

# ESE 5310: Digital Signal Processing

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Lecture 7: February 2, 2023  
Sampling and Reconstruction



# Lecture Outline

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- ❑ Sampling
  - Frequency Response of Sampled Signal
- ❑ Reconstruction
- ❑ Anti-aliasing Filter



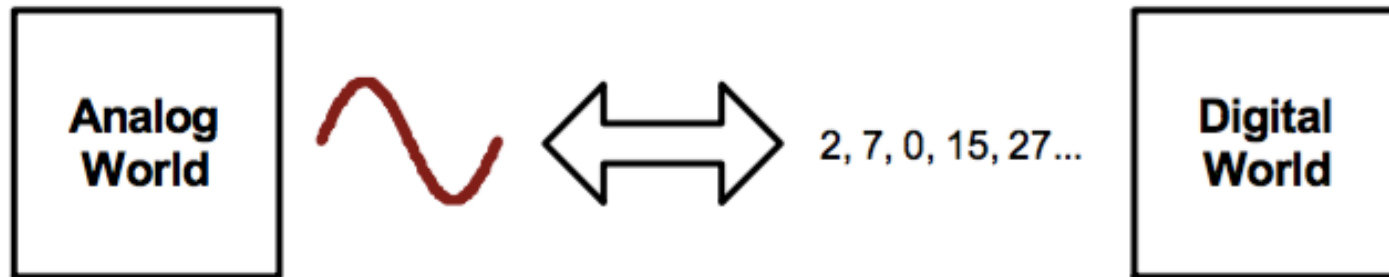
# Video Example

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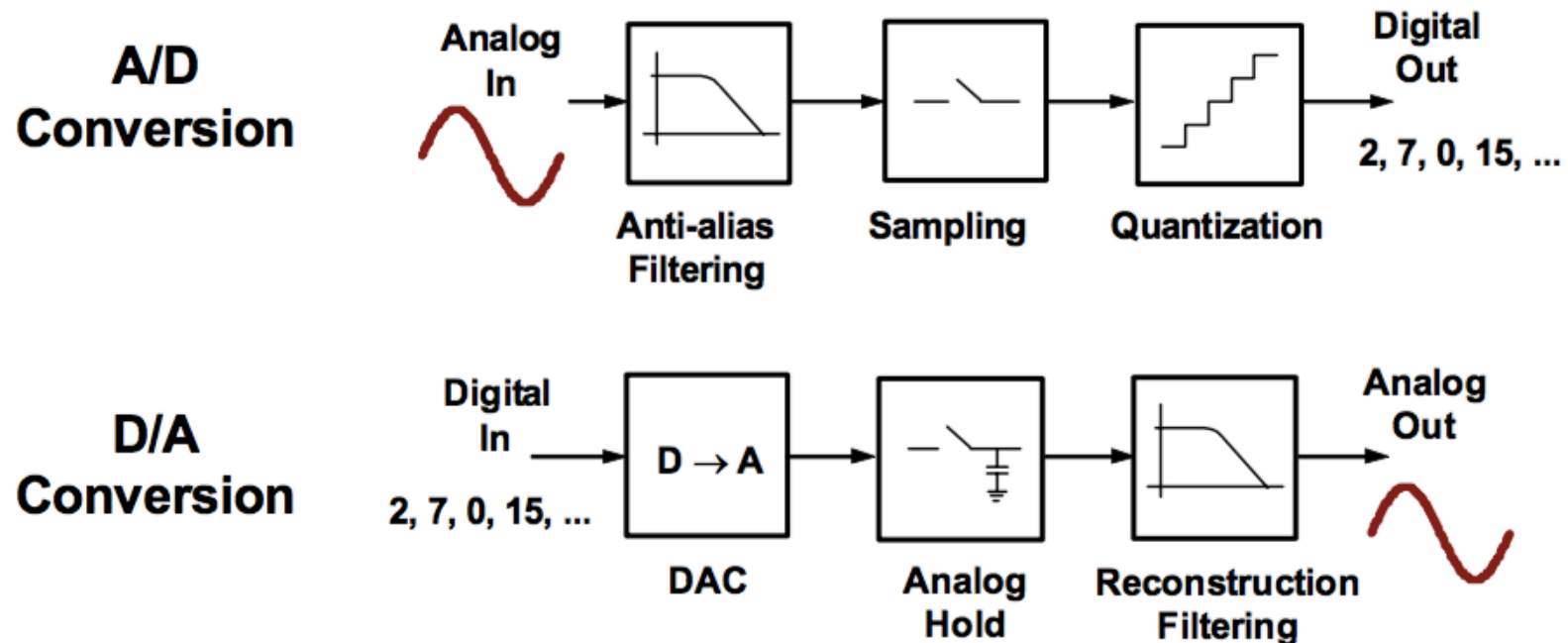
❑ <https://www.youtube.com/watch?v=ByTsISFXUoY>

# The Data Conversion Problem



- ❑ Real world signals
  - Continuous time, continuous amplitude
- ❑ Digital abstraction
  - Discrete time, discrete amplitude
- ❑ Two problems
  - How to go discretize in time and amplitude
    - A/D conversion
  - How to "undescretize" in time and amplitude
    - D/A conversion

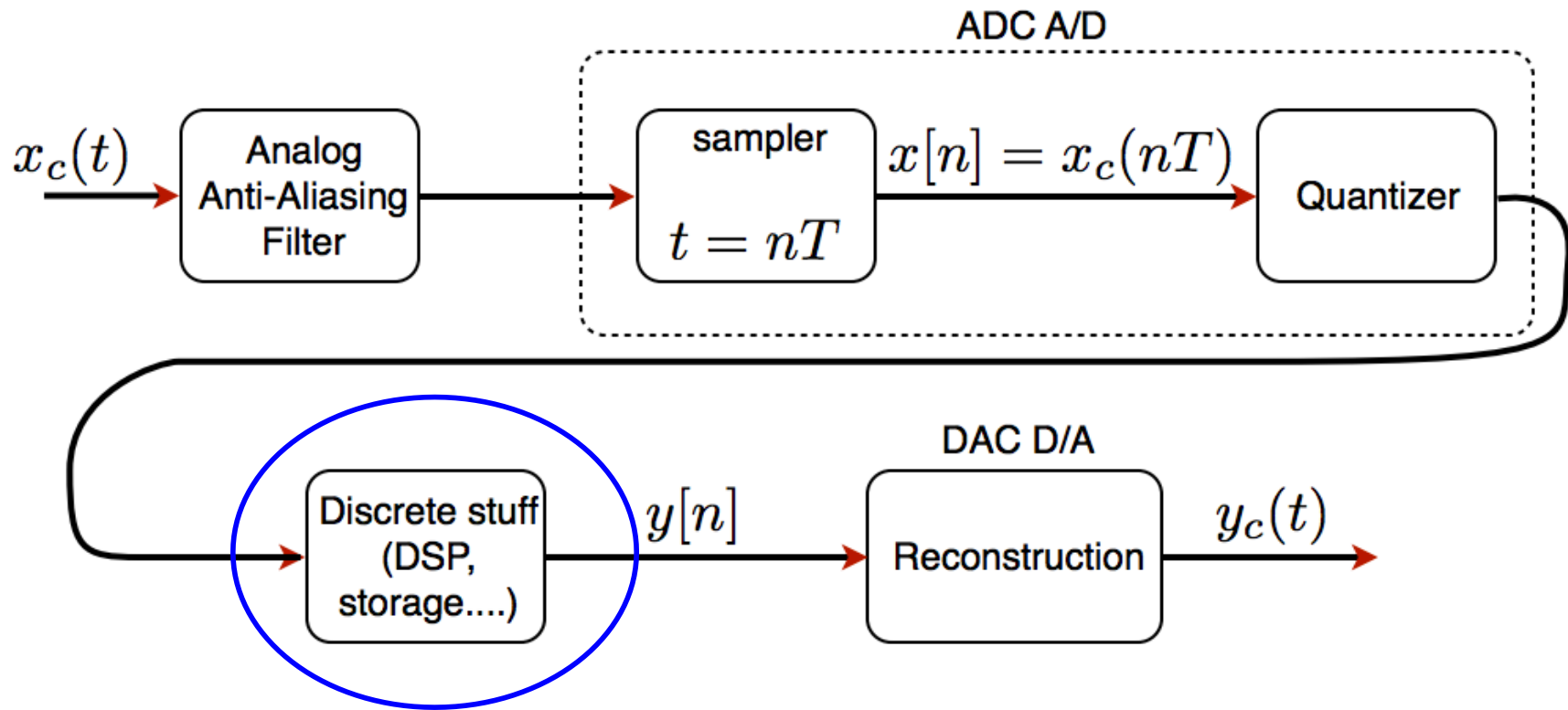
# Overview



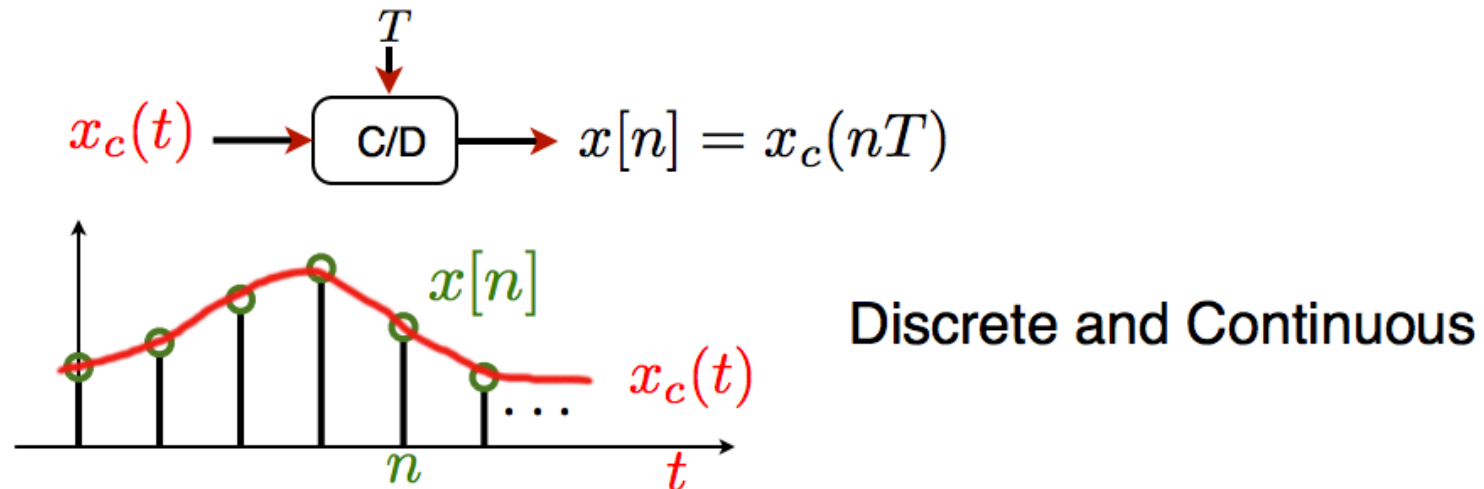
- ❑ We'll first look at these building blocks from a functional, "black box" perspective



# DSP System

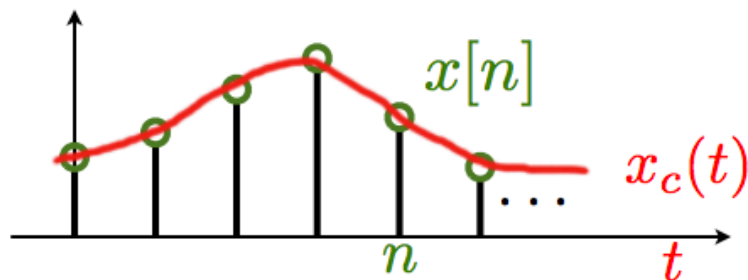
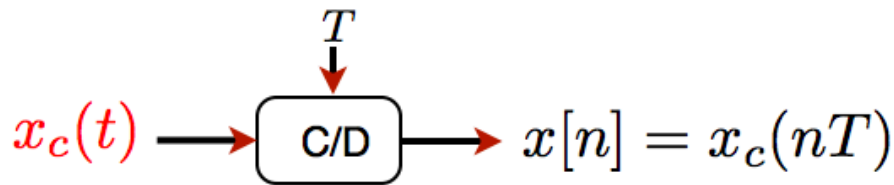


# Ideal Sampling Model



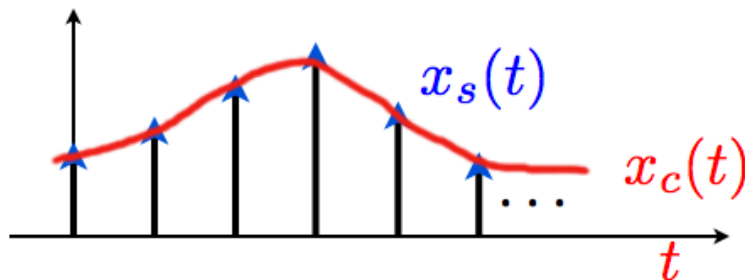
- ❑ Ideal continuous-to-discrete time (C/D) converter
  - $T$  is the sampling period
  - $f_s = 1/T$  is the sampling frequency
  - $\Omega_s = 2\pi/T$

# Ideal Sampling Model



Discrete and Continuous

define impulsive sampling:

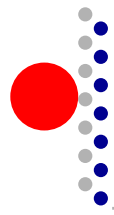


Continuous

$$x_s(t) = \cdots + x_c(0)\delta(t) + x_c(T)\delta(t - T) + \cdots$$

$$x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$





# Ideal Sampling Model

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$$x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

- Dirac delta function,  $\delta(t)$ 
  - Infinitely high and thin, area of 1
  - Not physical—for modeling

$$x[n] \leftrightarrow x_s(t) \leftrightarrow x_c(t)$$

- Three signals. How are they related? In time? In frequency?



