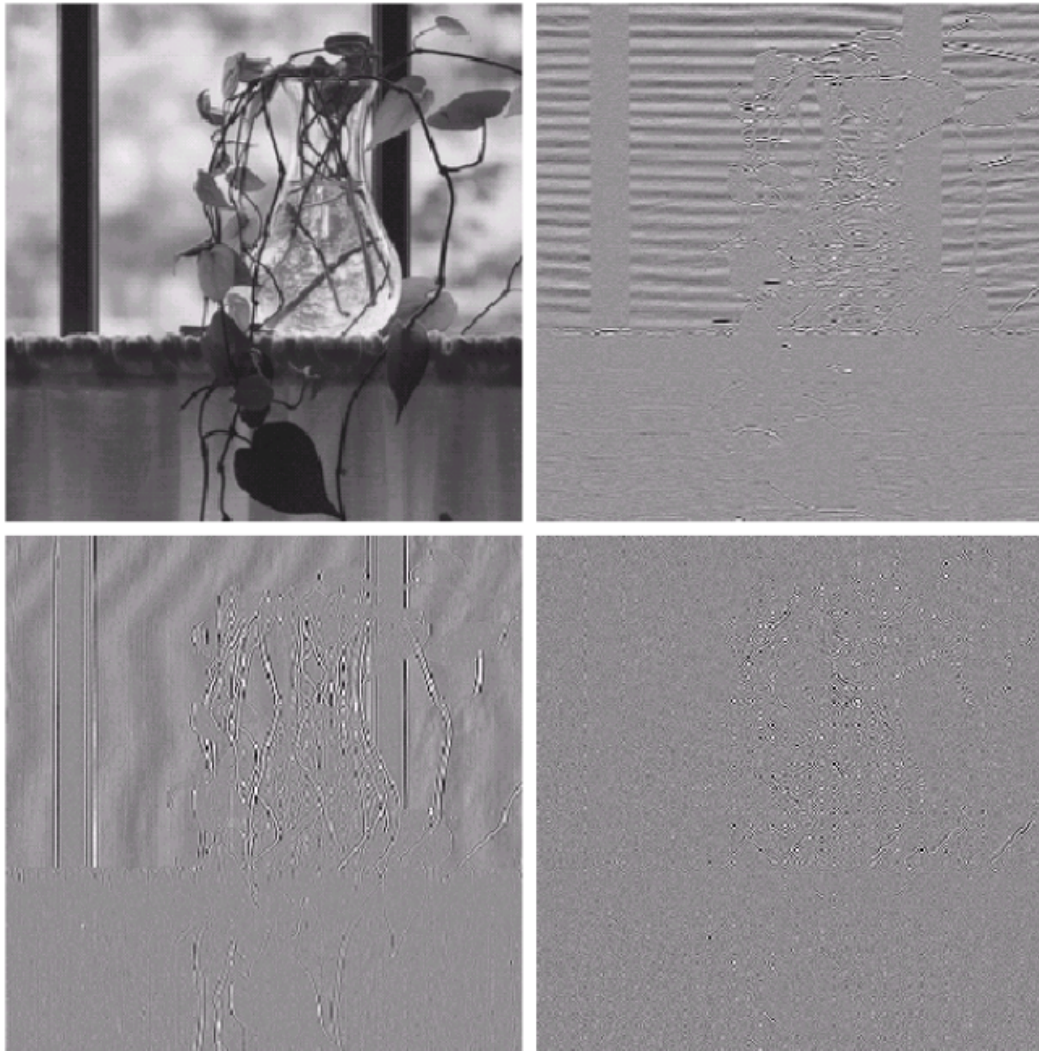


FIGURE 7.7 A four-band split of the vase in Fig. 7.1 using the subband coding system of Fig. 7.5.