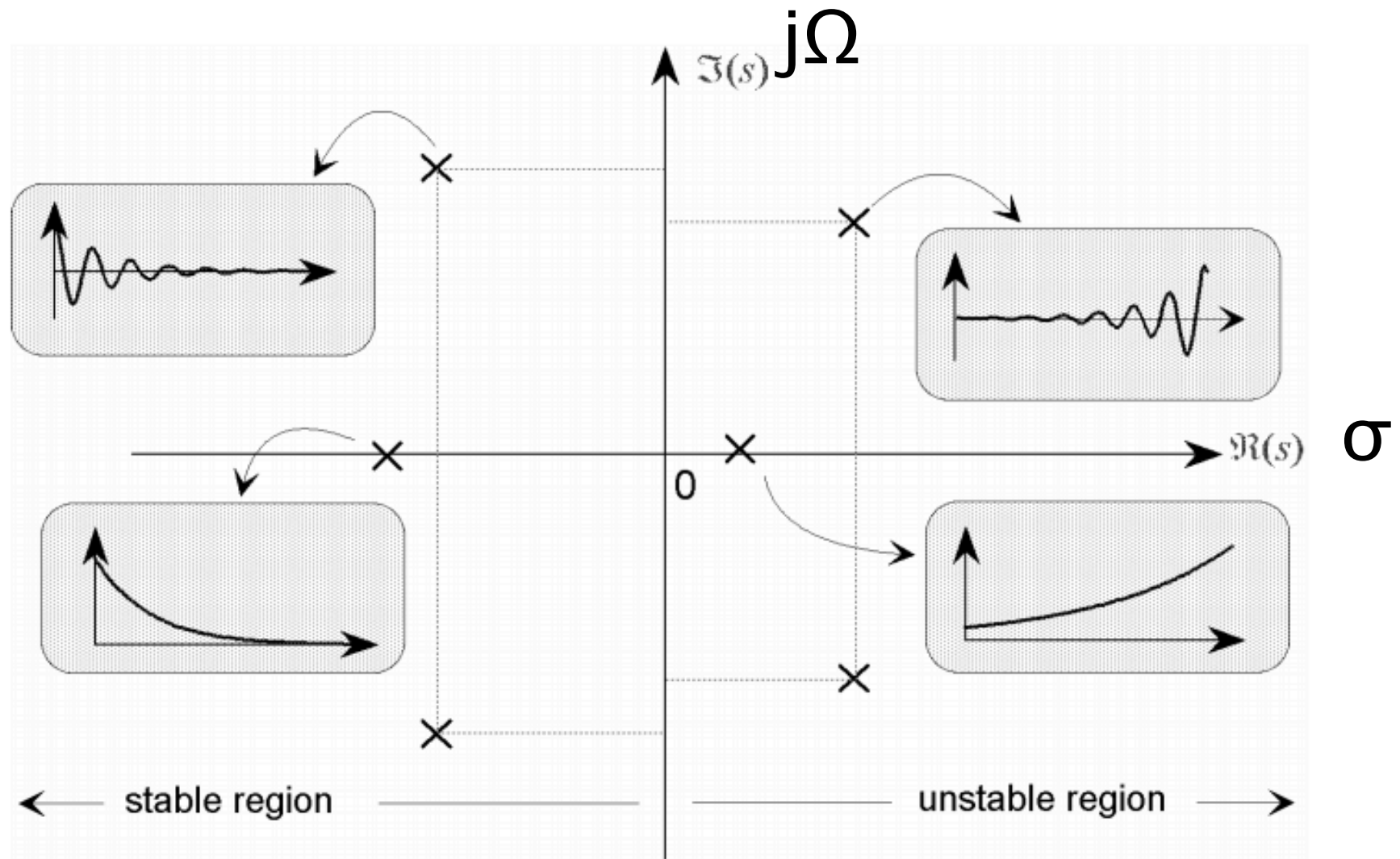
# Reminder: Laplace Transform

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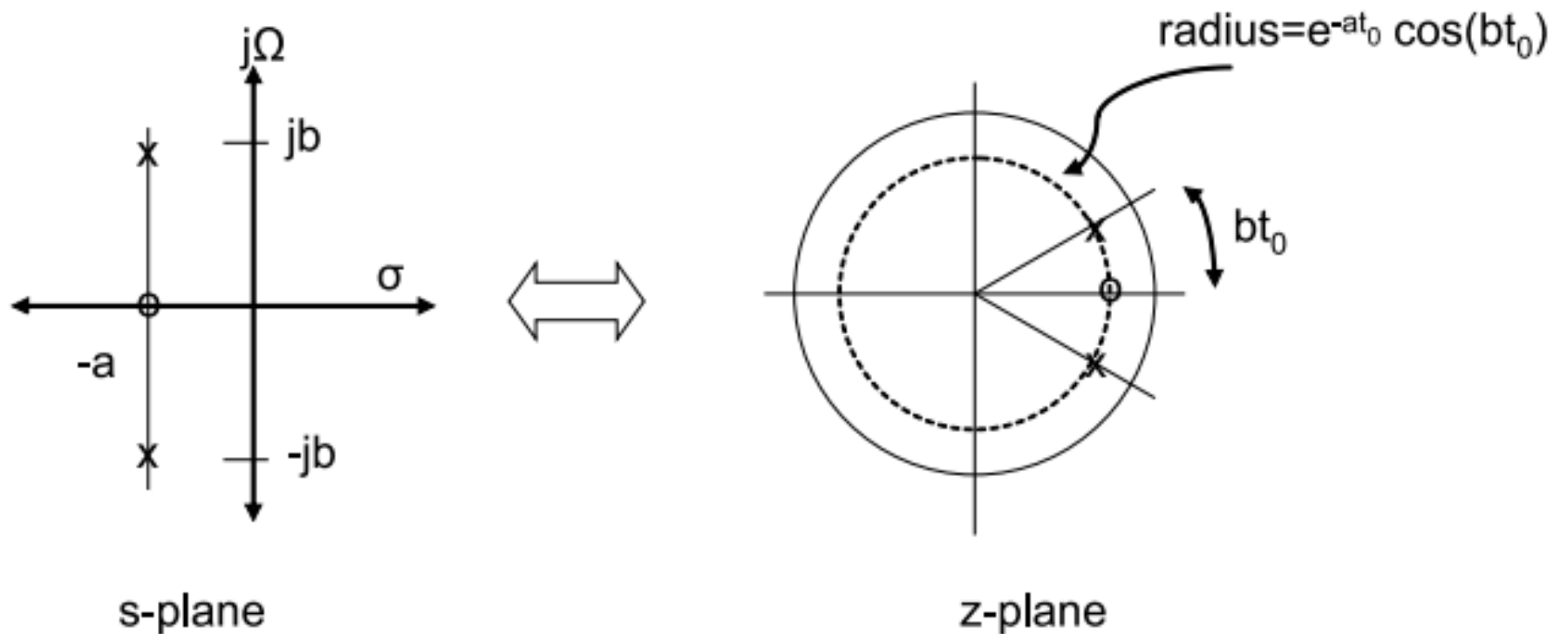
- ❑ The Laplace transform takes a function of time,  $t$ , and transforms it to a function of a complex variable,  $s$ .
- ❑ Because the transform is invertible, no information is lost, and it is reasonable to think of a function  $f(t)$  and its Laplace transform  $F(s)$  as two views of the same phenomenon.
- ❑ Each view has its uses, and some features of the phenomenon are easier to understand in one view or the other.



# S-plane and stability



# S-Plane Mapping to Z-Plane



$t_0$  = sampling period



# Frequency Domain Analysis

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- How is  $x[n]$  related to  $x_s(t)$  in frequency domain?

$$x[n] = x_c(nT) \qquad x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

# Frequency Domain Analysis

- How is  $x[n]$  related to  $x_s(t)$  in frequency domain?

$$x[n] = x_c(nT) \qquad x_s(t) = x_c \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

$$x_s(t) \quad : \text{C.T.}$$

$$X_s(j\Omega) = \sum_n x_c(nT) e^{-j\Omega nT}$$

$$x[n] \quad : \text{D.T.}$$

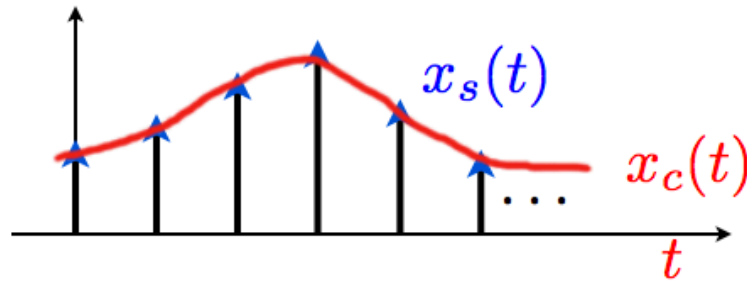
$$X(e^{j\omega}) = \sum_n x[n] e^{-j\omega n}$$


$$\omega = \Omega T$$

$$X(e^{j\omega}) = X_s(j\Omega)|_{\Omega=\omega/T}$$

$$X_s(j\Omega) = X(e^{j\omega})|_{\omega=\Omega T}$$

# Frequency Domain Analysis



Continuous

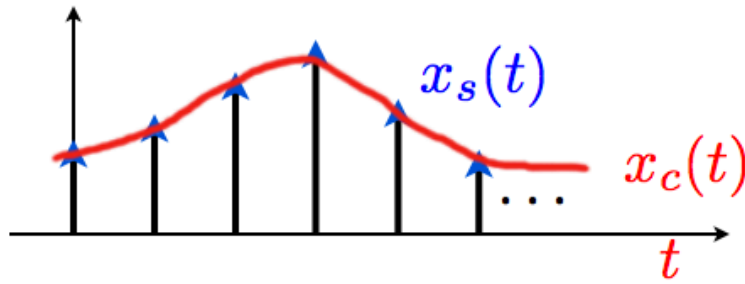
$$x_s(t) = \cdots + x_c(0)\delta(t) + x_c(T)\delta(t - T) + \cdots$$

$$x_s(t) = x_c \underbrace{\sum_{n=-\infty}^{\infty} \delta(t - nT)}_{\triangleq s(t)}$$

$$s(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$$



# Frequency Domain Analysis



Continuous

$$x_s(t) = \cdots + x_c(0)\delta(t) + x_c(T)\delta(t - T) + \cdots$$

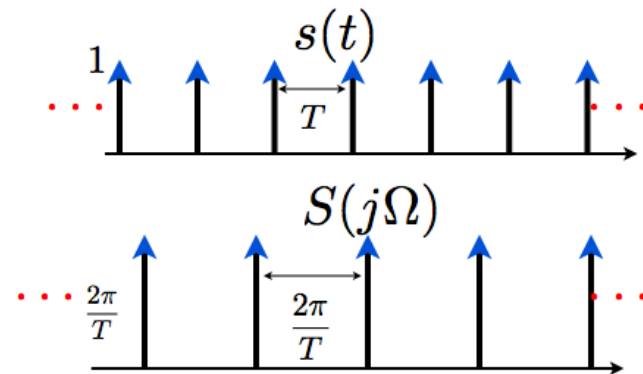
$$x_s(t) = x_c \underbrace{\sum_{n=-\infty}^{\infty} \delta(t - nT)}_{\triangleq s(t)}$$

$$s(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

$$s(t) \leftrightarrow S(j\Omega)$$

$$S(j\Omega) = \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(\Omega - \frac{2\pi}{T}k)$$

$\frac{2\pi}{T} \equiv \Omega_s$







# Frequency Domain Analysis

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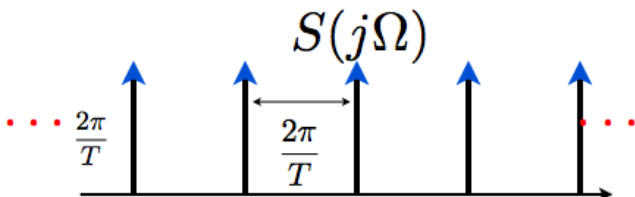
$$x_s(t) = s(t) \cdot x_c(t)$$



# Frequency Domain Analysis

$$x_s(t) = s(t) \cdot x_c(t)$$

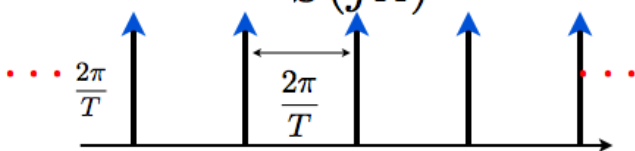
$$X_s(j\Omega) = \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega)$$

$$S(j\Omega) = \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta(\Omega - \frac{2\pi}{T}k)$$


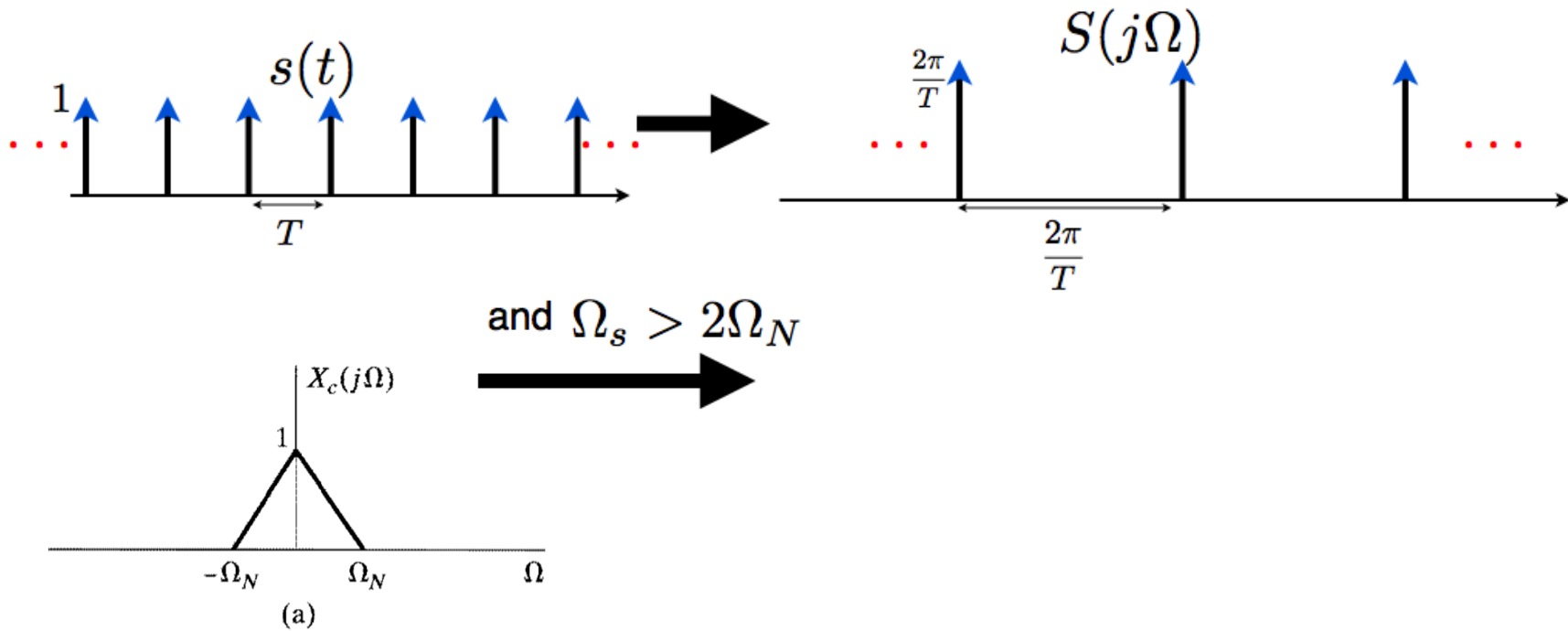
# Frequency Domain Analysis

$$x_s(t) = s(t) \cdot x_c(t)$$

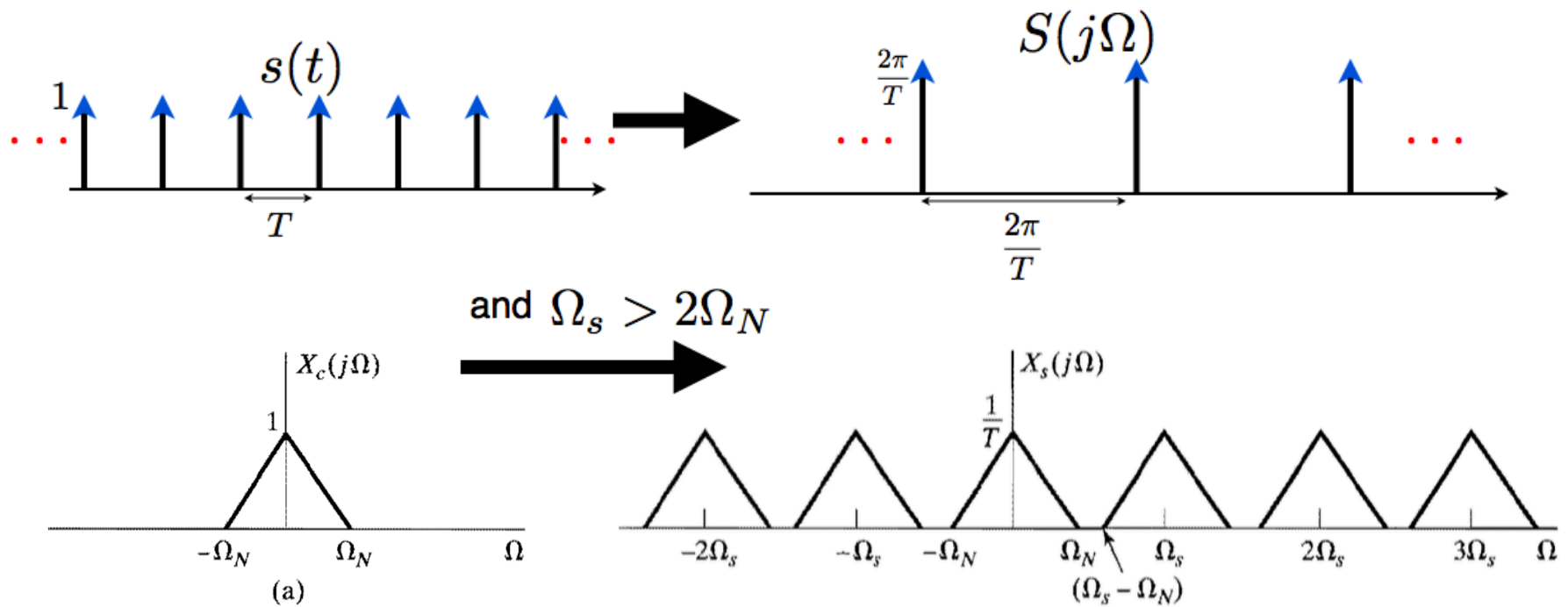
$$\begin{aligned} X_s(j\Omega) &= \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega) \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_s)) \quad | \quad \Omega_s = \frac{2\pi}{T} \end{aligned}$$

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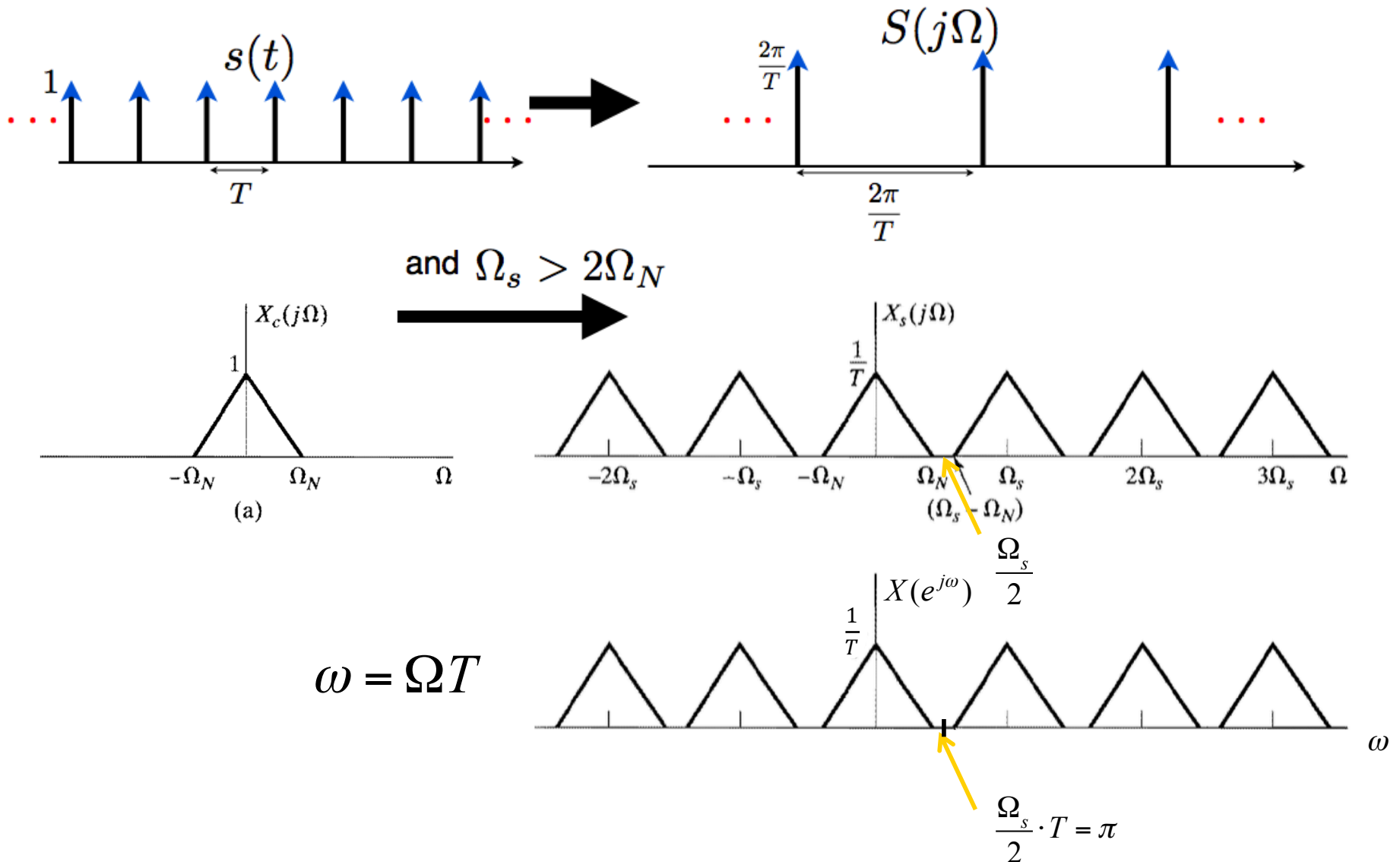
# Frequency Domain Analysis