a	d^V
d^H	d^D

$a(m,n)$:
approximation

$d^V(m,n)$: detail in
vertical

$d^H(m,n)$: detail in
horizontal

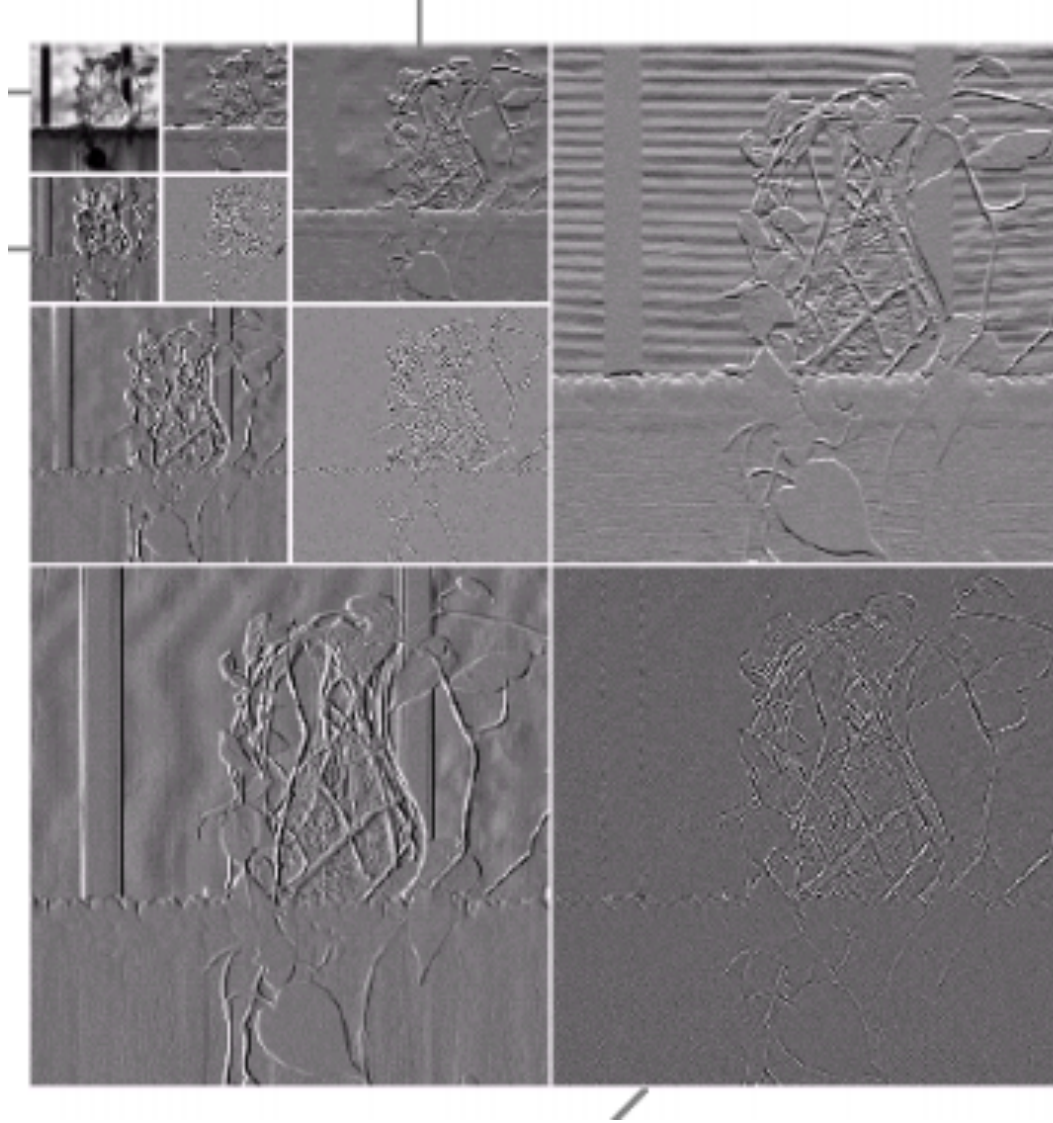
$d^D(m,n)$: detail in
diagonal

Colorado School of Mines

Image and Multidimensional Signal Processing

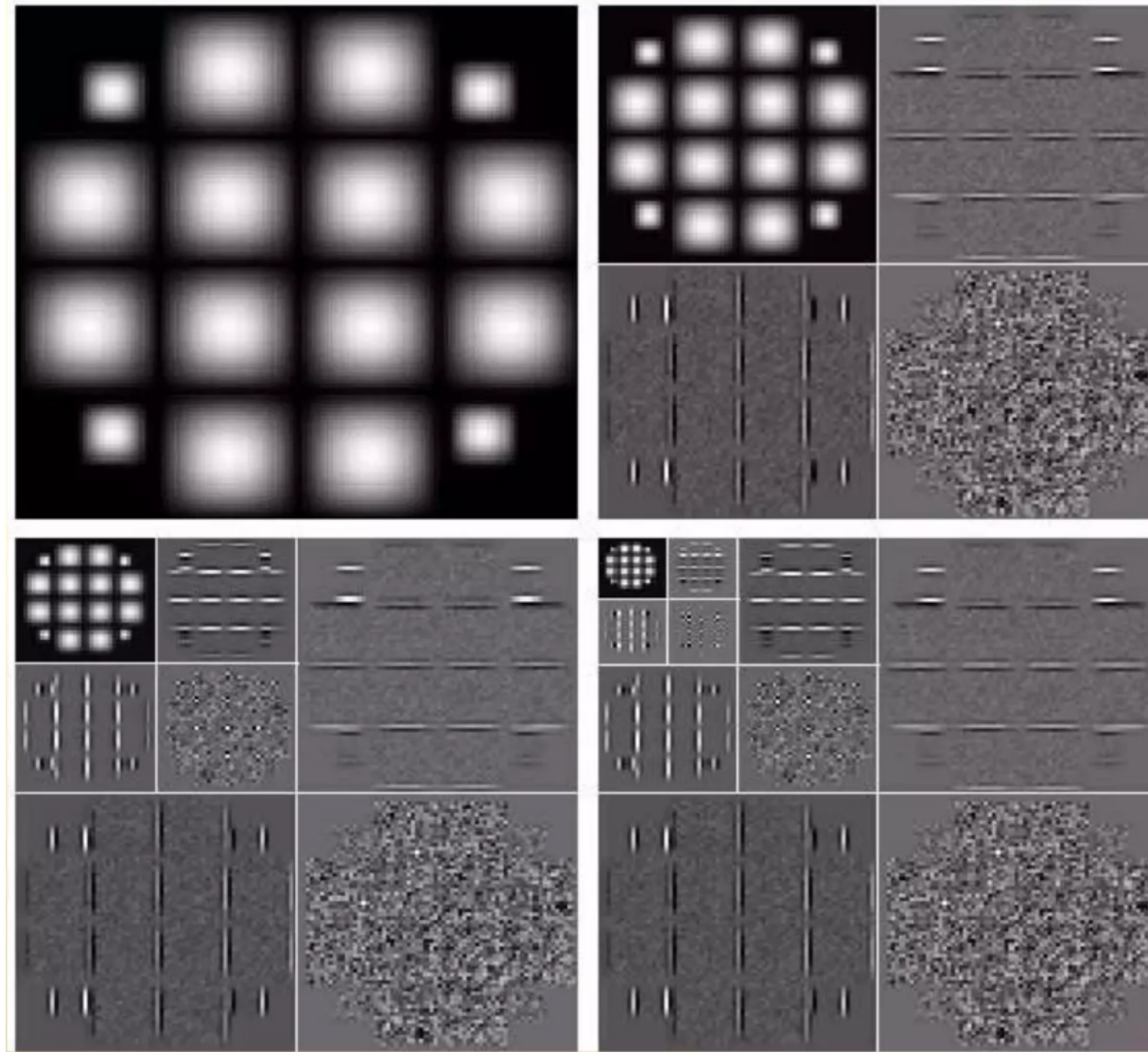


Expanding to Two Dimensions





Expanding to Two Dimensions





Big Ideas

❑ Wavelet transform

- Capture temporal data with fewer coefficients than STFT
- Use scaling and translation to get different resolution at different levels



Admin

- ❑ Project 2
 - Due 4/26
- ❑ Final Exam – 5/1
 - Covers lec 1-23*
 - Doesn't include lecture 12 (Data converters and noise shaping)
 - All old exams online
 - Disclaimers: old exams had different coverage for different years



Admin

- ❑ TA Office hour schedule for Zhihan and Jiyue
 - <https://edstem.org/us/courses/33619/discussion/2954005>
 - 18th (Tue):
 - Jiyue He, 10-noon
 - Jiyue He, 7-9 pm
 - 19th (Wed):
 - Jiyue He, 7-9 pm
 - 24th (Mon):
 - Zhihan Xu, 10-11:30 am
 - 25th (Tue):
 - Zhihan Xu, 10-11:30 am
 - Zhihan Xu, 7-8:30 pm
- ❑ Shuang and my office hours as usual