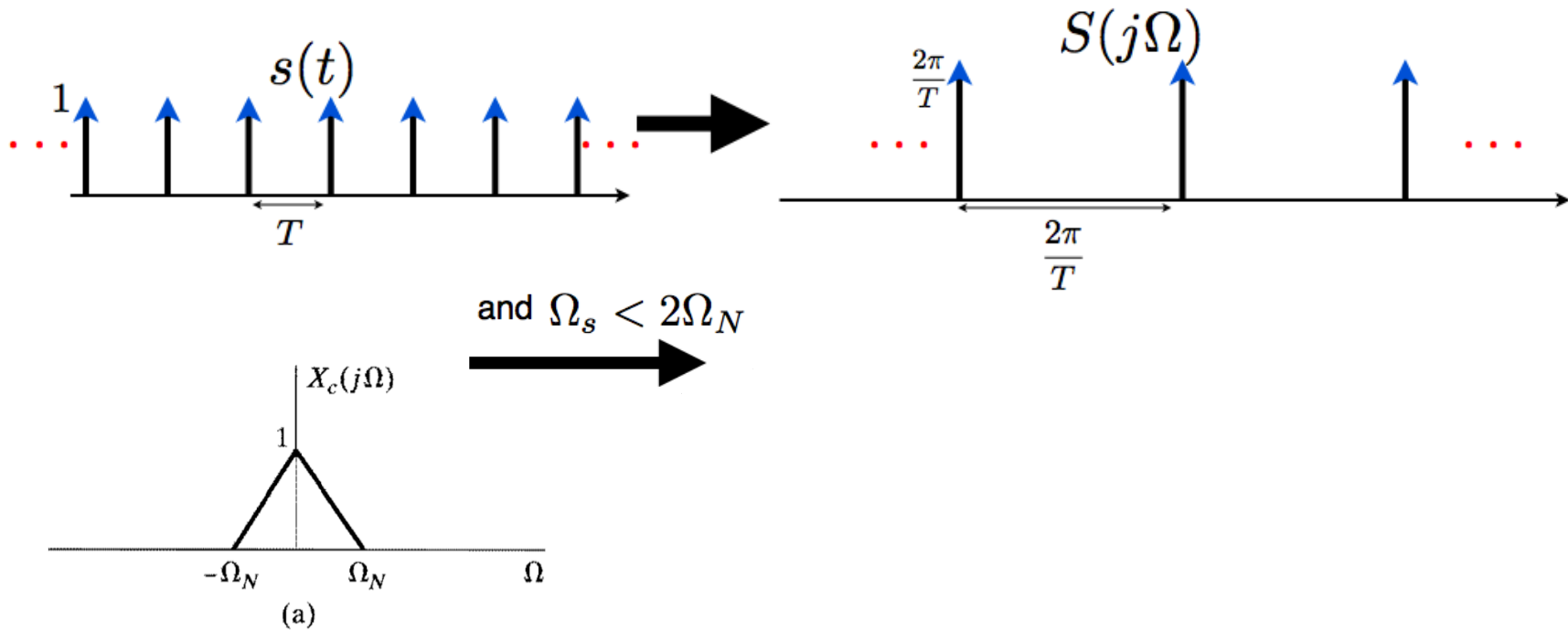
# Frequency Domain Analysis



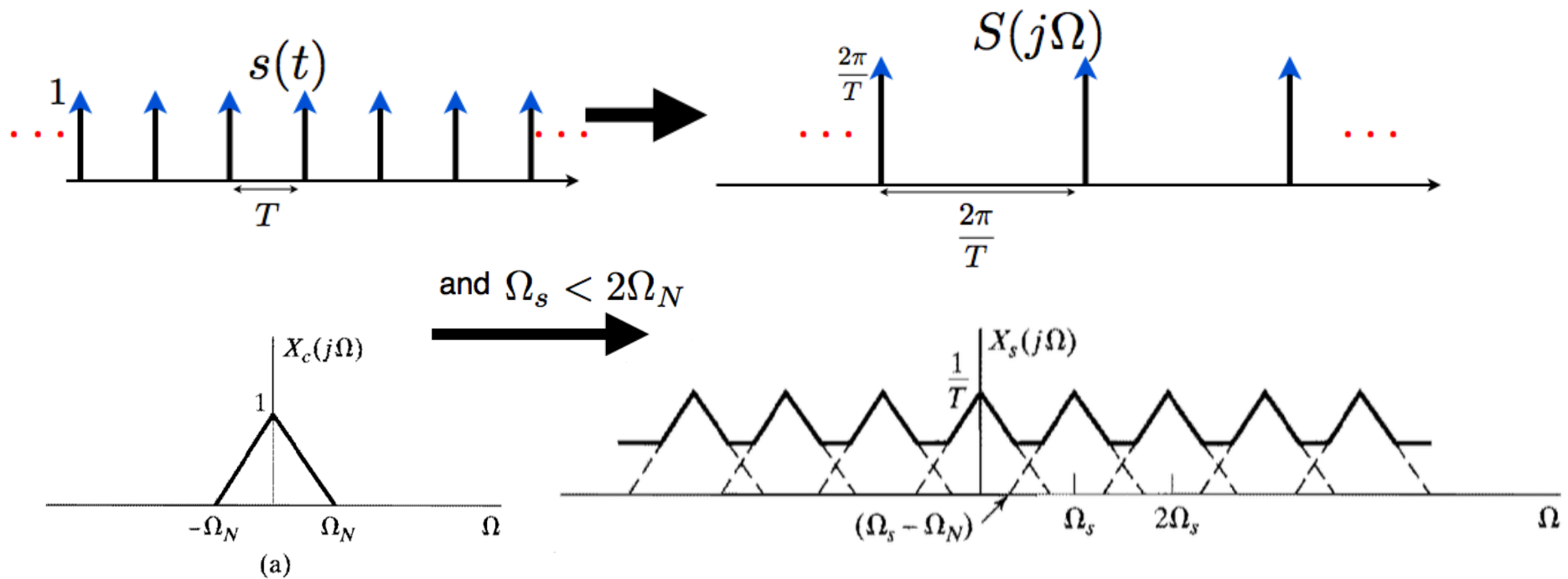
# Frequency Domain Analysis



# Frequency Domain Analysis w/ Aliasing

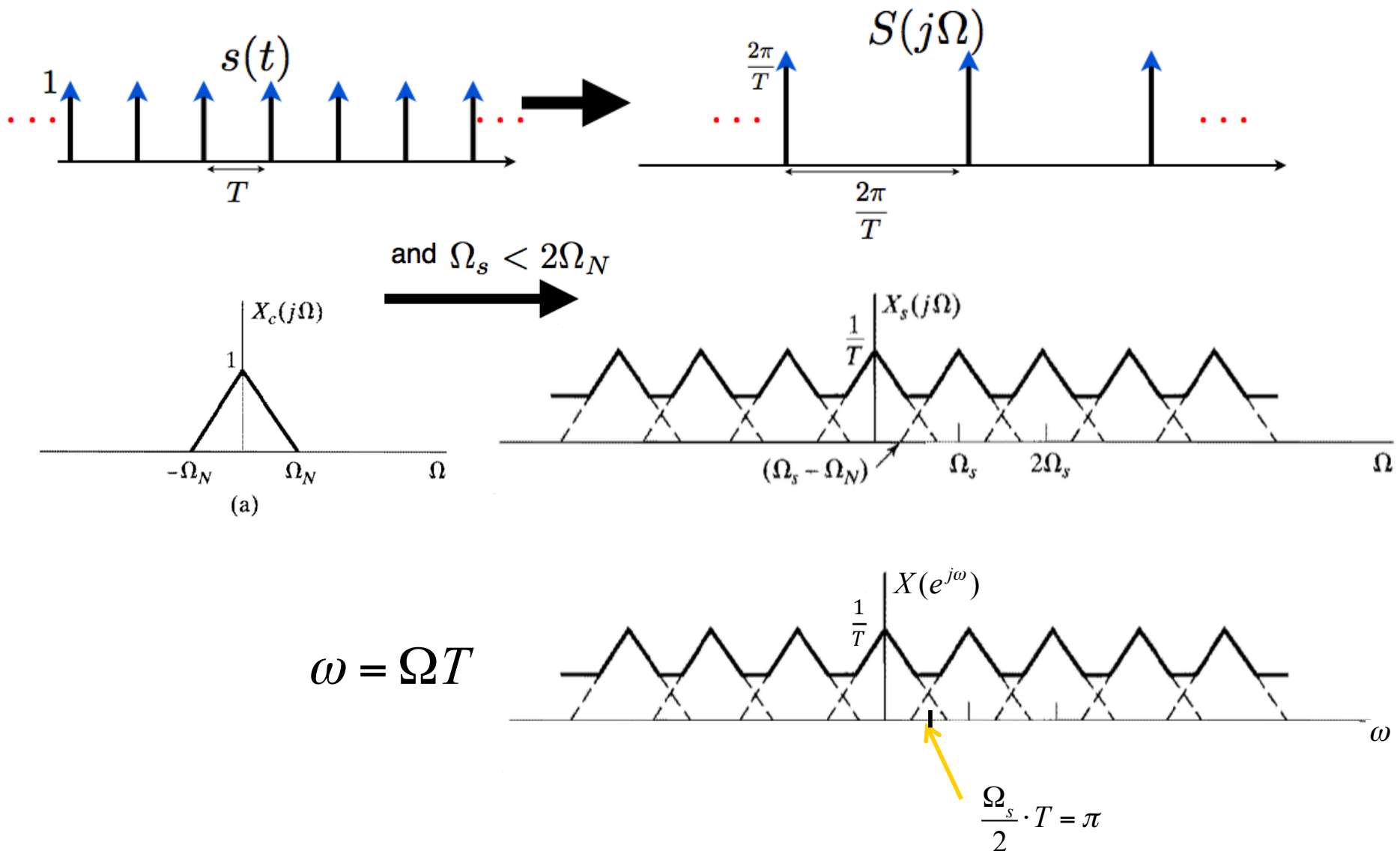


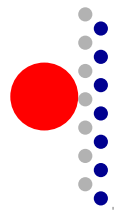
# Frequency Domain Analysis w/ Aliasing





# Frequency Domain Analysis w/ Aliasing

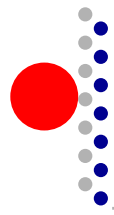




## Example: Cosine Input

- Sample the continuous-time signal  $x_c(t) = \cos(4000\pi t)$  with sampling period  $T = 1/6000$  s ( $f_s = 6$  kHz)

$$\begin{aligned} X_s(j\Omega) &= \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega) \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_s)) \quad | \quad \Omega_s = \frac{2\pi}{T} \end{aligned}$$



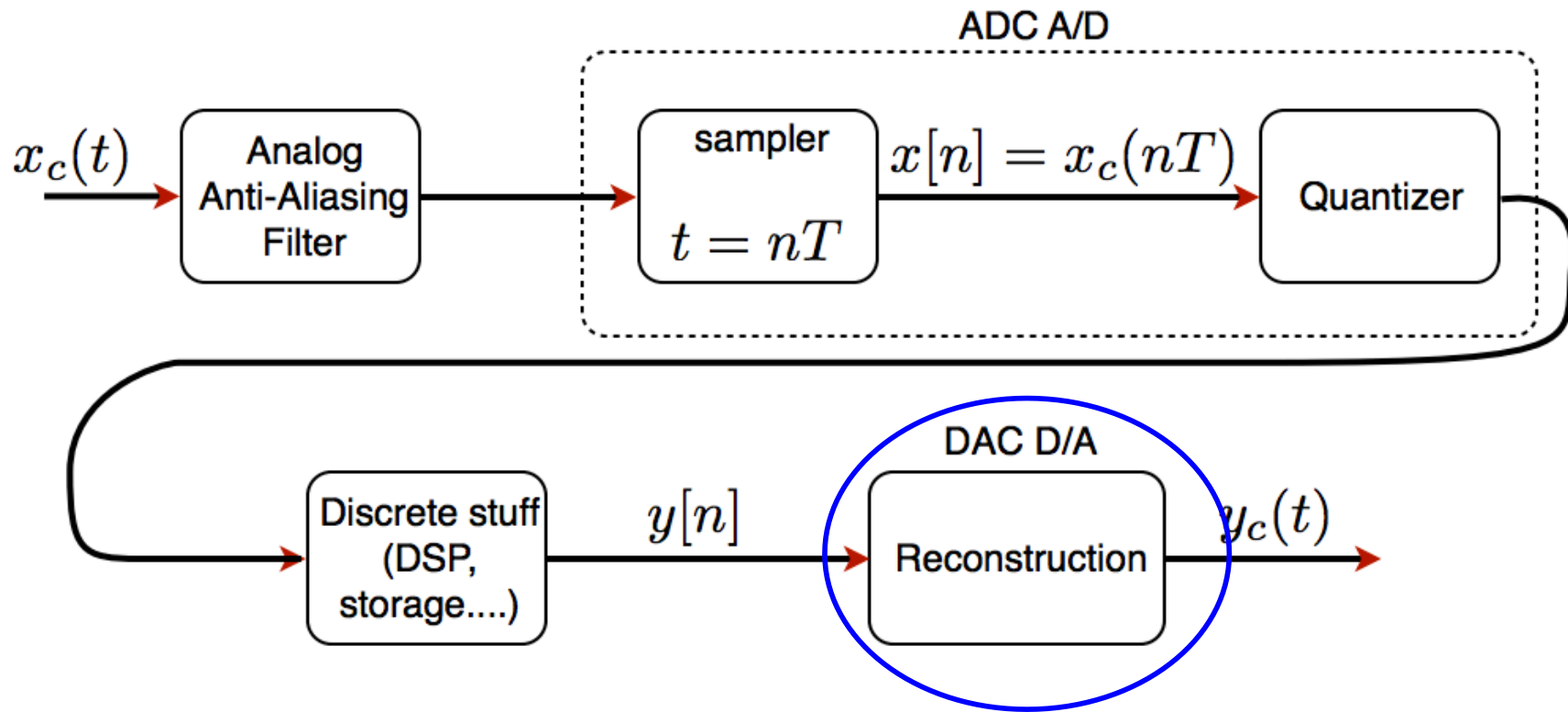
## Example: Cosine Input

- Sample the continuous-time signal  $x_c(t) = \cos(16000\pi t)$  with sampling period  $T = 1/6000$  s ( $f_s = 6$  kHz)

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# DSP System

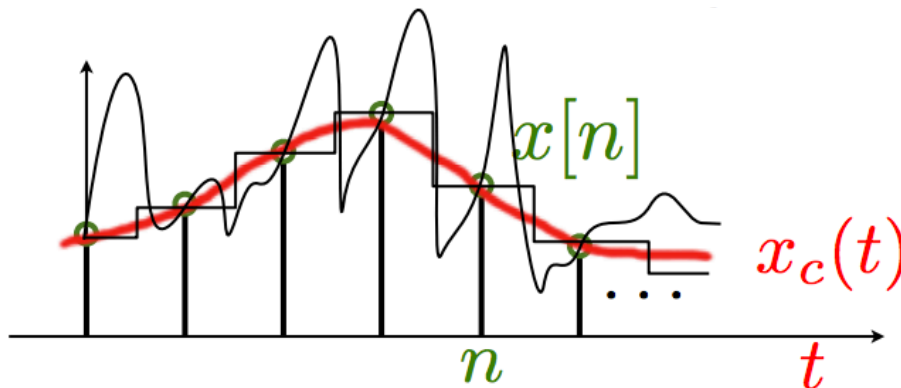


# Reconstruction of Bandlimited Signals

- Nyquist Sampling Theorem: Suppose  $x_c(t)$  is bandlimited. I.e.

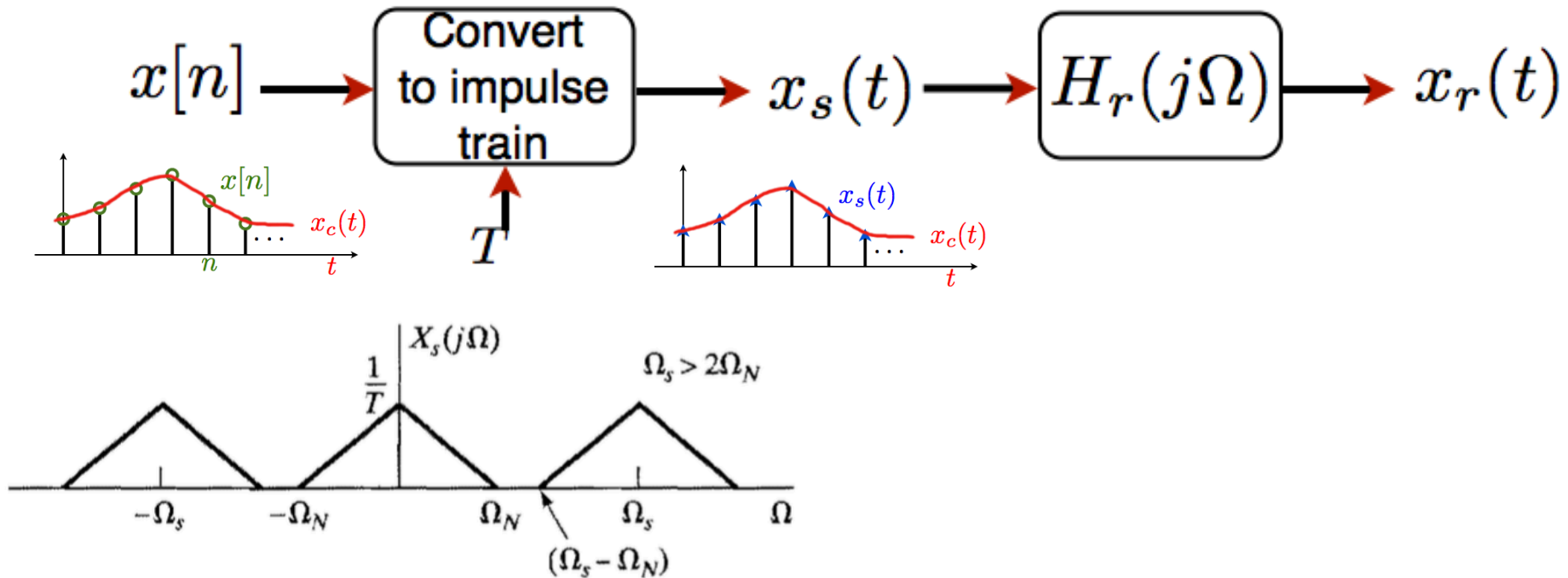
$$X_c(j\Omega) = 0 \quad \forall \quad |\Omega| \geq \Omega_N$$

- If  $\Omega_s \geq 2\Omega_N$ , then  $x_c(t)$  can be uniquely determined from its samples  $x[n] = x_c(nT)$
- Bandlimitedness is the key to uniqueness

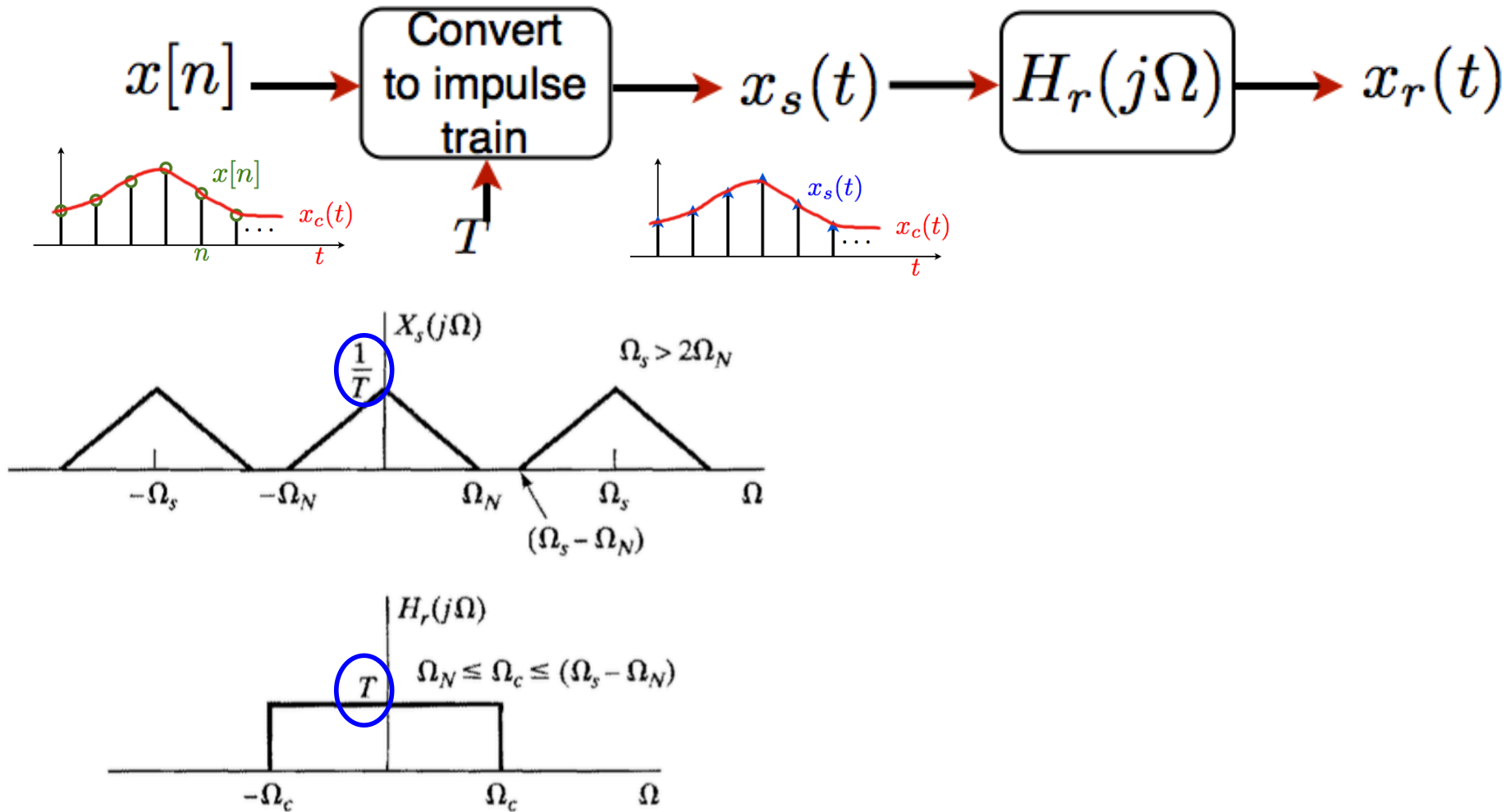


Multiple signals go through the samples, but only one is bandlimited within our sampling band

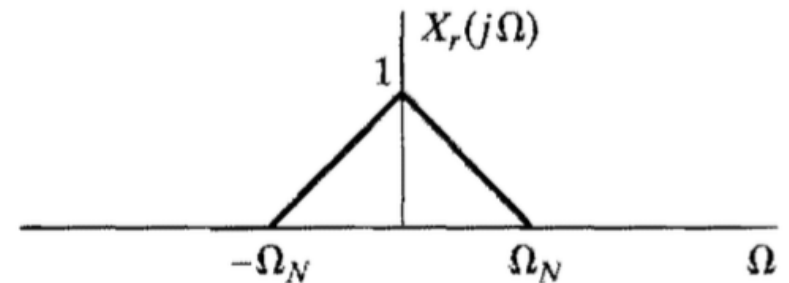
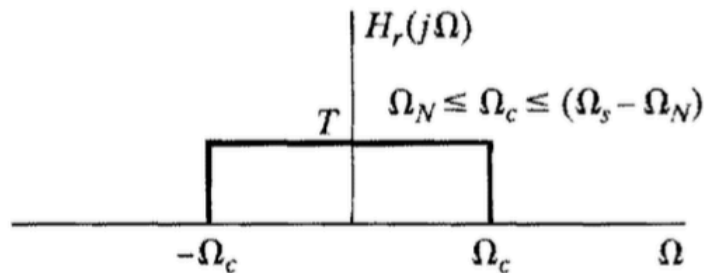
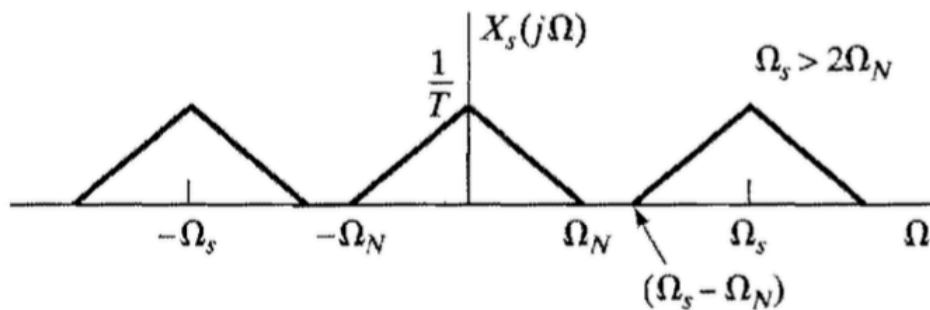
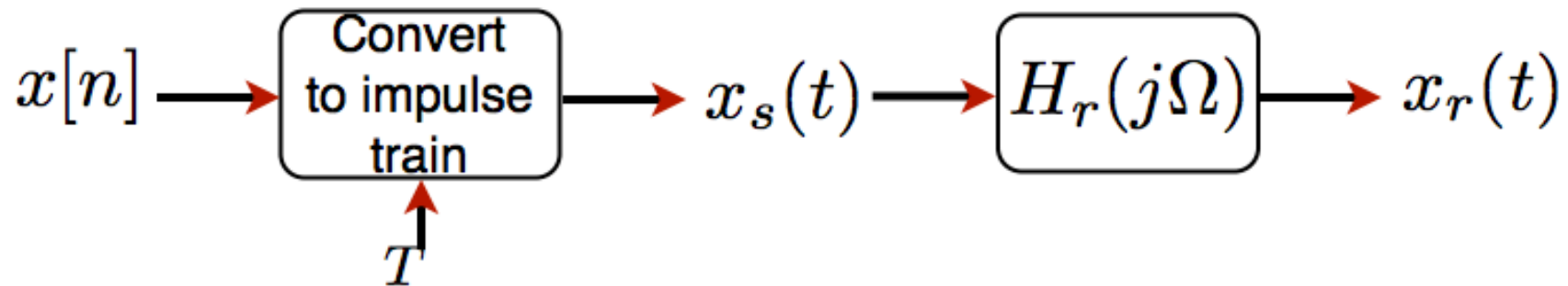
# Reconstruction in Frequency Domain



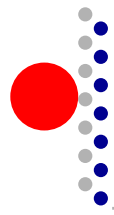
# Reconstruction in Frequency Domain



# Reconstruction in Frequency Domain







## Example: Cosine Input

- Sample and reconstruct the continuous-time signal  $x_c(t) = \cos(4000\pi t)$  with sampling period  $T = 1/6000$  s ( $f_s = 6$  kHz)

$$\begin{aligned} X_s(j\Omega) &= \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega) \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - k\Omega_s)) \quad | \quad \Omega_s = \frac{2\pi}{T} \end{aligned}$$

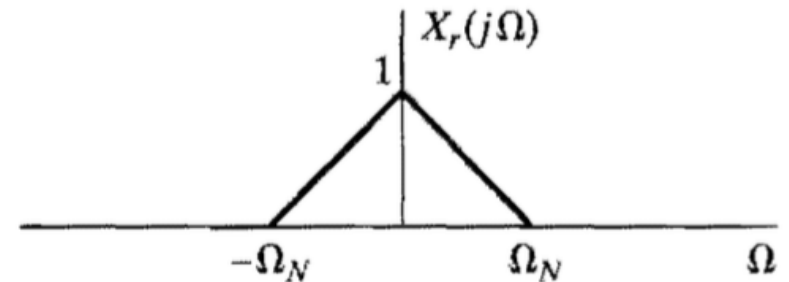
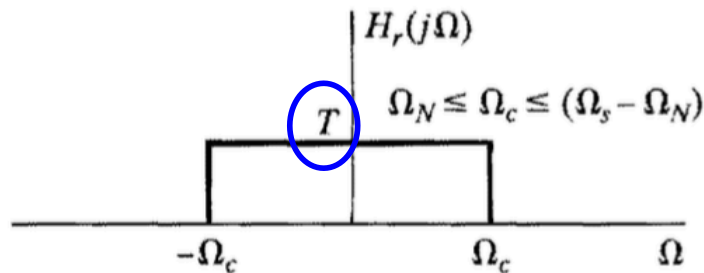
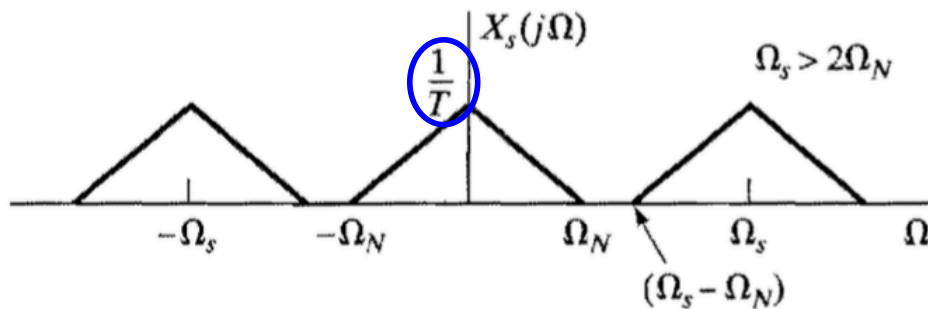
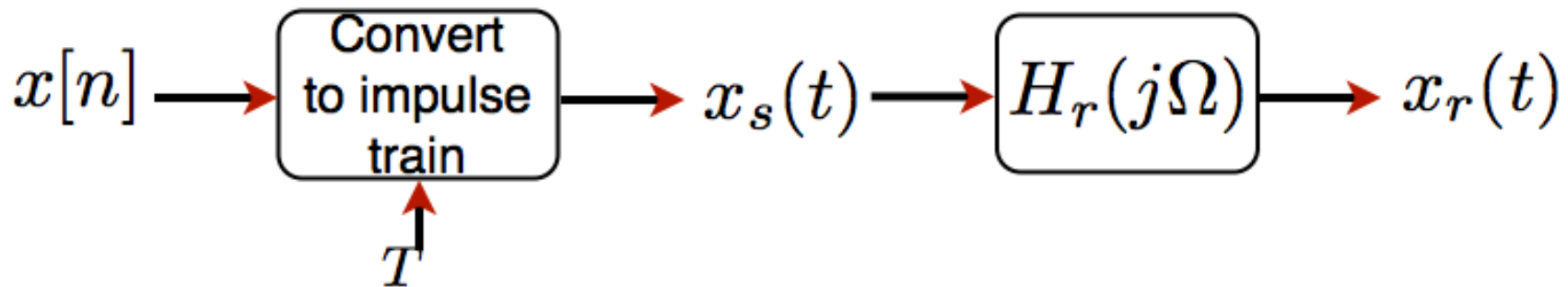


## Example: Cosine Input

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$$\begin{aligned} X_s(j\Omega) &= \frac{1}{2\pi} X_c(j\Omega) * S(j\Omega) \\ &= \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c(j(\Omega - \Omega_s)) \quad | \quad \Omega_s = \frac{2\pi}{T} \end{aligned}$$

# Reconstruction in Frequency Domain



$$2\Omega_N = \Omega_s$$

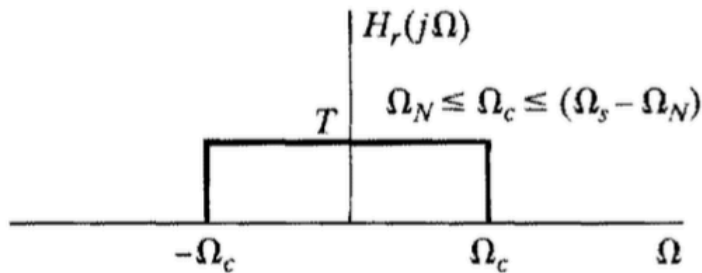
$$\Omega_N = \Omega_C = \Omega_s / 2$$

Sample at Nyquist  
and filter at signal  
bandwidth



# Reconstruction in Time Domain

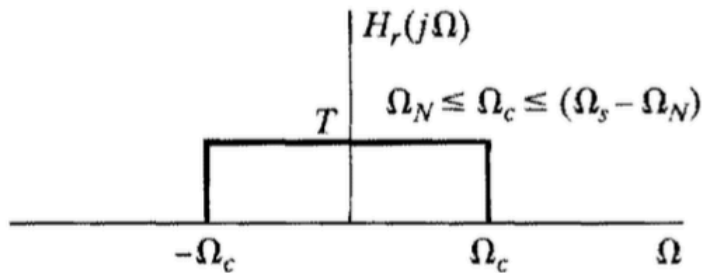
$$\Omega_N = \Omega_C = \Omega_s/2$$



$$h_r(t) = \frac{1}{2\pi} \int_{-\Omega_s/2}^{\Omega_s/2} T e^{j\Omega t} d\Omega$$

# Reconstruction in Time Domain

$$\Omega_N = \Omega_C = \Omega_s/2$$

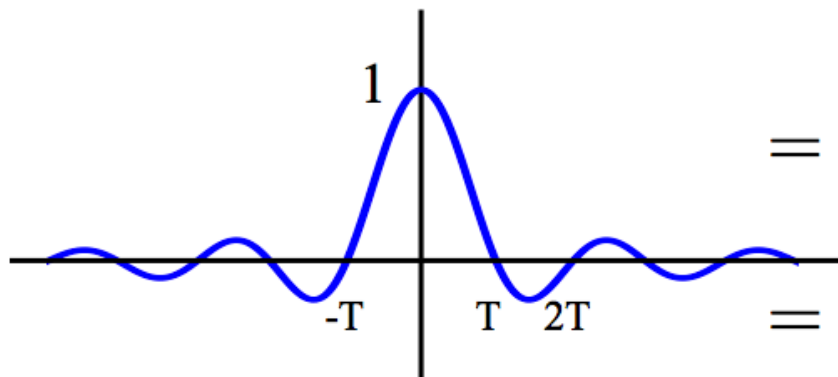


$$h_r(t) = \frac{1}{2\pi} \int_{-\Omega_s/2}^{\Omega_s/2} T e^{j\Omega t} d\Omega$$

$$= \frac{T}{2\pi} \frac{1}{jt} e^{j\Omega t} \Big|_{-\Omega_s/2}^{\Omega_s/2}$$

$$= \frac{T}{\pi t} \frac{e^{j\frac{\Omega_s}{2}t} - e^{-j\frac{\Omega_s}{2}t}}{2j}$$

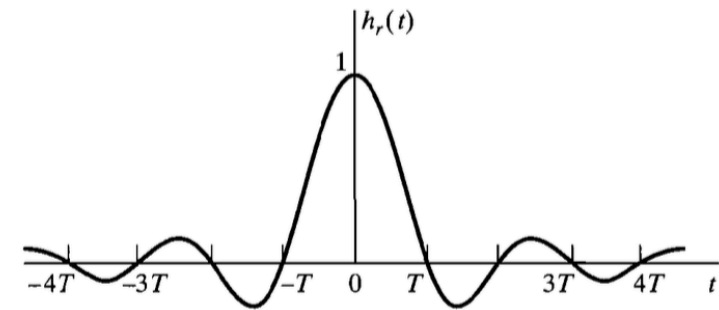
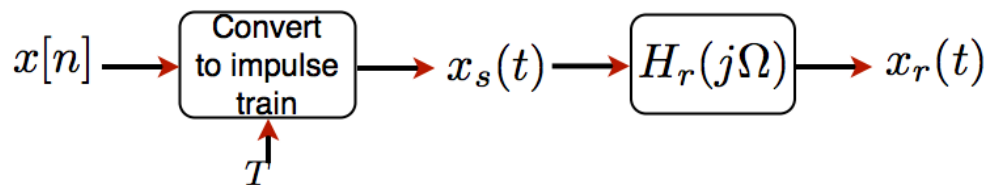
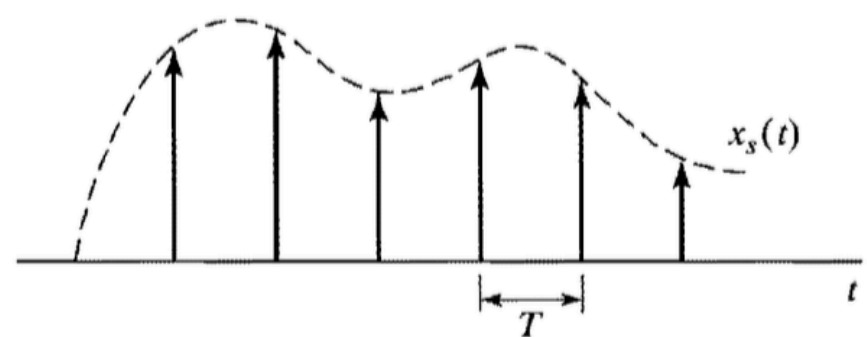
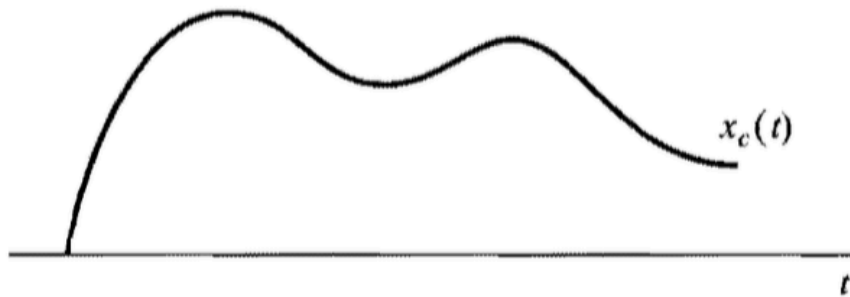
$$= \frac{T}{\pi t} \sin\left(\frac{\Omega_s}{2}t\right) = \frac{T}{\pi t} \sin\left(\frac{\pi}{T}t\right)$$



$$= \text{sinc}\left(\frac{t}{T}\right)$$

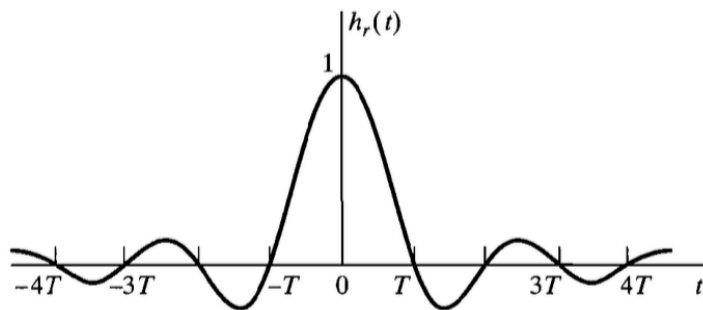
# Reconstruction in Time Domain

$$\begin{aligned}
 x_r(t) = x_s(t) * h_r(t) &= \left( \sum_n x[n] \delta(t - nT) \right) * h_r(t) \\
 &= \sum_n x[n] h_r(t - nT)
 \end{aligned}$$

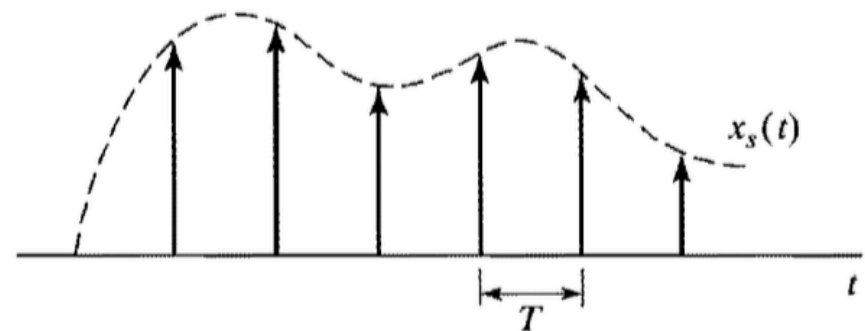


# Reconstruction in Time Domain

$$\begin{aligned}x_r(t) = x_s(t) * h_r(t) &= \left( \sum_n x[n] \delta(t - nT) \right) * h_r(t) \\&= \sum_n x[n] h_r(t - nT)\end{aligned}$$



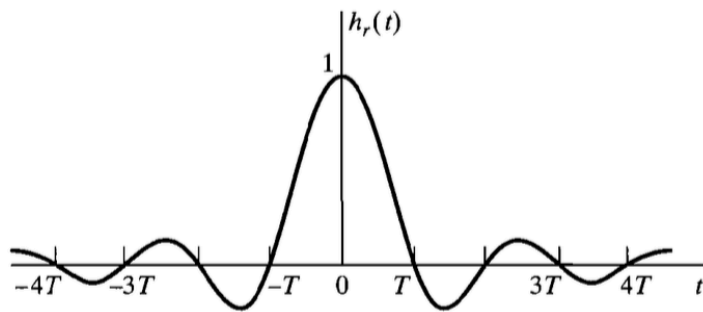
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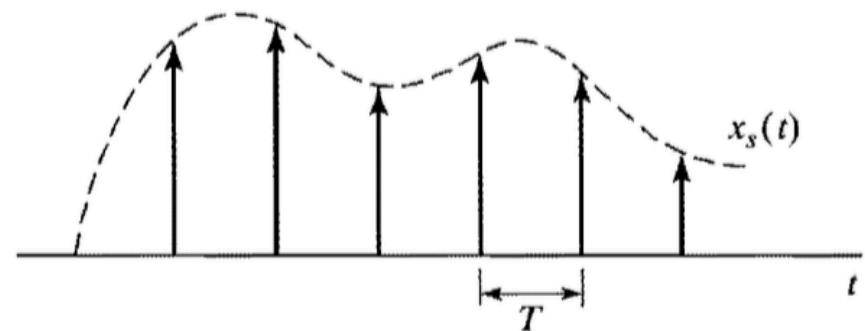
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# Reconstruction in Time Domain

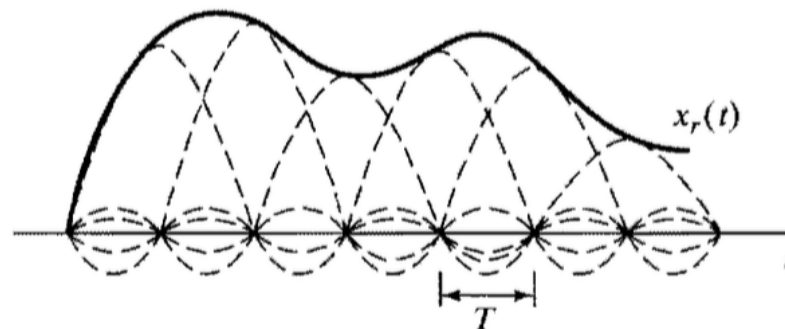
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 \end{aligned}$$



\*



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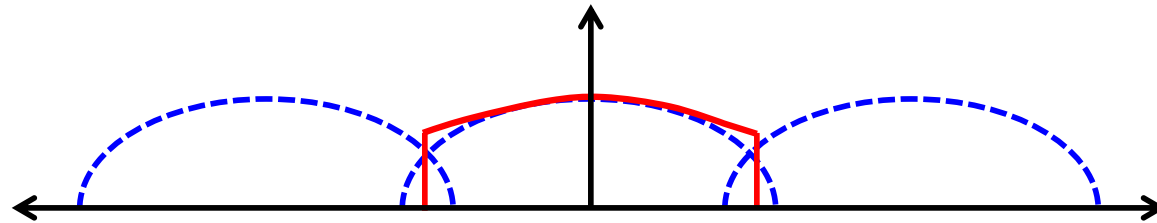
The sum of "sincs" gives  $x_r(t) \rightarrow$  unique signal that is bandlimited by sampling bandwidth





# Aliasing

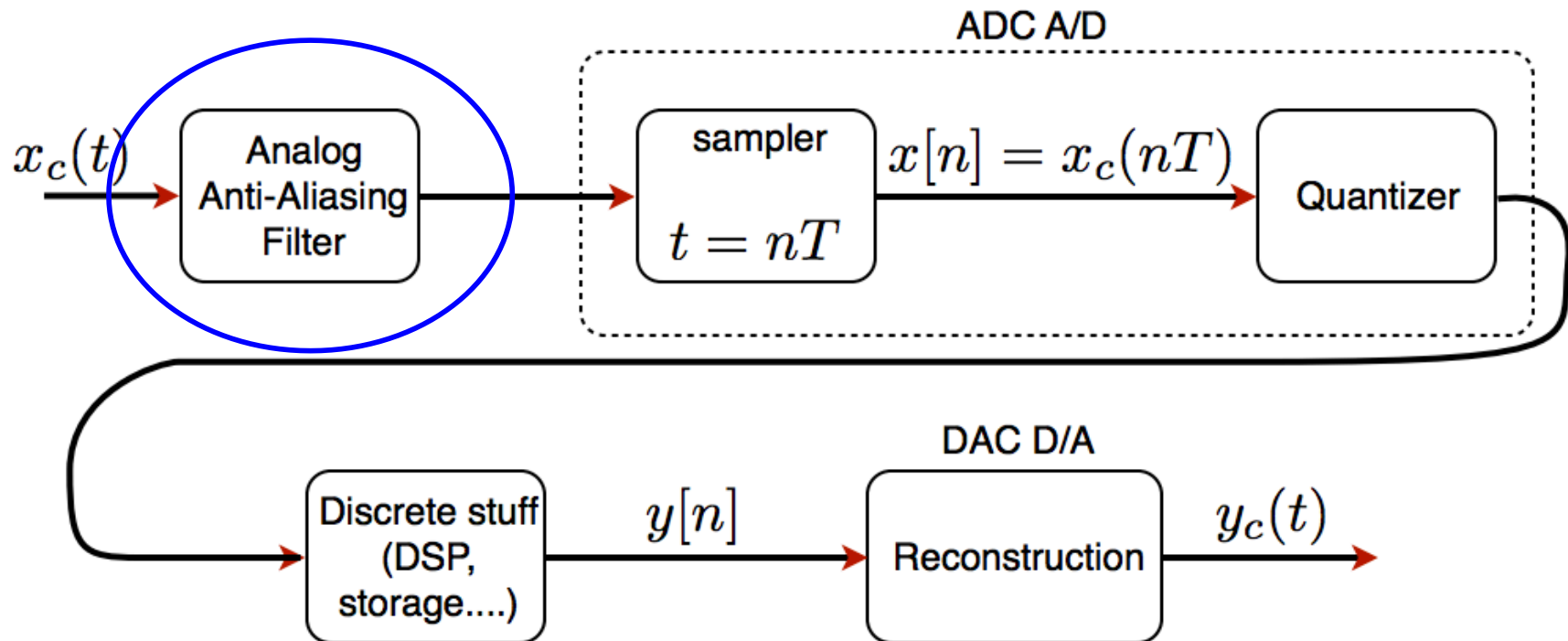
- If  $\Omega_N > \Omega_s/2$ ,  $x_r(t)$  is an aliased version of  $x_c(t)$



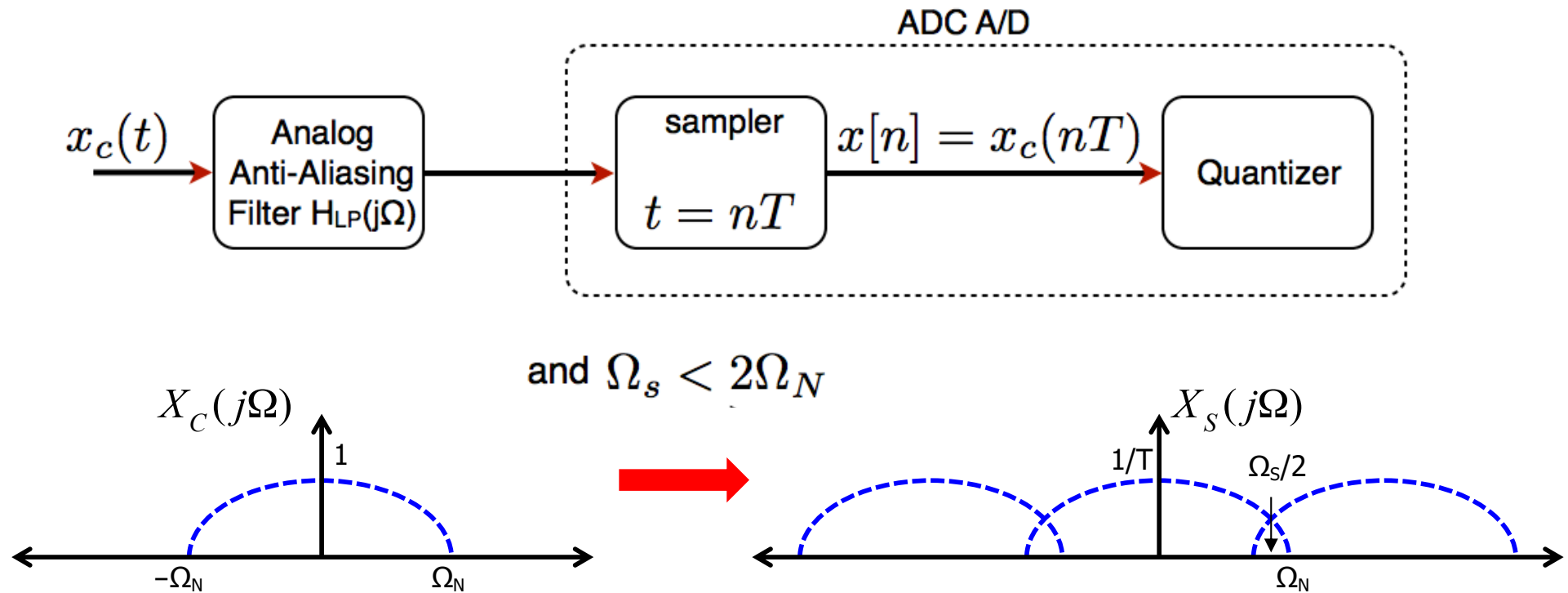
$$X_r(j\Omega) = \begin{cases} TX_s(j\Omega) & \text{if } |\Omega| \leq \Omega_s/2 \\ 0 & \text{otherwise} \end{cases}$$



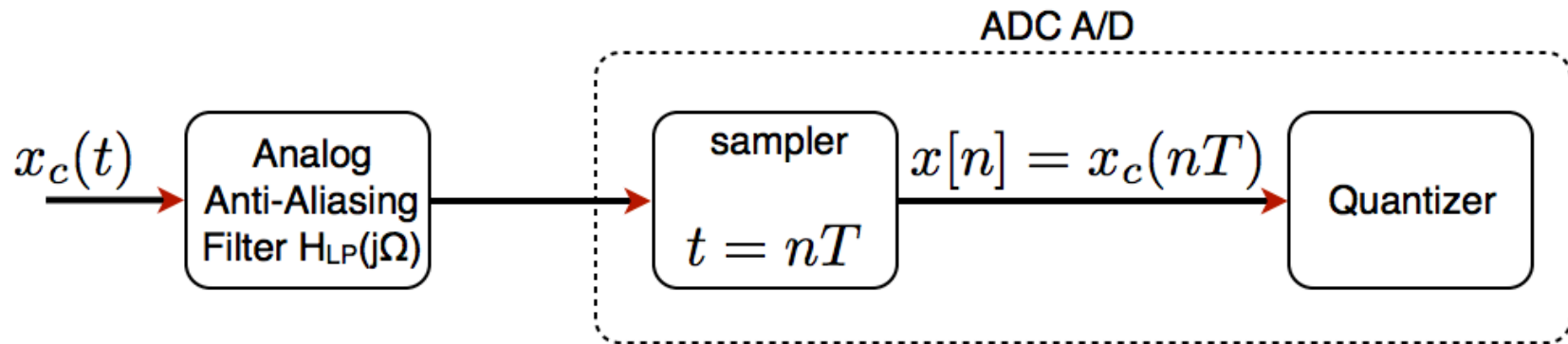
# DSP System



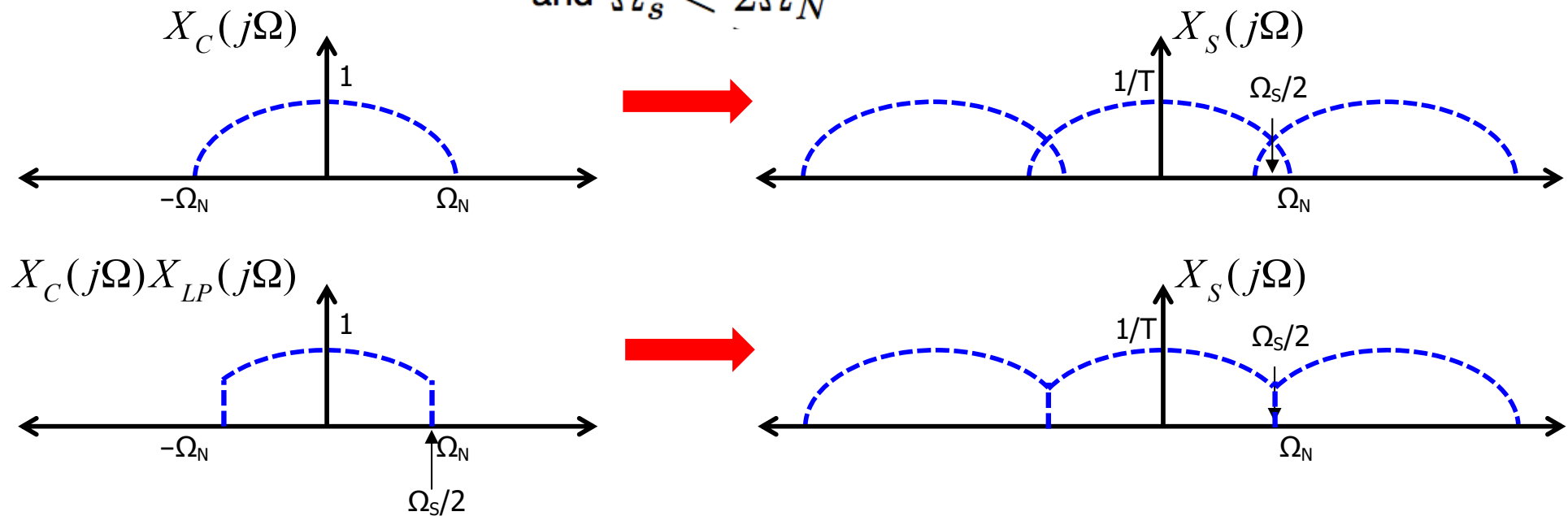
# Anti-Aliasing Filter



# Anti-Aliasing Filter



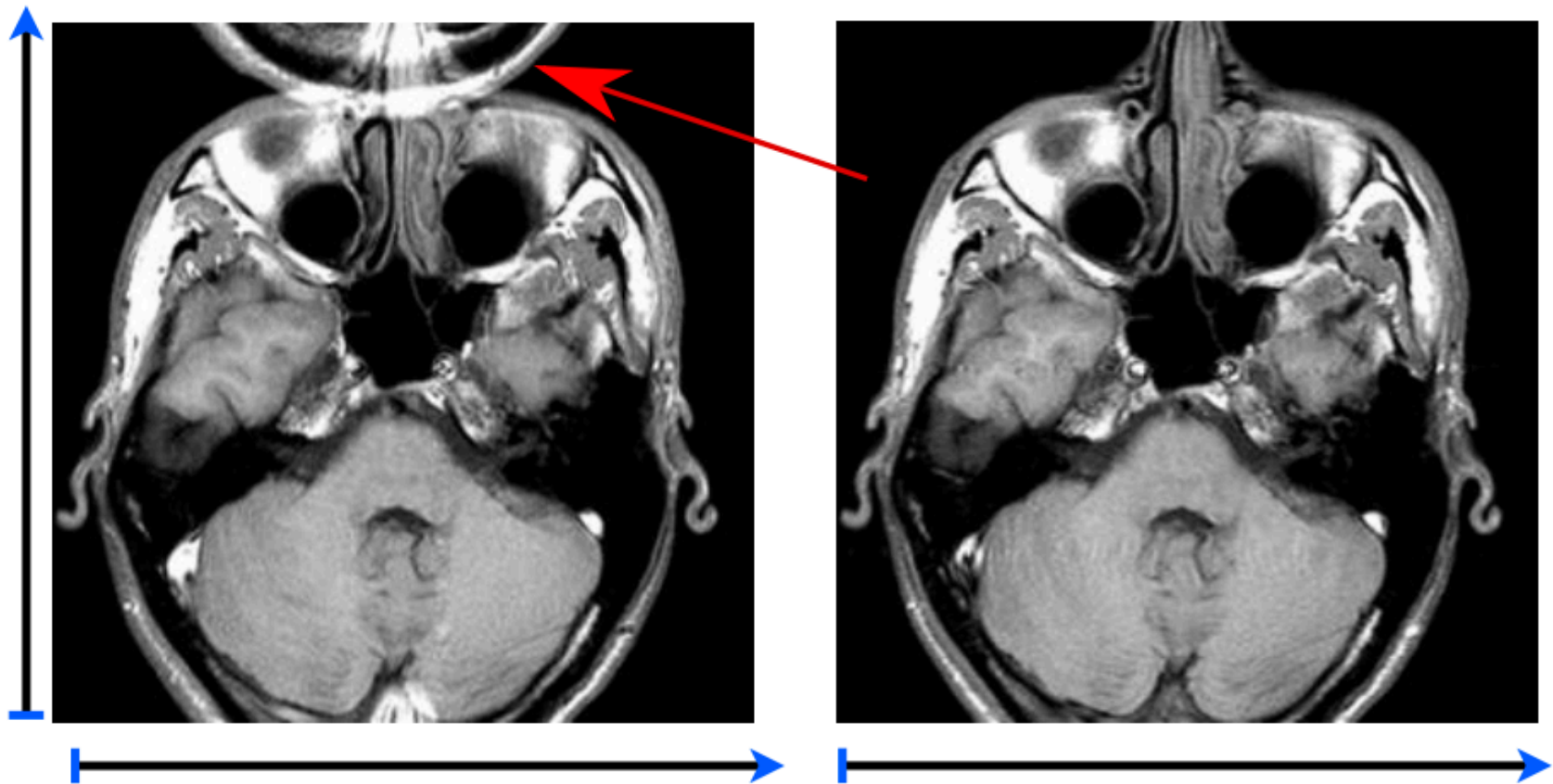
and  $\Omega_s < 2\Omega_N$





# MRI aliasing example

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# MRI anti-aliasing example

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# MRI anti-aliasing example

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# Big Ideas

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- ❑ Sampling
  - Ideal sampling modeled as impulsive sampling
  - Sample at Nyquist rate for recovery of unique bandlimited signal (i.e. avoid aliasing)
- ❑ Frequency Response of Sampled Signal
  - Sampled signal is period replicated input CT signal
- ❑ Reconstruction
  - Low pass/reconstruction filter results in sum of sines
- ❑ Anti-aliasing filtering
  - Force input signal to be bandlimited





# Admin

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- ❑ HW 2 out due Sunday 2/5 at midnight
  - Double check that your submission by viewing it!